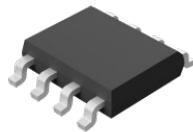


MLX90377 - Triaxis® Position Sensor IC

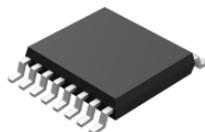
Datasheet

Features and Benefits

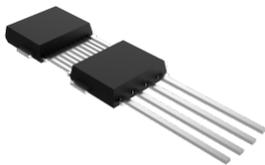
- Triaxis® Hall Technology
- On Chip Signal Processing for Robust Absolute Position Sensing
- Programmable Measurement Range
- Programmable Linear Transfer Characteristic (4 or 8 Multi-points or 16 or 32 Piece-Wise-Linear)
- Selectable (fast) SENT, PWM, SPC or Ratiometric Analog Output modes
- SAE J2716 APR2016 SENT
- ISO26262 **ASIL** READY BY MELIXIS
ASIL-C capable for (fast) SENT, PWM, SPC Output modes, and ASIL-B capable for analog output mode, Safety Element out of Context
- Enhanced serial data communication
- 48 bit ID Number option
- Single Die - SOIC-8, DMP-4, SMP-3, and Dual Die (Full Redundant) - TSSOP-16 package RoHS Compliant
- Robustness against stray-field



SOIC-8



TSSOP-16



DMP-4



SMP-3

Application Examples

- Absolute Rotary or Linear Position Sensor
- Pedal Position Sensor
- Throttle Position Sensor

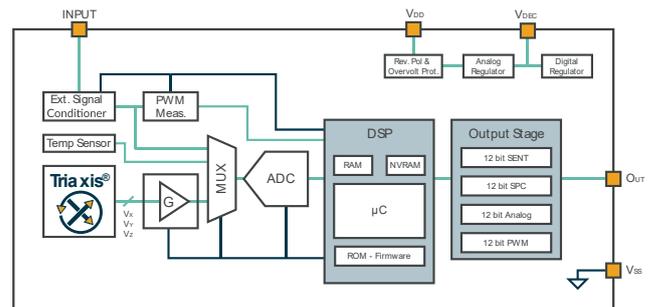
- Ride Height Position Sensor
- Steering Wheel Position Sensor
- Float-Level Sensor
- Non-Contacting Potentiometer
- Small Angle Position Sensor
- Small Stroke Position Sensor

Description

The MLX90377 is a monolithic magnetic position sensor IC. It consists of a Triaxis® Hall magnetic front end, an analog to digital signal conditioner, a DSP for advanced signal processing and an output stage driver.

The MLX90377 is sensitive to the three components of the magnetic flux density applied to the IC (i.e. Bx, By and Bz). This allows the MLX90377 with the correct magnetic circuit to decode the absolute position of any moving magnet (e.g. rotary position from 0 to 360 Degrees or linear displacement, see Figure 2). It enables the design of non-contacting position sensors that are frequently required for both automotive and industrial applications.

The MLX90377 provides 4 output modes. Firstly, the IC supports (fast) SENT and SPC frames encoded according to a Secure Sensor format. The output delivers enhanced serial messages providing error codes, and user-defined values. Through programming, the MLX90377 can also be configured to output a PWM (Pulse Width Modulated) signal. Finally, analog ratiometric output is also supported.



Ordering Information

| Product | Temperature | Package | Option Code | Packing Form | Definition |
|----------|-------------|---------|-------------|--------------|--|
| MLX90377 | G | DC | ADB-300 | RE | Standard/Legacy Mode, Analog/PWM output, Low Power Mode |
| MLX90377 | G | DC | ADB-310 | RE | Standard/Legacy Mode, Analog/PWM output, High Speed Mode |
| MLX90377 | G | DC | ADB-330 | RE | Standard / Legacy Mode, SENT output |
| MLX90377 | G | DC | ADB-380 | RE | Standard / Legacy Mode, SPC output |
| MLX90377 | G | GO | ADB-210 | RE | Linear Stray-Field Immune Mode, Analog/PWM output, High Speed Mode |
| MLX90377 | G | GO | ADB-230 | RE | Linear Stray-Field Immune Mode, SENT output |
| MLX90377 | G | GO | ADB-280 | RE | Linear Stray-Field Immune Mode, SPC output |
| MLX90377 | G | GO | ADB-300 | RE | Standard/Legacy Mode, Analog/PWM output, Low Power Mode |
| MLX90377 | G | GO | ADB-310 | RE | Standard/Legacy Mode, Analog/PWM output, High Speed Mode |
| MLX90377 | G | GO | ADB-330 | RE | Standard / Legacy Mode, SENT output |
| MLX90377 | G | GO | ADB-380 | RE | Standard / Legacy Mode, SPC output |
| MLX90377 | G | GO | ADB-510 | RE | Rotary Stray-Field Immune Mode, Analog/PWM output, High Speed Mode |
| MLX90377 | G | GO | ADB-530 | RE | Rotary Stray-Field Immune Mode (High Field variant), SENT output |
| MLX90377 | G | VS | ADB-300 | RE/RX | Standard/Legacy Mode, Analog/PWM output, Low Power Mode |
| MLX90377 | G | VS | ADB-303 | RE/RX | Standard/Legacy Mode, Analog/PWM output, Low Power Mode |
| MLX90377 | G | VS | ADB-308 | RE/RX | Standard/Legacy Mode, Analog/PWM output, Low Power Mode |
| MLX90377 | G | VS | ADB-310 | RE/RX | Standard/Legacy Mode, Analog/PWM output, High Speed Mode |
| MLX90377 | G | VS | ADB-313 | RE/RX | Standard/Legacy Mode, Analog/PWM output, High Speed Mode |
| MLX90377 | G | VS | ADB-318 | RE/RX | Standard/Legacy Mode, Analog/PWM output, High Speed Mode |
| MLX90377 | G | VS | ADB-330 | RE/RX | Standard / Legacy Mode, SENT output |

| Product | Temperature | Package | Option Code | Packing Form | Definition |
|----------|-------------|---------|-------------|--------------|---|
| MLX90377 | G | VS | ADB-333 | RE/RX | Standard / Legacy Mode, SENT output |
| MLX90377 | G | VS | ADB-338 | RE/RX | Standard / Legacy Mode, SENT output |
| MLX90377 | G | VE | ADB-110 | RE/RX | Rotary Stray-Field Immune Mode (Low Field variant), Analog/PWM output, High Speed Mode |
| MLX90377 | G | VE | ADB-210 | RE/RX | Linear Stray-Field Immune Mode, Analog/PWM output, High Speed Mode |
| MLX90377 | G | VE | ADB-230 | RE/RX | Linear Stray-Field Immune Mode, SENT output |
| MLX90377 | G | VE | ADB-300 | RE/RX | Standard/Legacy Mode, Analog/PWM output, Low Power Mode |
| MLX90377 | G | VE | ADB-310 | RE/RX | Standard/Legacy Mode, Analog/PWM output, High Speed Mode |
| MLX90377 | G | VE | ADB-330 | RE/RX | Standard / Legacy Mode, SENT output |
| MLX90377 | G | VE | ADB-510 | RE/RX | Rotary Stray-Field Immune Mode (High Field variant), Analog/PWM output, High Speed Mode |

Table 1 - Ordering Codes

| | |
|---|---|
| Temperature Code: | G: -40°C to 160°C |
| Package Code: | DC: SOIC-8 package (see 19.1) GO: TSSOP-16 package (see 19.2) VS: DMP-4 package (PCB-less dual mold, see 19.3) VE: SMP-3 package (PCB-less single mold, see 19.4) |
| Option Code - Chip revision | ADB-123: Chip Revision <ul style="list-style-type: none"> ▪ ADB : Standard revision |
| Option Code - Application | ADB-123: 1-Application - Magnetic configuration <ul style="list-style-type: none"> ▪ 1: Angular Rotary Strayfield Immune - Low field Variant ▪ 2: Linear position Strayfield Immune ▪ 3: Legacy / Angular Rotary / Linear position ▪ 5: Angular Rotary Strayfield Immune - High field Variant |
| Option Code - SW configuration: output mode, protocol | ADB-123: 2-SW configuration: output mode, protocol <ul style="list-style-type: none"> ▪ 0: Analog/PWM output mode, low power mode ▪ 1: Analog/PWM output mode, high speed mode ▪ 3: SENT output mode, 3μs tick time ▪ 8: SPC output mode, 1.5μs tick time <p>Important: Product must be ordered by the option code corresponding to the desired output protocol. The output protocols are not interchangeable by reprogramming at the end-user.</p> |
| Option Code - DMP-4 Trim & Form configuration | ADB-123: 3-DMP-4 Trim & Form configuration <ul style="list-style-type: none"> ▪ 0: Standard straight leads. See section 19.3.1 ▪ 3: Trim and Form STD2 2.54. See section 19.3.2 ▪ 8: Trim and Form STD4 2.54. See section 19.3.3 |
| Packing Form: | -RE : Tape & Reel <ul style="list-style-type: none"> ▪ VS:2500 pcs/reel ▪ VE:2500 pcs/reel ▪ DC:3000 pcs/reel ▪ GO:4500 pcs/reel -RX : Tape & Reel, similar to RE with parts face-down (VS and VE package) |
| Ordering Example: | MLX90377GDC-ADB-300-RE For a legacy version in SOIC-8 package, low-power analog output, delivered in Reel of 3000pcs. |

Table 2 - Ordering Codes Information

Contents

| | |
|---|----|
| Features and Benefits..... | 1 |
| Application Examples..... | 1 |
| Description..... | 1 |
| Ordering Information | 2 |
| 1. Functional Diagram and Application Modes..... | 7 |
| 2. Glossary of Terms | 8 |
| 3. Pin Definitions and Descriptions..... | 9 |
| 3.1. Pin Definition for SOIC-8 package..... | 9 |
| 3.2. Pin Definition for TSSOP-16 Dual-Die package | 10 |
| 3.3. Pin Definition for DMP-4 package | 10 |
| 3.4. Pin Definition for SMP-3 package | 11 |
| 4. Absolute Maximum Ratings..... | 12 |
| 5. Isolation Specification | 12 |
| 6. General Electrical Specifications..... | 13 |
| 7. Timing Specifications | 15 |
| 7.1. General Timing Specifications | 15 |
| 7.2. Timing Definitions..... | 15 |
| 7.3. Analog output timing specifications | 17 |
| 7.4. PWM output timing specifications | 18 |
| 7.5. SENT output timing specifications..... | 19 |
| 7.6. SPC Timing Specifications | 21 |
| 8. Magnetic Field Specifications | 23 |
| 8.1. Rotary Stray-field Immune Mode – Low Field Variant (-1xx code)..... | 23 |
| 8.2. Linear Stray-field Immune Mode (-2xx code) | 24 |
| 8.3. Standard/Legacy Mode (-3xx code)..... | 25 |
| 8.4. Rotary Stray-field Immune Mode - High Field Variant (-5xx code)..... | 26 |
| 9. Accuracy Specifications..... | 27 |
| 9.1. Definitions | 27 |
| 9.2. Rotary Stray-field Immune Mode – Low Field Variant (-1xx code)..... | 28 |
| 9.3. Linear Stray-field Immune Mode (-2xx Code)..... | 29 |
| 9.4. Standard/Legacy Mode (-3xx Code)..... | 30 |
| 9.5. Rotary Stray-field Immune Mode – High Field Variant (-5xx Code)..... | 31 |

| | |
|--|------------|
| 10. Memory Specifications | 31 |
| 11. Output Accuracy and Protocol | 32 |
| 11.1. Analog Output | 32 |
| 11.2. PWM (Pulse Width Modulation) Output | 32 |
| 11.3. SENT (Single Edge Nibble Transmission) Output | 33 |
| 11.4. SPC (Short PWM Code) Output..... | 41 |
| 12. End User Programmable Items | 50 |
| 13. End User Identification Items | 58 |
| 14. Description of End-User Programmable Items | 59 |
| 14.1. Output Modes | 59 |
| 14.2. Output Transfer Characteristic | 62 |
| 14.3. Sensor Front-End | 70 |
| 14.4. Filtering..... | 71 |
| 14.5. Programmable Diagnostics Settings | 73 |
| 15. Functional Safety | 75 |
| 15.1. Safety Manual..... | 75 |
| 15.2. Safety Mechanisms..... | 75 |
| 15.3. Fault Handling Time Interval..... | 79 |
| 16. Recommended Application Diagrams | 81 |
| 16.1. Wiring with the MLX90377 in SOIC-8 Package | 81 |
| 16.2. Wiring with the MLX90377 in TSSOP-16 Package..... | 82 |
| 16.3. Wiring with the MLX90377 in DMP-4 Package (built-in capacitors)..... | 83 |
| 16.4. Wiring with the MLX90377 in SMP-3 Package (built-in capacitors) | 84 |
| 17. Standard Information Regarding Manufacturability of Melexis Products with Different Soldering Processes | 85 |
| 18. ESD Precautions | 85 |
| 19. Package Information | 86 |
| 19.1. SOIC-8 Package | 86 |
| 19.2. TSSOP-16 Package | 89 |
| 19.3. DMP-4 - Package..... | 92 |
| 19.4. SMP-3 - Package | 97 |
| 19.5. Packages Thermal Performances..... | 101 |
| 20. Contact | 102 |
| 21. Disclaimer | 102 |

1. Functional Diagram and Application Modes

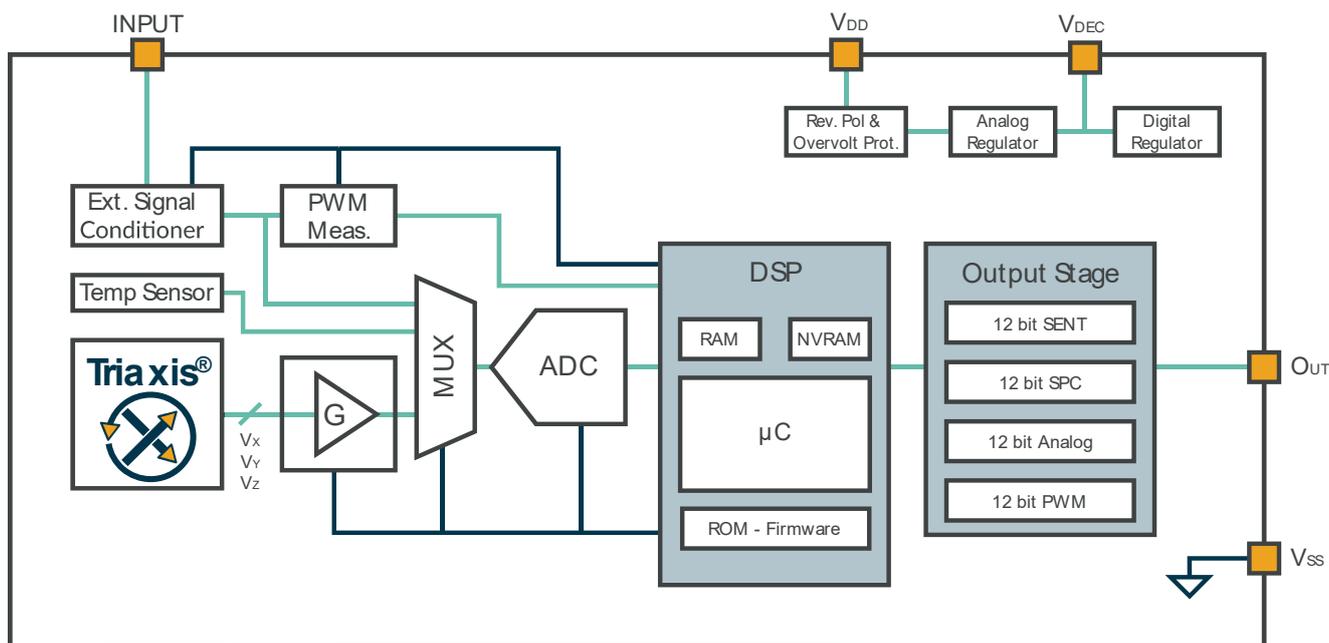


Figure 1 - MLX90377 Block diagram

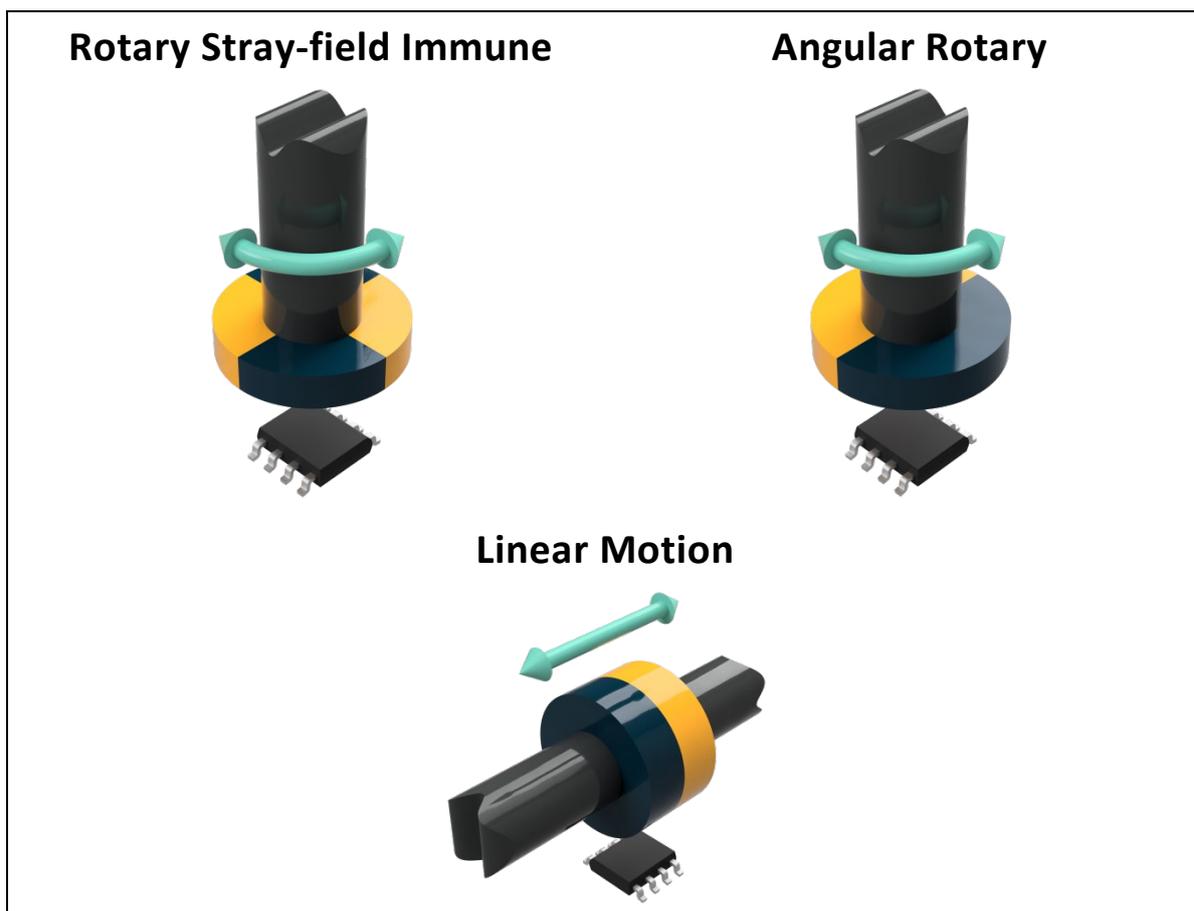


Figure 2 - Application Modes

2. Glossary of Terms

| Name | Description |
|-----------|--|
| ADC | Analog-to-Digital Converter |
| AoU | Assumption of Use |
| ASP | Analog Signal Processing |
| AWD | Absolute WatchDog |
| CPU | Central Processing Unit |
| CRC | Cyclic Redundancy Check |
| DC | Duty Cycle of the output signal i.e. $T_{ON} / (T_{ON} + T_{OFF})$ |
| DMP | Dual Mould Package |
| DP | Discontinuity Point |
| DSP | Digital Signal Processing |
| DTI | Diagnostic Test Interval |
| ECC | Error Correcting Code |
| EMA | Exponential Moving Average |
| EMC | Electro-Magnetic Compatibility |
| EoL | End of Line |
| FIR | Finite Impulse Response |
| FHTI | Fault Handling Time Interval |
| Gauss (G) | Alternative unit for the magnetic flux density (10G = 1mT) |
| HW | Hardware |
| IMC | Integrated Magnetic Concentrator |
| INL/DNL | Integral Non-Linearity / Differential Non-Linearity |
| IWD | Intelligent WatchDog |
| LNR | LiNeaRization |
| LSB/MSB | Least Significant Bit / Most Significant Bit |
| NC | Not Connected |
| NVRAM | Non Volatile RAM |
| POR | Power On Reset |
| PSF | Product Specific Functions |
| PWL | Piecewise Linear |
| PWM | Pulse Width Modulation |

| Name | Description |
|-----------|--|
| RAM | Random Access Memory |
| ROM | Read-Only Memory |
| SPC | Short PWM Code |
| SCN | Status & Communication Nibble |
| SEooC | Safety Element out of Context |
| TC | Temperature Coefficient (in ppm/°C) |
| Tesla (T) | SI derived unit for the magnetic flux density (Vs/m ²) |

Table 3 - Glossary of Terms

3. Pin Definitions and Descriptions

3.1. Pin Definition for SOIC-8 package

| Pin # | Name | Description |
|-------|------------------|--|
| 1 | V _{DD} | Supply |
| 2 | Input | For Melexis test or External Measurement |
| 3 | Test | For Melexis test |
| 4 | N.C. | Not connected |
| 5 | OUT | Output |
| 6 | N.C. | Not connected |
| 7 | V _{DEC} | Decoupling pin |
| 8 | V _{SS} | Ground |

Table 4 - SOIC-8 Pin definitions and descriptions

For optimal EMC behavior, it is recommended to connect the unused pins (N.C. and Test) to the Ground.

3.2. Pin Definition for TSSOP-16 Dual-Die package

| Pin # | Name | Description |
|-------|--------------------|--|
| 1 | V _{DEC1} | Decoupling pin die1 |
| 2 | V _{SS1} | Ground die1 |
| 3 | V _{DD1} | Supply die1 |
| 4 | Input ₁ | For Melexis test or External Measurement |
| 5 | Test ₂ | For Melexis test |
| 6 | OUT ₂ | Output die2 |
| 7 | N.C. | Not connected |
| 8 | N.C. | Not connected |
| 9 | V _{DEC2} | Decoupling pin die2 |
| 10 | V _{SS2} | Ground die2 |
| 11 | V _{DD2} | Supply die2 |
| 12 | Input ₂ | For Melexis test or External Measurement |
| 13 | Test ₁ | For Melexis test |
| 14 | N.C. | Not connected |
| 15 | OUT ₁ | Output die1 |
| 16 | N.C. | Not connected |

Table 5 - TSSOP-16 Pin definitions and descriptions

For optimal EMC behavior, it is recommended to connect the unused pins (N.C. and Test) to the Ground.

3.3. Pin Definition for DMP-4 package

DMP-4 package offers a pin to pin compatibility with the previous generation of Triaxis® products.

| Pin # | Name | Description |
|-------|-----------------|-------------|
| 1 | V _{SS} | Ground |
| 2 | V _{DD} | Supply |
| 3 | OUT | Output |
| 4 | V _{SS} | Ground |

Table 6 - DMP-4 Pin definitions and descriptions

3.4. Pin Definition for SMP-3 package

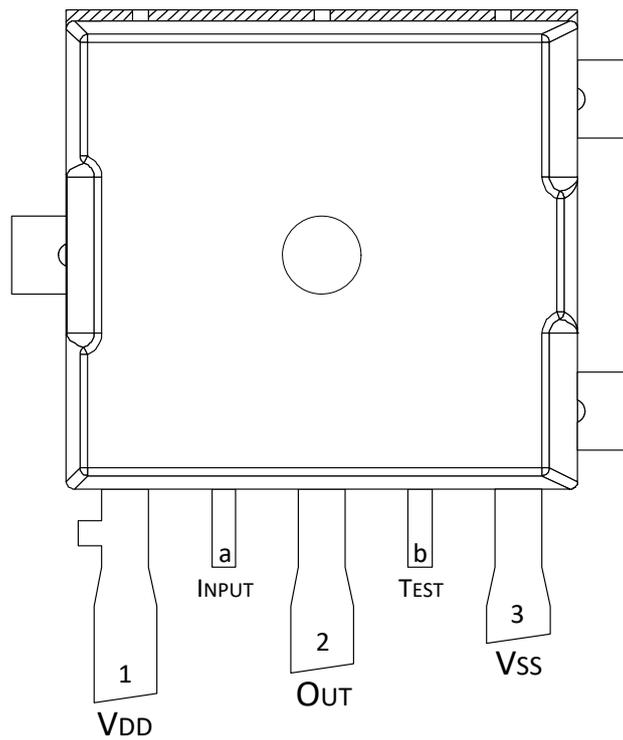


Figure 3 – Pins definition for SMP-3 Package

| Pin # | Name | Description |
|-------|-------|-------------|
| 1 | VDD | Supply |
| 2 | OUT | Output |
| 3 | VSS | Ground |
| a | INPUT | Test |
| b | TEST | Test |

Table 7 - SMP-3 Pins definition and description

4. Absolute Maximum Ratings

| Parameter | Symbol | Min | Typ | Max | Unit | Condition |
|---|------------------------|------------|-----|----------|------|----------------|
| Supply Voltage ⁽¹⁾ | V _{DD} | | | 28 37 | V | < 48h < 60s |
| Reverse Voltage Protection ⁽¹⁾ | V _{DD-rev} | -14 -18 | | | V | < 48h < 1h |
| Positive Output Voltage ⁽¹⁾ | V _{OUT} | | | 28 | V | < 48h |
| Reverse Output Voltage ⁽¹⁾ | V _{OUT-rev} | -14 -18 | | | V | < 48h < 1h |
| Internal Voltage | V _{DEC} | | | 3.6 | V | |
| | V _{DEC-rev} | -0.3 | | | V | |
| Positive Input pin Voltage | V _{Input} | | | 6 | V | |
| Reverse Input pin Voltage | V _{Input-rev} | -3 | | | V | |
| Operating Temperature | T _{AMB} | -40 | | +160 | °C | |
| Junction Temperature | T _J | | | +175 | °C | |
| Storage Temperature | T _{ST} | -55 | | +170 | °C | |
| Magnetic Flux Density | B _{max} | -1 | | 1 | T | |

Table 8 - Absolute maximum ratings

Exceeding any of the absolute maximum ratings may cause permanent damage.

Exposure to absolute maximum ratings conditions for extended periods may affect device reliability.

The values in Table 8 are valid for a single die, and will degrade when simultaneously applied to multiple dice in the same package.

5. Isolation Specification

Only valid for the TSSOP-16 package (code GO, i.e. dual die version).

| Parameter | Symbol | Min | Typ | Max | Unit | Condition |
|----------------------|-------------------|-----|-----|-----|------|---|
| Isolation Resistance | R _{isol} | 4 | - | - | MΩ | Between dice, measured between V _{SS1} and V _{SS2} with +/-20V bias |

Table 9 - Isolation specification

¹ Valid for full operating temperature range.

6. General Electrical Specifications

General electrical specifications are valid for temperature range [-40;160] °C and supply voltage range [4.5;5.5] V unless otherwise noted.

| Electrical Parameter | Symbol | Min | Typ | Max | Unit | Condition |
|--|----------------------------|-----------|--------------|-----------|-------------|--|
| Supply Voltage | V_{DD} | 4.5 | 5 | 5.5 | V | Powered by regulated supply |
| | V_{DD} | 6 | 12 | 18 | V | Powered by vehicle battery ⁽²⁾ |
| Supply Current for Legacy / Angular Rotary | I_{DD} | 8 | 9 | 11 | mA | -30x code |
| | | 10 | 11 | 13.5 | | -31x, -38x code |
| | | 8 | 9 | 11.5 | | -33x code |
| Supply Current for Strayfield Immune | I_{DD} | 12 9 | 13.5 11.5 | 15 13 | mA | -21x, -28x, -51x code -23x, -53x code |
| Start-up Level (rising) | $V_{DDstartH}$ | 3.95 | 4.1 | 4.25 | V | |
| Start-up Hysteresis | $V_{DDstartHyst}$ | - | 200 | - | mV | |
| PTC Entry Level (rising) | V_{PROV0} ⁽³⁾ | 6.3 | 6.55 | 6.8 | V | Powered by regulated supply |
| | V_{PROV1} | 21.5 | 23 | 24.5 | V | Powered by vehicle battery ⁽²⁾ |
| PTC Entry Level Hysteresis | $V_{PROV0Hyst}$ | 400 | 500 | 600 | mV | |
| Output Short Circuit Current | $I_{OUTshort}$ | -35 10 | | -10 35 | mA | $V_{out} = 0\text{ V}$ $V_{out} = 5\text{ V or }18\text{ V}$ |
| Output Load | R_L | 5 | 10 | | k Ω | Analog pull-up/pull-down |
| | | 5 | | 100 | k Ω | PWM push-pull, pull-up to 5V, PWM push-pull, pull-down to 0V |
| | | 10 | - | 55 | k Ω | SENT push-pull, pull-up ⁽⁴⁾ |
| | | 3 | - | 55 | k Ω | SPC push-pull and improved emission modes ⁽⁴⁾ |
| | | 1 | - | 100 | k Ω | Open drain pull-up/pull-down ⁽⁴⁾ |
| Digital open drain output level | $V_{satLoOd}$ | 0 | | 10 | % V_{ext} | Pull-up to any external voltage $V_{ext} \leq 18\text{ V}$, $I_L \leq 3.4\text{ mA}$ |
| | $V_{satHiOd}$ | 90 | | 100 | % V_{ext} | Pull-down to GND with any supply voltage $V_{DD} \leq 18\text{ V}$, $I_L \leq 3.4\text{ mA}$ |

² The battery mode is exclusively for the PWM protocol

³ Programming through Connector (PTC) requires raising supply voltage above V_{prov0} or V_{prov1} . This is customer configurable by setting a bit in the NVRAM

⁴ Output resistance should be selected together with the output capacitive load to correspondingly match the application, i.e. tick time, SPC ID, to allow appropriate time window for the trigger pulse reception. More details see sections 11.3.5 and 11.4.3.

| Electrical Parameter | Symbol | Min | Typ | Max | Unit | Condition |
|---|------------------------|--|--|--|------------------|--|
| Digital open drain output leakage ⁽⁵⁾ | I _{leakpuOd} | | | 400 | μA | V _{ext} ≤ 18V, 4.5V ≤ V _{DD} ≤ 5.5V, with a pull-up resistor |
| | | | | 100 | μA | V _{vdd} = V _{ext} , with a pull-up resistor |
| | I _{leakpdOd} | | | 150 | μA | with a pull-down resistor |
| Digital output Ron | R _{on} | 27 | 50 | 130 | Ω | Push-pull mode |
| Analog Saturation Output Level | V _{satA_lo} | | 0.5 ⁽⁶⁾ 3.3 ⁽⁶⁾ | 1.2 ⁽⁷⁾ 7.4 ⁽⁷⁾ | %V _{DD} | Pull-up load R _L ≥ 10 kΩ to 5 V Pull-up load R _L ≥ 5 kΩ to 18 V |
| | V _{satA_hi} | 97 ⁽⁷⁾ 95 ⁽⁷⁾ | 99 ⁽⁶⁾ 98 ⁽⁶⁾ | | %V _{DD} | Pull-down load R _L ≥ 10 kΩ Pull-down load R _L ≥ 5 kΩ |
| Digital Output Level | V _{satD_lopp} | | | 0.2 ⁽⁷⁾ 0.5 ⁽⁷⁾ | V | Pull-up load R _L ≥ 10 kΩ Pull-up load 10 kΩ > R _L ≥ 5 kΩ |
| | V _{satD_hipp} | V _{DD} - 0.2 ⁽⁷⁾ V _{DD} - 0.5 ⁽⁷⁾ | | | V | Pull-down load R _L ≥ 10 kΩ Pull-down load 10 kΩ > R _L ≥ 5 kΩ |
| Active Diagnostic Output Level | Diag_lo | | 0.5 ⁽⁶⁾ 3.3 ⁽⁶⁾ | 1.2 ⁽⁷⁾ 7.4 ⁽⁷⁾ | %V _{DD} | Pull-up load R _L ≥ 10 kΩ to 5 V Pull-up load R _L ≥ 5 kΩ to 18 V |
| | Diag_hi | 97 ⁽⁷⁾ 95 ⁽⁷⁾ | 99 ⁽⁶⁾ 98 ⁽⁶⁾ | | %V _{DD} | Pull-down load R _L ≥ 10 kΩ Pull-down load R _L ≥ 5 kΩ |
| Passive Diagnostic Output Level for Broken V _{SS} (Broken-Wire Detection) ⁽⁸⁾ | BV _{SSPD} | | 2.5 1 | 4 1.6 | %V _{DD} | Pull-down load R _L ≤ 25 kΩ Pull-down load R _L ≤ 10 kΩ |
| | BV _{SSPU} | 99.5 | 100 | | %V _{DD} | Pull-up load R _L ≥ 5 kΩ |
| Passive Diagnostic Output Level for Broken V _{DD} (Broken-Wire Detection) ⁽⁸⁾ | BV _{DDPD} | | 0 | 0.5 | %V _{DD} | Pull-down load R _L ≥ 5 kΩ |
| | BV _{DDPU} | 92.5 97 | 97.5 99 | | %V _{DD} | Pull-up load R _L ≤ 25 kΩ Pull-up load R _L ≤ 10 kΩ |
| Leakage current in fail-safe (HiZ) state ⁽⁹⁾ | I _{leak_HiZ} | | 10 ⁽¹⁰⁾ | 20 ⁽¹⁰⁾ | μA | Pull-up at output ⁽¹¹⁾ |

Table 10 – Electrical Specifications

⁵ The digital output level is thereby defined by the external voltage and pull-up or pull-down resistor.

⁶ at 27Deg.C and 5V supply voltage with typical process parameters

⁷ at 160Deg.C and 4.5V supply voltage with worst case (6 sigma) process parameters

⁸ Valid for dual-die configuration as well, i.e. TSSOP-16 package, where the two dies have the same supply and ground level, while the output of one die is connected with PU and the output of the other one is connected with PD.

⁹ In the case of open-drain mode, see the parameters I_{leakpuOd}/I_{leakpdOd}.

¹⁰ As the leakage current significantly increases at high temperature, the typical value is at 35Deg.C, and the maximum value is at 160Deg.C.

¹¹ The worst case scenario for the leakage occurs when the output is pulled up. The leakage current will be ~30 times smaller if the output is pulled down.

7. Timing Specifications

Timing specifications are valid for temperature range [-40; 160] °C and supply voltage range [4.5; 5.5] V unless otherwise noted.

7.1. General Timing Specifications

| Parameter | Symbol | Min. | Typ | Max. | Unit | Condition |
|---|---------------------|---------------|----------|---------------|------------|--|
| Main Clock Frequency ⁽¹²⁾ | F_{CK} | 11.4 22.8 | 12 24 | 12.6 25.2 | MHz | -23x, -30x, -33x, -53x code -21x, -28x, -31x, -38x, -51x code |
| Main Clock initial tolerances at room temperature | $\Delta F_{CK,0}$ | 11.75 23.5 | 12 24 | 12.25 24.5 | MHz | -23x, -30x, -33x, -53x code -21x, -28x, -31x, -38x, -51x code |
| Main Clock Frequency Thermal Drift | $\Delta F_{CK,T}$ | -2.5 | - | 2.5 | % F_{ck} | Relative to 35°C |
| Main Clock Frequency Total Drift | $\Delta F_{CK,TOT}$ | -3.5 | - | 3.5 | % F_{ck} | Relative tolerances, including thermal and lifetime drift, after factory trimming at Melexis |
| 1MHz Clock Frequency | F_{1M} | | 1 | | MHz | |

Table 11 - General Timing Specifications

7.2. Timing Definitions

7.2.1. Startup Time

In analog mode, the startup time is the time between the power on cycle and the time the first valid angle transmitted on the output. During startup, the sensor output is in High-Z state, and the driver is only enabled when the sensor is able to transmit a valid output.

In SPC mode, the startup time is the time when the device is able to receive the 1st trigger pulse with valid angle information. Prior to it, there is a programmable SPC-SCN_INIT parameter (see Table 49 in chapter 12) to report an error in the status and communication nibble (SCN, see section 11.4.5.3).

In SENT mode, the startup time consists of two values. The first one, T_{init} , is the time needed for the circuit to be ready to start acquiring an angle. In SENT mode, at that time, the IC starts transmitting initialisation frames. The second value, T_{stup} , is the time when the first valid angle is transmitted.

In PWM mode, startup is defined by three values, $T_{stup[1..3]}$. The first value is reached when the output is ready and starts to drive a voltage. The second value T_2 is the start of the first value angle transmission and the third one T_3 the moment the first angle has been transmitted.

¹² Including thermal and lifetime drift

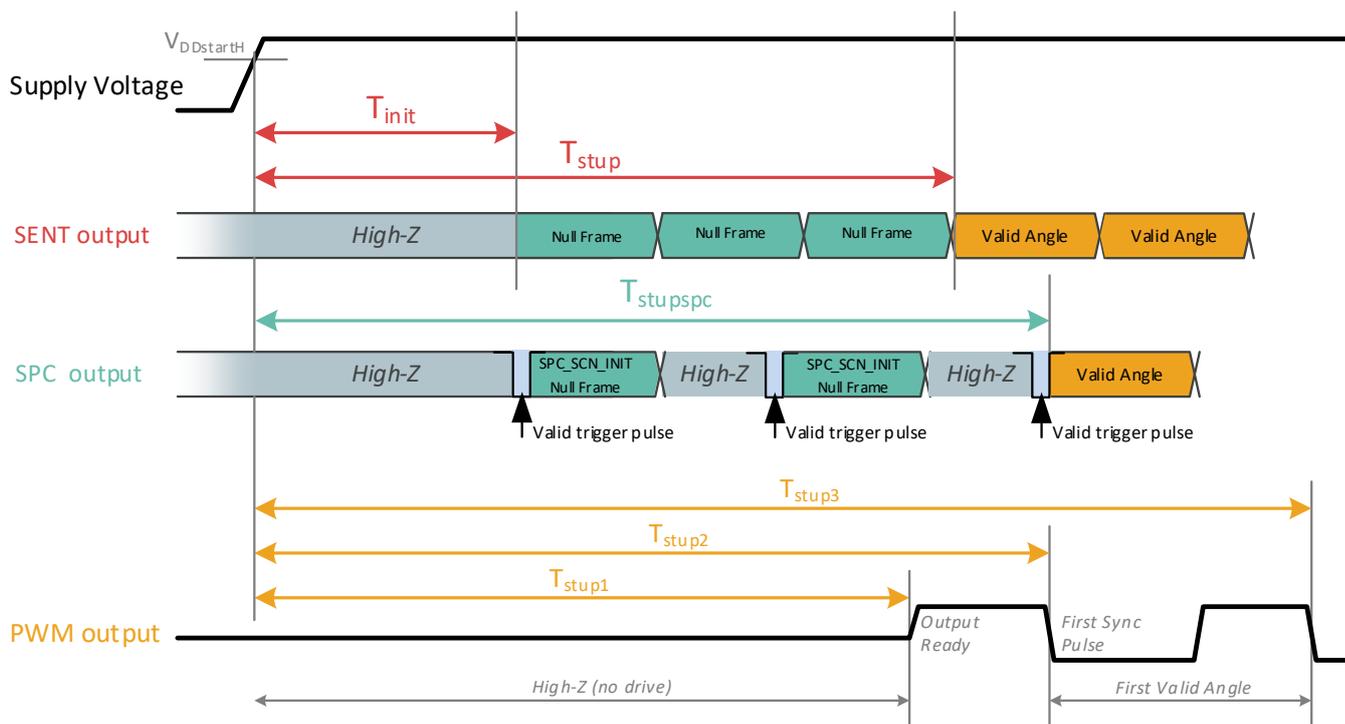


Figure 4 - Startup Time Definition

7.2.2. Latency (average)

Latency is the average lag between the movement of the detected object (magnet) and the response of the sensor output, as shown in Figure 5, where $\theta_{ECU}(t)$ is sampled as the beginning of the synchronization pulse. This value is representative of the time constant of the system for regulation calculations.

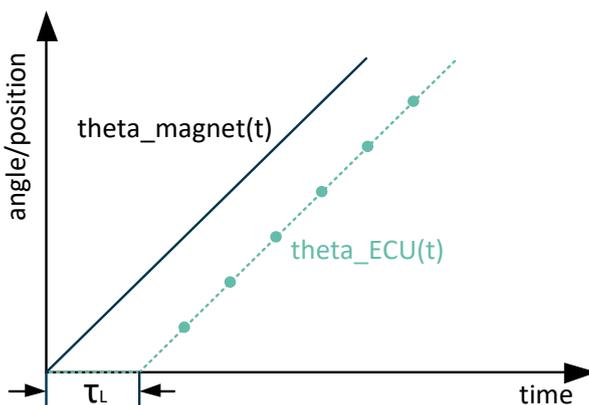


Figure 5 - Definition of Latency

7.2.3. Step Response (worst case)

The step response is a suitable metric for the "delay" of the sensor in case of an abrupt step in the magnetic change, considering 100% settling time without any DSP filter. Full settling is typically achieved in just two

steps. The sensor is asynchronous with the magnetic step change: the 100% settling time will fall in a time window; worst case is illustrated in the figure below.

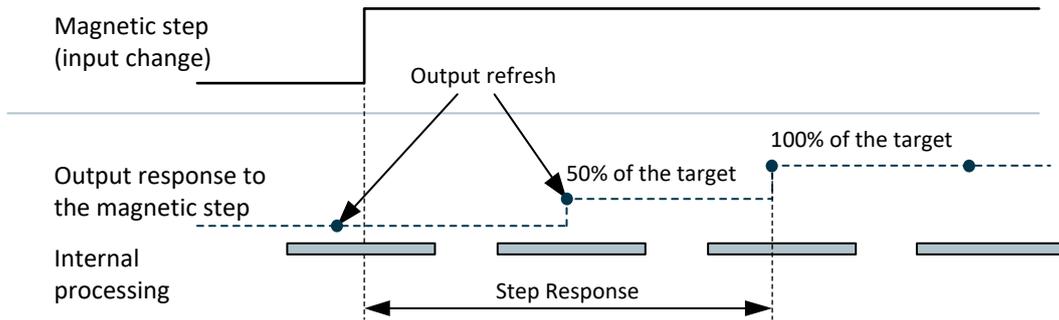


Figure 6 - Step Response Definition

7.3. Analog output timing specifications

For the analog output configurations, specifications are valid for the operational temperature range, and the supply voltage defined in section 6.

7.3.1. High Speed Mode (-21x,-31x,-51x code)

For the High Speed Mode, the timing specifications are listed in Table 12.

| Parameter | Symbol | Min | Typ | Max | Unit | Condition |
|---------------------------------------|-------------|-----|-----|------------|---------------|---|
| Output refresh period | τ_R | 190 | 200 | 210 | μs | |
| Latency | τ_L | | | 140 | μs | no external load |
| Step response | τ_S | | | 380 | μs | no external load |
| Start-up time | τ_{SU} | | | 3.2 | ms | |
| Slew Rate | SR | | | 120 200 | V/ms | capacitive load at output is 100nF capacitive load at output is 10nF |
| Analog Diagnostics Cycle | DTI_{ANA} | | | 7.2 | ms | see Table 73, section 15.3 |
| Digital Diagnostics Cycle | DTI_{DIG} | | | 6.1 | ms | see Table 73, section 15.3 |
| Safe startup Time ⁽¹³⁾⁽¹⁴⁾ | | | | 13.3 | ms | |

Table 12 - Analog General Timing Specifications for High Speed Mode

¹³ Fully programmable depending on the power/magnetic/output mode.

¹⁴ Time between reset due to digital fault to first valid data transmission. Min. value defined by $OUT_DIAG_HIZ_TIME$ (see Table 42 in chapter 12 for details).

7.3.2. Low Power Mode (-30x code)

For the Standard/Legacy Mode, the timing specifications are listed in Table 13:

| Parameter | Symbol | Min | Typ | Max | Unit | Condition |
|---------------------------------------|-------------|-----|-----|------------|---------------|---|
| Output refresh period | τ_R | 380 | 400 | 420 | μs | |
| Latency | τ_L | | | 300 | μs | no external load |
| Step response | τ_S | | | 680 | μs | no external load |
| Start-up time | τ_{SU} | | 4 | 4.2 | ms | |
| Slew Rate | SR | | | 120 200 | V/ms | capacitive load at output is 100nF capacitive load at output is 10nF |
| Analog Diagnostics Cycle | DTI_{ANA} | | | 15 | ms | see Table 73, section 15.3 |
| Digital Diagnostics Cycle | DTI_{DIG} | | | 9.8 | ms | see Table 73, section 15.3 |
| Safe startup Time ⁽¹³⁾⁽¹⁴⁾ | | | | 21.7 | ms | |

Table 13 - Analog General Timing Specifications for the Standard/Legacy Low Power Mode

7.4. PWM output timing specifications

The factory calibrated parts with order code -x0x and -x1x default to ratiometric analog output, the end user must reprogram the part to use the PWM output. Details can be found in Table 39. The Table 14 below shows the timing specifications for PWM output.

| Parameter | Symbol | Min | Typ | Max | Unit | Condition |
|---|--------------------|------|-----|------|---------------|---|
| Internal angle measurement period (-x0x code) | τ_R | 380 | 400 | 420 | μs | |
| Internal angle measurement period (-x1x code) | τ_R | 190 | 200 | 210 | μs | |
| PWM Frequency | F_{PWM} | 100 | | 2000 | Hz | Details see section 14.1.4 |
| PWM Frequency Initial Tolerances | $\Delta F_{PWM,0}$ | -2 | | 2 | % F_{PWM} | $T=35^\circ\text{C}$, can be trimmed at EOL |
| PWM Frequency Thermal Drift | $\Delta F_{PWM,T}$ | -2.5 | | 2.5 | % F_{PWM} | |
| PWM Frequency Drift | ΔF_{PWM} | -5.0 | | 5.0 | % F_{PWM} | Over temperature and lifetime |
| PWM startup Time (up to output ready) ⁽¹⁵⁾ | T_{stup1} | | 3.2 | | ms | |
| PWM startup Time (up to first sync. Edge) ⁽¹⁵⁾ | T_{stup2} | | 4.2 | | ms | $T_{stup1} + T_{PWM}$, $F_{PWM} = 1\text{kHz}$ |

¹⁵ Typical value specified according to the typical PWM frequency. Max. value can be obtained by scaling with the PWM frequency drift accordingly.

| Parameter | Symbol | Min | Typ | Max | Unit | Condition |
|--|----------------------|-----|-----|-----|------|---|
| PWM startup Time (up to first data received) ⁽¹⁵⁾ | T _{stup3} | | 5.2 | | ms | T _{stup1} + 2 * T _{PWM} , F _{PWM} = 1kHz ⁽¹⁶⁾ |
| Rise Time PWM push-pull mode ⁽¹⁷⁾ | T _{rise_pp} | | 3.1 | 5.0 | μs | VDD in the range of [4.5V, 5.5V] |
| Fall Time PWM push-pull mode ⁽¹⁷⁾ | T _{fall_pp} | | 2.6 | 4.5 | μs | VDD in the range of [4.5V, 5.5V] |
| Fall Time PWM NMOS open-drain mode ^{(17) (18)} | T _{fall_od} | | | 200 | μs | VDD=4.5V, V _{pull-up} = 18V, R _L = 10kΩ ⁽¹⁹⁾ |

Table 14 - PWM timing specifications with order code

7.5. SENT output timing specifications

The general SENT timing specifications are listed in Table 15.

| Parameter | Symbol | Min | Typ | Max | Unit | Condition |
|---|-------------------|-----|------------|-----|------|---|
| Tick time ⁽²⁰⁾ | | | 3 | 6 | μs | 3μs = Normal SENT (default) 6μs = Slow SENT |
| SENT edge rise Time | T _{rise} | | 12.5 | 18 | μs | between 1.1V and 3.8V |
| SENT edge fall Time | T _{fall} | | 5.3 | 6.5 | μs | for 3μs tick time |
| Slow Message cycle length (3μs tick time) | | | 791 475 | | ms | Extended sequence (40 frames) Short sequence (24 frames) |

Table 15 - SENT General Timing Specifications

7.5.1. Standard/Legacy Mode (-x3x code)

7.5.1.1. Continuous Synchronous Acquisition Mode

For the SENT output protocol in the Standard/Legacy Mode, MLX90377 with the default factory calibration operates with constant SENT frame length (SENT with pause), where a constant latency and step response time is guaranteed. The length of the SENT frame is defined by the parameter T_FRAME in number of ticks (see Table 49). The sensor is configured in continuous synchronous timing mode, acquires angles based on the output frequency, and the sensor output has a fixed frame frequency.

The detailed default setting of the SENT protocol for order code -33x is as follows:

¹⁶ First frame transmitted has no synchronization edge; Therefore the second frame transmitted is the first complete one.

¹⁷ 10nF capacitive load, calculated between 10%-90% voltage level.

¹⁸ Rise time in PWM NMOS open-drain mode is not specified as it depends on external components and the pull-up voltage.

¹⁹ V_{pull-up} stands for the external pull up voltage. The worst-case scenario for this parameter is in the NMOS open-drain mode, where there is the maximal voltage difference between V_{pull-up} and VDD. Furthermore, this parameter highly depends on the V_{pull-up}, the capacitive and resistive load at the output. The rise/fall time should be adjusted accordingly, with a different V_{pull-up} value and RC time constant resulting from the output load.

²⁰ The device also supports SENT formats with other tick time, see Table 31.

- Protocol: SENT with pause
- Tick time: 3µs/tick
- SENT format: A.3 (H.4)
- Number of conversions per SENT frame: 2
- ADC clock frequency: 3MHz
- DSP linearization: 16 points

The corresponding timing specifications are listed in Table 16.

| Parameter | Symbol | Min | Typ | Max | Unit | Condition |
|---------------------------------------|--------------|-------------------------|------|------|-------|--|
| SENT frame tick count | N_{Tframe} | 320 ⁽²¹⁾⁽²²⁾ | | | ticks | |
| Output refresh period | τ_R | | 0.96 | | ms | |
| SENT startup time | T_{stup} | | | 6.4 | ms | Until first valid angle received |
| Average Latency | T_{latcy} | | | 1.7 | ms | Filter = 0 (no filter) ⁽²³⁾ |
| | | | | 1.9 | | Filter = 1 (FIR11), default setting |
| | | | | 2.5 | | Filter = 2 (FIR1111) |
| Step Response (worst case) | T_{wcStep} | | 2.24 | 3.0 | ms | Filter = 0 (no filter) ⁽²³⁾ |
| | | | 2.71 | 3.5 | | Filter = 1 (FIR11), default setting |
| | | | 3.62 | 4.4 | | Filter = 2 (FIR1111) |
| Analog Diagnostics Cycle | DTI_{ANA} | | | 10.4 | ms | see Table 73, section 15.3 |
| Digital Diagnostics Cycle | DTI_{DIG} | | | 8.5 | ms | see Table 73, section 15.3 |
| Safe startup Time ⁽¹³⁾⁽¹⁴⁾ | | | | 12.1 | ms | |

Table 16 - Synchronous SENT Mode Timing Specifications

7.5.1.2. Continuous Asynchronous Acquisition Mode

Optionally to the default factory setting of the -x3x code, the MLX90377 can be configured to operate with variable SENT frame length (SENT without pause), where latency and step response is dependent on the transmitted data. The length of the SENT frame is defined by the data content. The sensor is configured in the continuous asynchronous acquisition mode and periodically acquires a magnetic measurement asynchronously to the SENT transmission.

When configured in the continuous asynchronous acquisition mode, the detailed default setting of the SENT protocol for order code -x3x is as follows:

- Protocol: SENT without pause
- Tick time: 3µs/tick
- SENT format: A.3 (H.4)
- ADC clock frequency: 3MHz
- DSP linearization: 16 points

²¹ Defines the typical Output refresh period (SENT Frame Period).

²² If the sensor is configured from the default factory setting of -33x code to have 1 conversion per SENT frame, the minimum tick count is 282.

²³ See section 14.4 for details concerning Filter parameter. It is an option to improve the speed, but will degrade noise performance (see section 9.4) by a factor of 1.4.

The corresponding timing specifications are listed in Table 17.

| Parameter | Symbol | Min | Typ | Max | Unit | Condition |
|--|---------------------|-----|-----|-----|---------------|--|
| Internal angle measurement period | T_R | 475 | 500 | 525 | μs | Asynchronously to the output protocol |
| SENT startup time | T_{stup} | | | 5.9 | ms | Until first valid angle received |
| Average Latency | T_{latcy} | | | 1.8 | ms | Filter = 0 (no filter) ⁽²³⁾ |
| | | | | 2 | | Filter = 1 (FIR11), default setting |
| | | | | 2.5 | | Filter = 2 (FIR1111) |
| Step Response (worst case) | T_{wcStep} | | | 2.2 | ms | Filter = 0 (no filter) ⁽²³⁾ |
| | | | | 2.8 | | Filter = 1 (FIR11), default setting |
| | | | | 3.7 | | Filter = 2 (FIR1111) |
| Analog Diagnostics Cycle | DTI_{ANA} | | | 11 | ms | see Table 73, section 15.3 |
| Digital Diagnostics Cycle | DTI_{DIG} | | | 8.8 | ms | see Table 73, section 15.3 |
| Safe startup Time ^{(13) (14)} | | | | 18 | ms | |

Table 17 - Synchronous SENT Mode Timing Specifications

7.6. SPC Timing Specifications

In SPC mode, the MLX90377 starts data acquisition once the trigger pulse has been received, regardless of the configured mode. It will send the acquired data in the same SENT frame. This feature is available for any tick time greater than or equal to $1.5\mu\text{s}$. Please check the section 11.4 for more details on the configuration options.

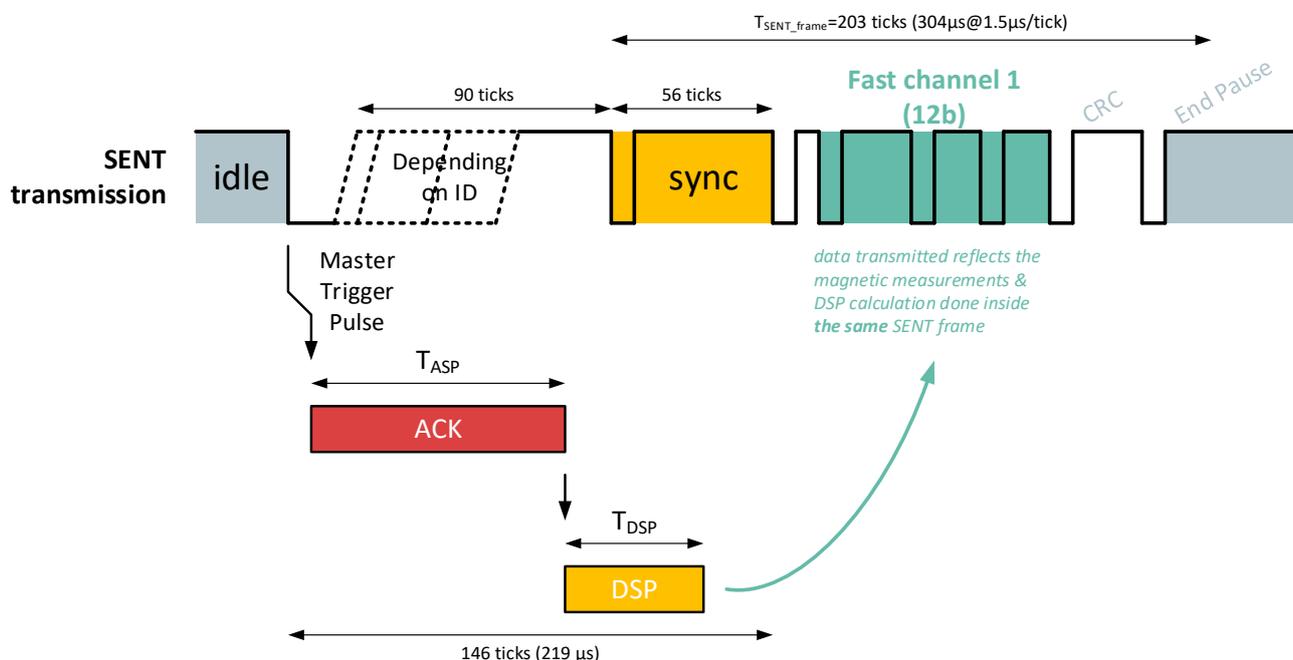


Figure 7 – SPC timing illustration in 1.5µs tick time mode and H.2 format

| Parameter | Symbol | Min | Typ | Max | Unit | Condition |
|---|---------------|------|------|------|---------------|---|
| Time between trigger received & acquisition started | T_{tr2acq} | | 10 | | μs | Tick: 1.5 μs SPC_RX_FILT_TIME ⁽²⁴⁾ : 6 SPC_TRIG_TH: 0 |
| Startup time | $T_{stupspc}$ | | 3.9 | | ms | |
| Latency | T_{latcy} | | 450 | | μs | |
| Step Response ⁽²⁵⁾ | T_{wcStep} | | 1 | | ms | |
| Jitter between the acquisition | J_{acq} | | | 5 | μs | |
| Inter-die synchronicity ⁽²⁶⁾ | | 8.45 | 9.08 | 9.71 | μs | |
| Analog Diagnostics Cycle ⁽²⁷⁾ | DTI_{ANA} | | | 9 | SPC triggers | |
| Digital Diagnostics Cycle ⁽²⁷⁾ | DTI_{DIG} | | | 9 | SPC triggers | |

Table 18 - SPC Mode Timing Specifications

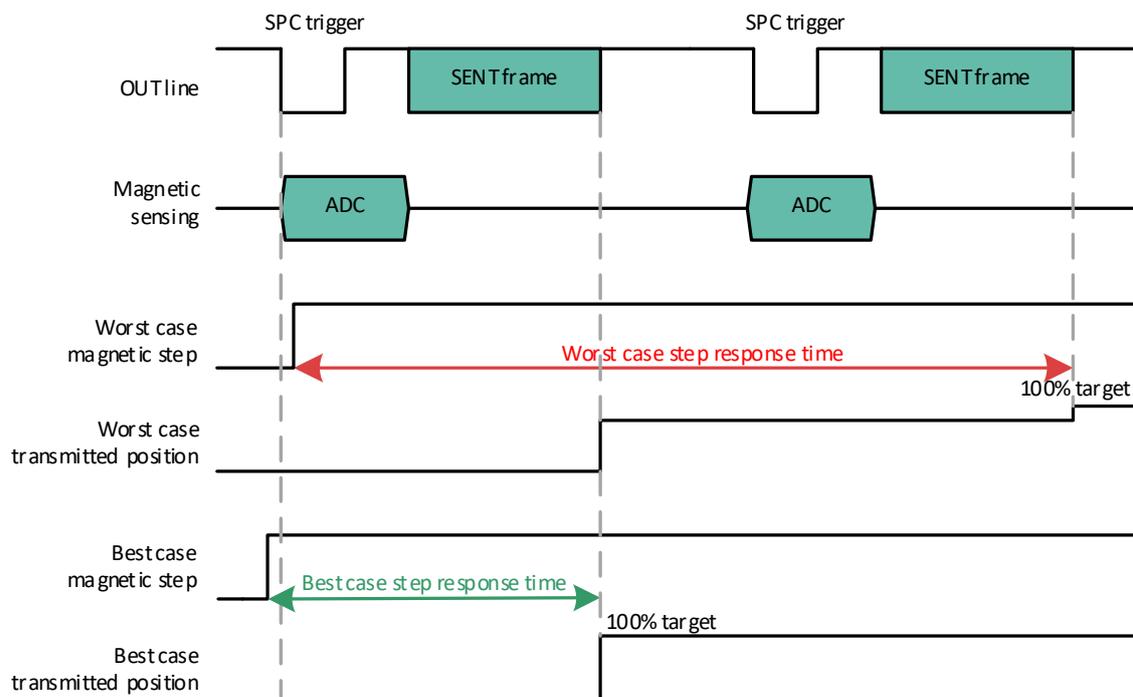


Figure 8 – Illustration of the best and worst case step response in SPC mode

²⁴ SPC_RX_FILT_TIME is the parameter for the SPC trigger pulse filtering. Its unit is tick time. It must be aligned with the longest NIBBLE low time on the bus, in order to separate SENT pulses from SPC trigger pulses (NIBBLE_PULSE_FIXED[]).

²⁵ The worst-case step response time is specified, which is equal to twice a SPC trigger plus its SENT response time, when the field change happens right after the trigger pulse, see Figure 8. It is then 1 ms if the SPC trigger period is 0.5 ms.

²⁶ The absolute time difference of the magnetic information acquisition between the two dice in bus transmission mode, see section 11.4, based on 1.5 μs tick time

²⁷ Diagnostics response time, detailed description see section 15.2, Table 71, e.g. with 500 μs ECU frame time, the max. value is 4.5ms

8. Magnetic Field Specifications

Magnetic field specifications are valid for temperature range [-40; 160] °C unless otherwise noted.

8.1. Rotary Stray-field Immune Mode – Low Field Variant (-1xx code)

| Parameter | Symbol | Min | Typ | Max | Unit | Condition |
|--|-----------------------------------|-------------------|-------|--------------------|------------------------|---|
| Number of magnetic poles | N_P | 4 ⁽²⁸⁾ | - | - | | |
| Magnetic Flux Density in X-Y plane | B_X, B_Y ⁽²⁹⁾ | | | 25 ⁽³⁰⁾ | mT | $\sqrt{B_X^2 + B_Y^2}$ this is not the useful signal |
| Magnetic Flux Density in Z | B_Z | | | 100 | mT | this is not the useful signal |
| Magnetic in-plane gradient of in-plane field component | $\frac{\Delta B_{XY}}{\Delta XY}$ | 3.8 | 10 | | $\frac{mT}{mm}$ | $\frac{1}{2} \sqrt{\left(\frac{dB_X}{dX} - \frac{dB_Y}{dY}\right)^2 + \left(\frac{dB_X}{dY} + \frac{dB_Y}{dX}\right)^2}$ this is the useful signal |
| Magnet Temperature Coefficient | TC_m | -2400 | | 0 | $\frac{ppm}{^\circ C}$ | |
| Field Strength Resolution ⁽³¹⁾ | $\frac{\Delta B_{XY}}{\Delta XY}$ | 0.075 | 0.100 | 0.125 | $\frac{mT}{mm}$ LSB | Magnetic field gradient norm (12bits data) |
| Field too Low Threshold | B_{TH_LOW} | 0.8 | 1.2 | ⁽³²⁾ | $\frac{mT}{mm}$ | Typ is recommended value to be set by user (see 14.5.4) |
| Field too High Threshold ⁽³³⁾ | B_{TH_HIGH} | 70 | 100 | 102 | $\frac{mT}{mm}$ | |

Table 19 - Magnetic specification for rotary stray-field immune- low field variant

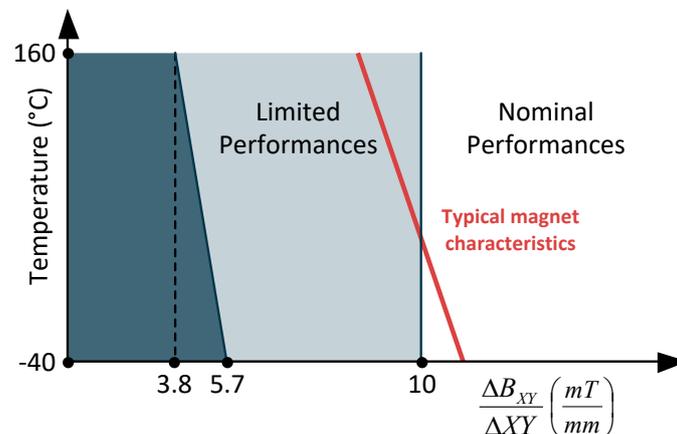


Figure 9 - Minimum useful signal definition for rotary stray-field immune application-low field variant

²⁸ Due to 4 poles magnet usage, maximum angle measurement range is limited to 180°

²⁹ The condition must be fulfilled for all combinations of B_X and B_Y .

³⁰ Above this limit, the IMC® starts to saturate, yielding to an increase of the linearity error.

³¹ Only valid with default `MAGNET_SREL_T[1..7]` configuration

³² Higher values of Field too Low threshold are not recommended by Melexis and shall only been set in accordance with the magnetic design and taking a sufficient safety margin to prevent false positive

³³ Due to the saturation effect of the IMC, the FieldTooHigh monitor detects only defects in the sensor.

Nominal performances apply when the useful signal $\Delta B_{XY}/\Delta XY$ is above the typical specified limit. Under this value, limited performances apply. See section 9.2 for accuracy specifications.

8.2. Linear Stray-field Immune Mode (-2xx code)

| Parameter | Symbol | Min | Typ | Max | Unit | Condition |
|---|----------------------------------|-------|-------------------|--------------------|-------------------------------------|--|
| Number of magnetic poles | N_p | | 2 | - | | Linear movement |
| Magnetic Flux Density in X | B_x | | | 80 ⁽³⁰⁾ | mT | $B_y \leq 20\text{mT}$ |
| Magnetic Flux Density in X-Y | B_x, B_y ⁽²⁹⁾ | | | 70 ⁽³⁰⁾ | mT | $\sqrt{B_x^2 + B_y^2}, B_y > 20\text{mT}$ |
| Magnetic Flux Density in Z | B_z | | | 100 | mT | |
| Magnetic gradient of X-Z field components | $\frac{\Delta B_{XZ}}{\Delta X}$ | 3 | 6 ⁽³⁴⁾ | | $\frac{\text{mT}}{\text{mm}}$ | $\sqrt{\left(\frac{\Delta B_x}{\Delta X}\right)^2 + \left(\frac{1}{G_{IMC}} \frac{\Delta B_z}{\Delta X}\right)^2}$ ⁽³⁵⁾ |
| Distance between the two IMC® | ΔX | | 1.91 | | | See chapter 19 for magnetic center definitions |
| IMC gain | G_{IMC} | | 1.19 | | | See ⁽³⁵⁾ |
| Magnet Temperature Coefficient | TC_m | -2400 | | 0 | $\frac{\text{ppm}}{^\circ\text{C}}$ | |
| Field Strength Resolution ⁽³¹⁾ | $\frac{\Delta B_{XZ}}{\Delta X}$ | 0.037 | 0.05 | 0.063 | $\frac{\text{mT}}{\text{mm LSB}}$ | Magnetic field gradient norm expressed in 12bits words |
| Field too Low Threshold | B_{TH_LOW} | 0.2 | 1.2 | ⁽³²⁾ | $\frac{\text{mT}}{\text{mm}}$ | Typ is recommended value to be set by user (see 14.5.4) |
| Field too High Threshold ⁽³³⁾ | B_{TH_HIGH} | 35 | 50 | 51 | $\frac{\text{mT}}{\text{mm}}$ | |

Table 20 - Magnetic specifications for linear stray-field application

Nominal performances apply when the useful signal $\Delta B_{xz}/\Delta x$ and temperature ranges are inside the values defined in the following Figure 10. At higher temperature or lower field gradients, the accuracy of MLX90377 is degraded and Limited Performances apply, as described in Table 24 in section 9.3.

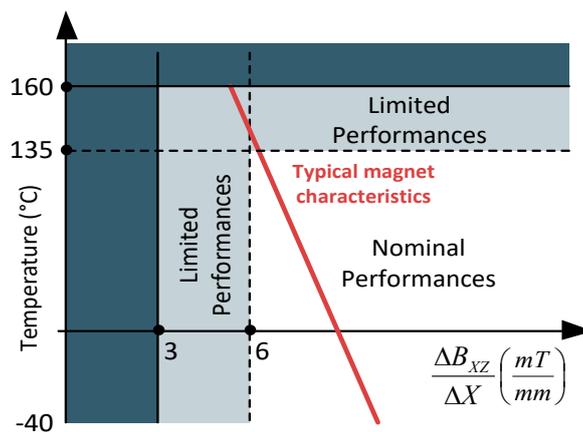


Figure 10 – Minimum useful signal definition for linear stray-field immune application

³⁴ Below this value, the performances are degraded due to a reduction of the signal-to-noise ratio, signal-to-offset ratio.

³⁵ IMC has better performance for concentrating in-plane (x-y) field components, resulting in a better magnetic sensitivity. A correction factor, called IMC gain has to be applied to the z field component to account for this difference.

8.3. Standard/Legacy Mode (-3xx code)

| Parameter | Symbol | Min. | Typ. | Max. | Unit | Condition |
|---|-------------------|--------------------|-------|--------------------|------------------------|---|
| Number of magnetic poles | N_p | - | 2 | - | | |
| Magnetic Flux Density in X-Y plane | $B_x, B_y^{(29)}$ | | | 70 ⁽³⁰⁾ | mT | $\sqrt{B_x^2 + B_y^2}$ |
| Magnetic Flux Density in Z | B_z | | | 100 | mT | in absolute value |
| Useful Magnetic Flux Density Norm | B_{Norm} | 10 ⁽³⁴⁾ | 20 | | mT | $\sqrt{B_x^2 + B_y^2}$ (x-y mode) $\sqrt{B_x^2 + \left(\frac{1}{G_{IMC}} B_z\right)^2}$ (x-z mode) $\sqrt{B_y^2 + \left(\frac{1}{G_{IMC}} B_z\right)^2}$ (y-z mode) see 14.3.1 for sensing mode description. |
| IMC gain | G_{IMC} | | 1.19 | | | see ⁽³⁵⁾ |
| Magnet Temperature Coefficient | TC_m | -2400 | | 0 | $\frac{ppm}{^\circ C}$ | |
| Field Strength Resolution ⁽³¹⁾ | B_{Norm} | 0.075 | 0.100 | 0.125 | $\frac{mT}{LSB}$ | Magnetic field gradient norm expressed in 12bits words |
| Field Too Low Threshold | B_{TH_LOW} | 0.4 | 4.0 | ⁽³²⁾ | mT | Typ is recommended value to be set by user (see 14.5.4) |
| Field Too High Threshold ⁽³³⁾ | B_{TH_HIGH} | 70 | 100 | 100 | mT | |

Table 21 - Magnetic specifications for Standard application

Nominal performances apply when the useful signal B_{Norm} is above the typical specified limit. Under this value, limited performances apply. See Table 25 in section 9.4 for accuracy specifications.

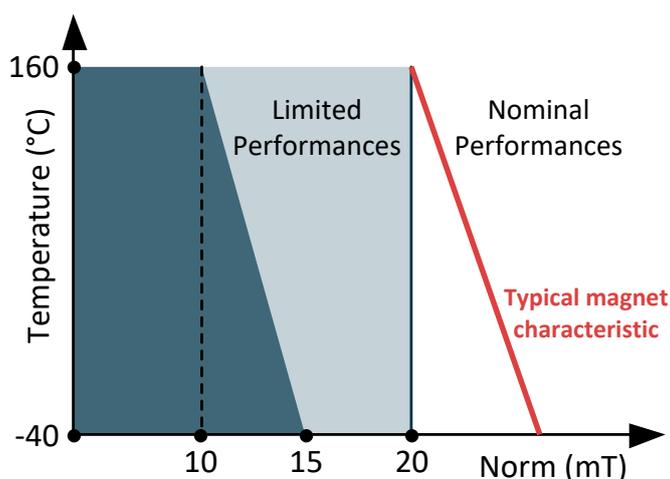


Figure 11 - Minimum useful signal definition for Standard/Legacy application

8.4. Rotary Stray-field Immune Mode- High Field Variant (-5xx code)

| Parameter | Symbol | Min | Typ | Max | Unit | Condition |
|--|-----------------------------------|-------------------|-------|--------------------|-----------------------------|--|
| Number of magnetic poles | N_p | 4 ⁽²⁸⁾ | - | - | | |
| Magnetic Flux Density in X-Y plane | B_x, B_y ⁽²⁹⁾ | | | 67 ⁽³⁰⁾ | mT | $\sqrt{B_x^2 + B_y^2}$ this is not the useful signal |
| Magnetic Flux Density in Z | B_z | | | 100 | mT | this is not the useful signal |
| Magnetic in-plane gradient of in-plane field component | $\frac{\Delta B_{XY}}{\Delta XY}$ | 8.25 | 21 | | $\frac{mT}{mm}$ | $\frac{1}{2} \sqrt{\left(\frac{dB_x}{dx} - \frac{dB_y}{dy}\right)^2 + \left(\frac{dB_x}{dy} + \frac{dB_y}{dx}\right)^2}$ this is the useful signal. |
| Magnet Temperature Coefficient | TC_m | -2400 | | 0 | $\frac{ppm}{^\circ C}$ | |
| Field Strength Resolution ⁽³¹⁾ | $\frac{\Delta B_{XY}}{\Delta XY}$ | 0.075 | 0.100 | 0.125 | $\frac{mT}{mm \text{ LSB}}$ | Magnetic field gradient norm (12bits data) |
| Field too Low Threshold | B_{TH_LOW} | 1.2 | 2 | ⁽³²⁾ | $\frac{mT}{mm}$ | Typ is recommended value to be set by user (see 14.5.4) |
| Field too High Threshold ⁽³³⁾ | B_{TH_HIGH} | 80 | 100 | 102 | $\frac{mT}{mm}$ | |

Table 22 - Magnetic specification for rotary stray-field immune

See section 9.5 for accuracy specifications.

9. Accuracy Specifications

Accuracy specifications are valid for temperature range [-40;160] °C and supply voltage range [4.5; 5.5] V unless otherwise noted.

9.1. Definitions

This section defines several parameters, which will be used for the magnetic specifications.

9.1.1. Intrinsic Linearity Error

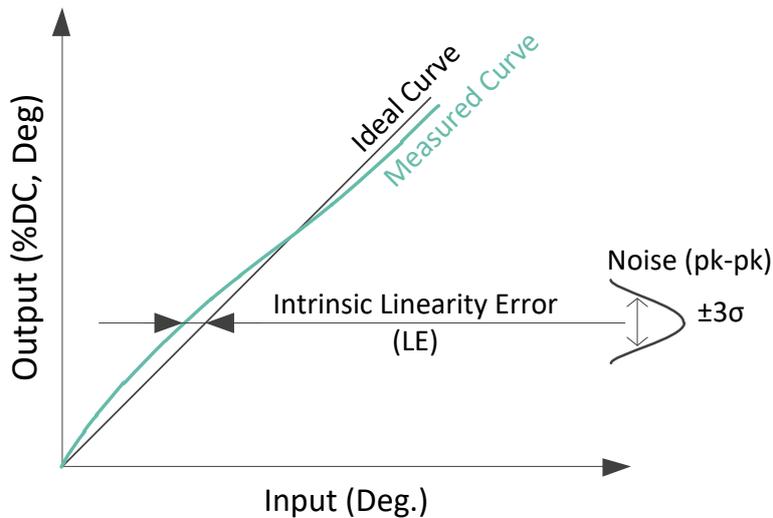


Figure 12 - Sensor accuracy definition

Figure 12 depicts the intrinsic linearity error in new parts. The Intrinsic Linearity Error refers to the IC itself (offset, sensitivity mismatch, orthogonality) considering an ideal magnetic field. Once associated to a practical magnetic construction and the associated mechanical and magnetic tolerances, the output linearity error increases. However, it can be significantly reduced with the multi-point end-user calibration (see section 14.2).

9.1.2. Total Angle Drift

After calibration, the output angle of the sensor might still change due to temperature change, aging, etc. This is defined as the total drift $\partial\theta_{TT}$:

$$\partial\theta_{TT} = |\max\{\theta(\theta_{IN}, T, t) - \theta(\theta_{IN}, T_{RT}, t_0)\}|$$

where θ_{IN} is the input angle, T is the temperature, T_{RT} is the room temperature, and t is the elapsed lifetime after calibration. t_0 represents the status at the start of the operating life. Note the total drift $\partial\theta_{TT}$ is always defined with respect to the angle at room temperature. In this datasheet, T_{RT} is typically defined at 35°C, unless stated otherwise. The total drift is valid for all angles along the full mechanical range. The total drift is expressed as an unsigned value, to indicate the ideal minimum drift should be 0. In reality, the drift can happen in both positive and negative directions.

9.2. Rotary Stray-field Immune Mode – Low Field Variant (-1xx code)

Valid before EoL calibration and for all applications under the conditions described in section 8.1.

| Parameter | Symbol | Min | Typ | Max | Unit | Condition |
|---|----------------------------|-----|-----|------|------|--|
| XY - Intrinsic Linearity Error | Le_{xy} | -1 | | 1 | Deg. | |
| Noise ⁽³⁶⁾ | | | | 0.55 | Deg. | Filter = 0 (FIR1), for nominal performance, see section 8.1 |
| | | | | 1 | Deg. | Filter = 0 (FIR1), for limited performance, see section 8.1 |
| | | | | 0.7 | Deg. | Filter = 1 (FIR1), for limited performance, see section 8.1 |
| | | | | 0.5 | Deg. | Filter = 2 (FIR1), for limited performance, see section 8.1 |
| XY - Total Drift | $\partial\theta_{TT_{XY}}$ | | | 0.85 | Deg. | |
| Hysteresis | | | 0.1 | 0.2 | Deg. | |
| Output Stray Field Immunity ⁽³⁷⁾ | $\partial\theta_{FF}$ | | | 0.6 | Deg. | In accordance of ISO 11452-8, at 30 Deg.C, with 10mT/mm magnetic In-plane gradient of In-plane field component, and stray-field strength of 4000A/m from any direction |

Table 23 - Rotary stray-field immune magnetic performance – Low Field Variant

³⁶ $\pm 3\sigma$

³⁷ Stray field induced error is linearly proportional to the stray field strength

9.3. Linear Stray-field Immune Mode (-2xx Code)

Valid before EoL calibration and for all applications under the conditions described in section 8.2.

| Parameter | Symbol | Min | Typ | Max | Unit | Condition |
|---|---------------------------|------|------------|---------------------|------|---|
| XZ - Intrinsic Maximum Error | L_{E_XZ} | -2.5 | ± 1.25 | 2.5 | Deg. | for nominal performance, see section 8.2 |
| | | -4 | ± 2 | 4 | | for limited performance, see section 8.2 |
| Noise in high speed mode (-21x code) ⁽³⁶⁾ | | | | 0.3 | Deg. | Filter = 1, 6mT/mm |
| | | | | 0.45 | | Filter = 0, 6mT/mm |
| | | | | 0.9 | | Filter = 0, 3mT/mm |
| Noise in low power mode (-23x code) ⁽³⁶⁾ | | | | 0.65 | Deg. | Filter = 0, 3mT/mm |
| XZ - Total Drift | $\partial\theta_{TT_XZ}$ | | | 0.8 ⁽³⁸⁾ | Deg. | for nominal performance, see section 8.2 |
| | | | | 1.4 ⁽³⁸⁾ | | for limited performance, see section 8.2 |
| Hysteresis | | | | 0.1 | Deg. | 6mT/mm gradient field |
| Output Stray Field Immunity ⁽³⁷⁾ | $\partial\theta_{FF}$ | | | 0.8 | Deg. | In accordance of ISO 11452-8, at 30 Deg.C, with 6mT/mm magnetic In-plane gradient of In-plane field component, and stray-field strength of 4000A/m from any direction |

Table 24 - Linear stray-field immune magnetic performances

³⁸ Valid for perfect field gradient. In most applications, an extra error caused by the bias field needs to be added. See Application note [Dual-Disk Linear Stray field Robust Position Sensing on melexis.com](#) for more information.

9.4. Standard/Legacy Mode (-3xx Code)

Valid before EoL calibration and for all applications under the conditions described in section 8.3.

| Parameter | Symbol | Min | Typ | Max | Unit | Condition |
|--|---------------------------|------|------------|------|------|--|
| XY - Intrinsic Linearity Error | L_{E_XY} | -1 | | 1 | Deg. | |
| XZ - Intrinsic Linearity Error | L_{E_XZ} | -2.5 | ± 1.25 | 2.5 | Deg. | |
| YZ - Intrinsic Linearity Error | L_{E_YZ} | -2.5 | ± 1.25 | 2.5 | Deg. | |
| Noise for nominal performance in section 8.3 in low power mode (-30x, -33x code) ⁽³⁶⁾ | | | 0.05 | 0.1 | Deg. | Filter = 0, $\sqrt{B_x^2 + B_y^2} > 40\text{mT}$ |
| | | | 0.1 | 0.2 | | Filter = 0 |
| | | | 0.05 | 0.1 | | Filter = 2 |
| Noise for limited performance in section 8.3 in low power mode (-30x, -33x code) ⁽³⁶⁾ | | | 0.2 | 0.4 | Deg. | Filter = 0 |
| | | | 0.14 | 0.28 | | Filter = 1 |
| | | | 0.1 | 0.2 | | Filter = 2 |
| Noise for nominal performance in section 8.3 in high speed mode (-31x code) ⁽³⁶⁾ | | | | 0.28 | Deg. | Filter = 0 |
| Noise for limited performance in section 8.3 in high speed mode (-31x code) ⁽³⁶⁾ | | | | 0.56 | Deg. | Filter = 0 |
| XY - Total Drift | $\partial\theta_{TT_XY}$ | | | 0.45 | Deg. | for nominal performance, see section 8.3 |
| | | | | 0.6 | | for limited performance, see section 8.3 |
| XZ - Total Drift | $\partial\theta_{TT_XZ}$ | | | 0.6 | Deg. | for nominal performance, see section 8.3 |
| | | | | 0.8 | | for limited performance, see section 8.3 |
| YZ - Total Drift | $\partial\theta_{TT_YZ}$ | | | 0.6 | Deg. | for nominal performance, see section 8.3 |
| | | | | 0.8 | | for limited performance, see section 8.3 |
| Hysteresis | | | 0.05 | 0.1 | Deg. | for nominal performance, see section 8.3 |
| | | | 0.1 | 0.2 | | for limited performance, see section 8.3 |

Table 25 - Standard Mode Nominal Magnetic Performances

9.5. Rotary Stray-field Immune Mode – High Field Variant (-5xx Code)

Valid before EoL calibration and for all applications under the conditions described in section 8.4.

| Parameter | Symbol | Min | Typ | Max | Unit | Condition |
|---|---------------------------|-----|-----|------|------|--|
| XY - Intrinsic Linearity Error | L_{E_XY} | -1 | | 1 | Deg. | |
| Noise in high speed mode (-51x code) ⁽³⁶⁾ | | | | 0.35 | | Filter = 2 |
| | | | | 0.5 | Deg. | Filter = 1 |
| | | | | 0.7 | | Filter = 0 |
| Noise in low power mode (-53x code) ⁽³⁶⁾ | | | | 0.5 | Deg. | Filter = 0 |
| XY - Total Drift | $\partial\theta_{TT_XY}$ | | | 0.67 | Deg. | |
| Hysteresis | | | 0.1 | 0.15 | Deg. | |
| Output Stray Field Immunity ⁽³⁷⁾ | $\partial\theta_{FF}$ | | | 0.3 | Deg. | In accordance of ISO 11452-8, at 30 Deg.C, with 21mT/mm magnetic In-plane gradient of In-plane field component, and stray-field strength of 4000A/m from any direction |

Table 26 - Rotary stray-field immune magnetic performance – High Field Variant

10. Memory Specifications

| Parameter | Symbol | Min | Typ | Max | Unit | Note |
|-----------|-----------|-----|------|-----|------|--|
| ROM | ROMsize | | 32 | | KB | 1 bit parity check (single error detection) |
| RAM | RAMsize | | 1024 | | B | 1 bit parity check (single error detection) |
| NVRAM | NVRAMsize | | 256 | | B | 6 bits ECC (single error correction, double error detection) |

Table 27 - Memory Specifications

11. Output Accuracy and Protocol

11.1. Analog Output

| Parameter | Symbol | Min. | Typ. | Max. | Unit | Condition |
|-----------------------------|------------------|------|------|------------|-------------------|---|
| Thermal analog output Drift | | | | 0.2 0.3 | %VDD | up to 125 °C up to 160 °C |
| Analog Output Resolution | R _{DAC} | | 12 | | bit | 12bit DAC (Theoretical) |
| | | -4 | | +4 | LSB ₁₂ | INL (before EoL calibration), output clamped between 3-97%VDD |
| | | -1.5 | | 1.5 | LSB ₁₂ | DNL |
| Ratiometric Error | | -0.1 | | 0.1 | %VDD | |

Table 28 – Analog output accuracy

11.2. PWM (Pulse Width Modulation) Output

11.2.1. Definition

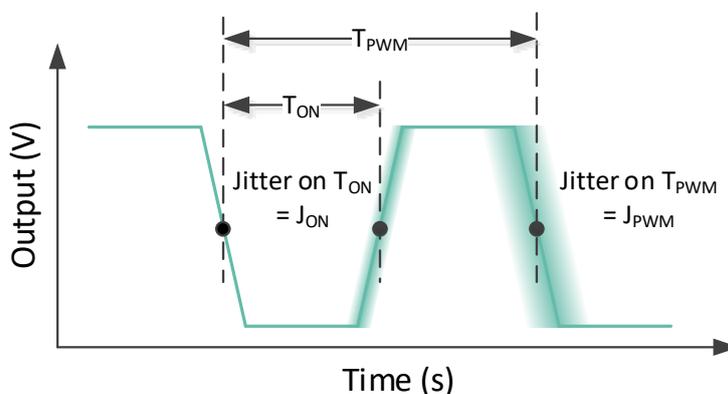


Figure 13 - PWM Signal definition

| Parameter | Symbol | Test Conditions |
|----------------------|---------------------------------------|--|
| PWM period | T _{PWM} | Trigger level = 50% VDD |
| Rise time, Fall time | t _{rise} , t _{fall} | Between 10% and 90% of VDD |
| Jitter | J _{ON} , J _{PWM} | ±3σ for 1000 successive acquisitions with clamped output |
| Duty Cycle | DC | T _{ON} / T _{PWM} |

Table 29 - PWM Signal definition

11.2.2. PWM performances

| Parameter | Symbol | Min | Typ | Max | Unit | Condition |
|-----------------------|------------------|-----|------|------|------|--|
| PWM Output Resolution | R _{pwm} | | 12 | | bit | |
| PWM %DC Jitter | J _{DC} | | | 0.03 | %DC | Push-Pull, 2kHz, C _L =4.7nF, R _{LPU} =4.7kΩ |
| PWM Period Jitter | J _{pwm} | - | - | 300 | ns | Push-Pull, 2kHz, C _L =4.7nF, R _{LPU} =4.7kΩ |
| PWM %DC thermal drift | | | 0.02 | 0.05 | %DC | Push-Pull, 2kHz, C _L =4.7nF, R _{LPU} =4.7kΩ |

Table 30 - PWM Signal Specifications

11.3. SENT (Single Edge Nibble Transmission) Output

The MLX90377 provides a digital output signal compliant with SAE J2716 Revised APR2016.

11.3.1. Sensor message definition

The MLX90377 repeatedly transmits a sequence of pulses, corresponding to a sequence of nibbles (4 bits), with the following sequence:

- Calibration/Synchronization pulse period 56 clock ticks to determine the time base of the SENT frame
- One 4-bit Status and Serial Communication nibble pulse
- A sequence of one up to six 4-bit data nibbles pulses representing the values of the signal(s) to be transmitted. The number of nibbles will be fixed for each application of the encoding scheme (i.e. Single Secure sensor format A.3, Throttle position sensor A.1)
- One 4-bit Checksum nibble pulse
- One optional pause pulse

See also SAE J2716 APR2016 for general SENT specification.

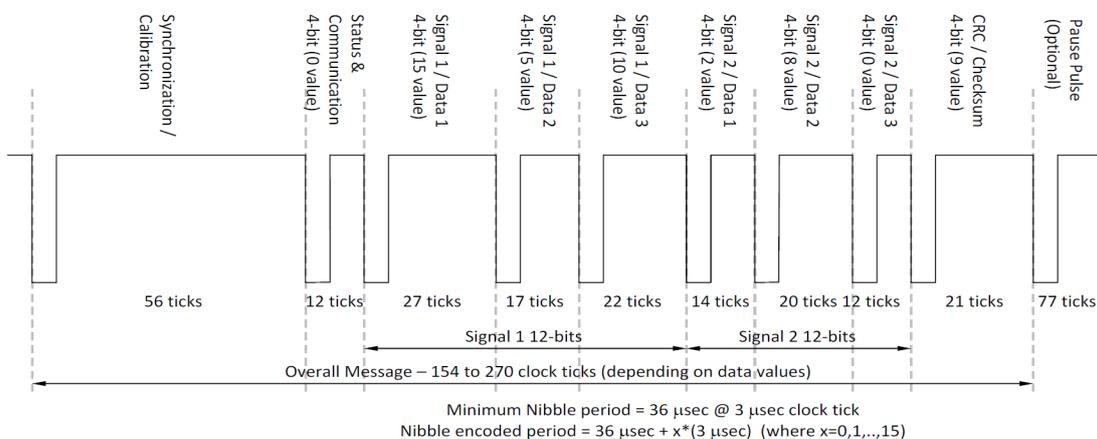


Figure 14 - SENT message encoding example for two 12bits signals

11.3.2. Sensor message frame contents

The SENT output transmits a sequence of data nibbles, according to the following configurations:

| Description | Symbol | Min | Typ | Max | Unit | Description |
|---------------------------------|-----------|-----|------------|-----|-------|---|
| Clock tick time | Tick Time | | 3 | 6 | μs | Main use cases: Normal SENT, 3μs tick time Slow SENT, 6μs tick time |
| Number of data nibbles | Xdn | 3 | 6 | | | |
| Frame duration (no pause pulse) | Npp | 154 | | 270 | ticks | 6 data nibbles |
| Frame duration with pause pulse | Ppc | 282 | 320 | 922 | ticks | |
| Sensor type | | | A.1 A.3 | | | Dual Throttle Position sensors Single Secure sensors |

Table 31 - SENT Protocol Frame Definition

11.3.3. SENT Format Option

The default SENT format option of MLX90377 is Single Secure Sensor A.3 (H.4). The MLX90377 SENT transmits a sequence of data nibbles; according single secure sensor format defined in SAE J2716 appendix H.4 ⁽³⁹⁾. The frame contains 12-bit angular value, an 8 bit rolling counter and an inverted copy of the most significant nibble of angular value.

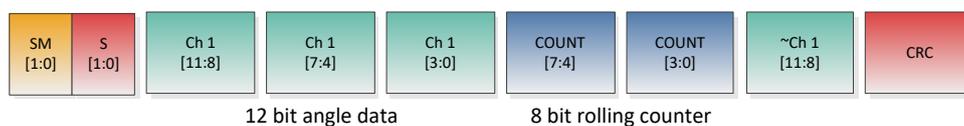


Figure 15 - H.4 Single Secure Sensor Frame Format

| Shorthand Description | Tick time | Data nibbles | Pause Pulse | Serial message | Data format |
|--|-----------|--------------|-------------|----------------|-------------|
| SENT2016-03.0us-6dn-ppc(366.0)-esp-A.3 | 3μs | 6 | Y | Enhanced | H.4 |
| SENT2016-03.0us-6dn-ppc(366.0)-nsp-A.3 | 3μs | 6 | Y | None | H.4 |
| SENT2016-03.0us-6dn-npp-nsp-A.3 | 3μs | 6 | N | None | H.4 |
| SENT2016-##-#us-#dn-###()-###-A.3 | 1..12 | 6 | Y/N | En/None | H.4 |

Table 32 - H.4 Single Secure Sensor Shorthand examples

³⁹ As referenced from section A.3 Single Secure Sensor of the standard.

11.3.4. Start-up behaviour

The circuit will send initialisation frames once digital start-up is done but angle measurement initialisation sequence is not yet complete. These initialisation frames content can be chosen by user with the following option:

| SENT_INIT_GM | Initialisation frame value | Comments |
|--------------|----------------------------|-----------------|
| 0 | 0x000 | SAE compliant |
| 1 | 0xFF | OEM requirement |

Table 33 - Initialization Frame Content Definition

11.3.5. Output configuration

In SENT mode, the MLX90377 can be configured in open drain mode, normal push-pull mode, as well as an enhanced emission mode, which is the default configuration, see also section 14.1.1.

The tick time is the unit reference for SENT. The default tick time value is 3µs.

The output resistive load, e.g. the external pull-up or pull-down resistor should be carefully selected, because the MLX90377 has a built-in high order low pass filter. A large resistive load will deteriorate the generated SENT signal, and could make the output signal not comply to the SENT specifications, such as the fall times and the minimum output voltages, e.g. parameters $V_{satD_lopp}/V_{satD_hipp}$ in Table 10 – Electrical Specifications in chapter 6. In principle, the values in Table 10 – Electrical Specifications in chapter 6 should be considered, which means it is not recommended to have a resistive load value smaller than 10kΩ, and a resistive load value smaller than 3kΩ should be avoided. The maximum output resistive load value should be less than 55kΩ to avoid unexpected impact from leakage current.

Furthermore, the output capacitance should also be properly chosen, together with the output resistive load to correspondingly match the application, e.g. tick time, to allow appropriate time constant for the transmission of the SENT signal. Details can be seen in sections 16.1 and 16.2.

11.3.6. SENT Output Timing configuration

Different SENT output timing configurations are listed in the following tables:

| SENT_TICK_TIME | Tick time configuration | Description |
|----------------|-------------------------|-----------------|
| 0 | 6 µs | |
| 1 | 1 µs | Not recommended |
| 2 | 2 µs | Not recommended |
| 3 | 3 µs | Standard SENT |
| 4 | 4 µs | Not recommended |
| 5 | 5 µs | Not recommended |

| SENT_TICK_TIME | Tick time configuration | Description |
|----------------|-------------------------|-------------|
| 6 | 12 μ s | Do not use |
| 7 | 24 μ s | Do not use |

Table 34 - SENT Tick Time Configuration (-33x code)

| NIBBLE_PULSE_CONFIG | High/low time configuration |
|---------------------|-----------------------------|
| 2 | Fixed low time (5 ticks) |
| 3 | Fixed high time (6 ticks) |

Table 35 - SENT Nibble configuration (high/low times)

11.3.7. Serial message channel (slow channel)

Serial data is transmitted sequentially in bit number 3 and 2 of the status and communication nibble. A serial message frame stretches over 18 consecutive SENT data messages from the transmitter. All 18 frames must be successfully received (no errors, calibration pulse variation, data nibble CRC error, etc.) for the serial value to be received.

11.3.7.1. Enhanced Serial Message (ESM)

Enhanced serial message with 12-bit data and 8-bit message ID is used (SAE J2716 APR2016 5.2.4.2, Figure 5.2.4.2-2). According to the standard, SM[0] contains a 6bits CRC followed by a 12-bits data. Message content is defined by a 8-bit message ID transmitted in the SM[1] channel. Correspondence between ID and message content is defined in the tables below (Table 36, Table 37 and Table 38).

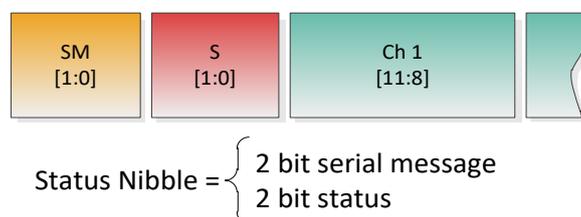


Figure 16 - SENT Status Nibble and Serial Message

By default, the short sequence consisting of a cycle of 24 data is transmitted (Table 36). An extended sequence can be used through configuration of SENT_SLOW_EXTENDED (Table 37). Additionally, the norm of the B field detected by the sensor can be returned at the end of the sequence by setting SENT_SLOW_BFIELD (Table 38)

| # | 8-bit ID | Item | Source data |
|---|----------|---------------------------|------------------------------|
| 1 | 0x01 | Diagnostic error code | Current status code from RAM |
| 2 | 0x06 | SENT standard revision | SENT_REV from NVRAM |
| 3 | 0x01 | Diagnostic error code | Current status code from RAM |
| 4 | 0x05 | Manufacturer code | SENT_MAN_CODE from NVRAM |
| 5 | 0x01 | Diagnostic error code | Current status code from RAM |
| 6 | 0x03 | Channel 1 / 2 Sensor type | SENT_SENSOR_TYPE from NVRAM |
| 7 | 0x01 | Diagnostic error code | Current status code from RAM |
| 8 | 0x07 | Fast channel 1: X1 | SENT_CHANNEL_X1 from NVRAM |
| 9 | 0x01 | Diagnostic error code | Current status code from RAM |

| # | 8-bit ID | Item | Source data |
|----|----------|------------------------|------------------------------|
| 10 | 0x08 | Fast channel 1: X2 | SENT_CHANNEL_X2 from NVRAM |
| 11 | 0x01 | Diagnostic error code | Current status code from RAM |
| 12 | 0x09 | Fast channel 1: Y1 | SENT_CHANNEL_Y1 from NVRAM |
| 13 | 0x01 | Diagnostic error code | Current status code from RAM |
| 14 | 0x0A | Fast channel 1: Y2 | SENT_CHANNEL_Y2 from NVRAM |
| 15 | 0x01 | Diagnostic error code | Current status code from RAM |
| 16 | 0x23 | (Internal) temperature | Current temperature from RAM |
| 17 | 0x01 | Diagnostic error code | Current status code from RAM |
| 18 | 0x29 | Sensor ID #1 | SENT_SENSOR_ID1 from NVRAM |
| 19 | 0x01 | Diagnostic error code | Current status code from RAM |
| 20 | 0x2A | Sensor ID #2 | SENT_SENSOR_ID2 from NVRAM |
| 21 | 0x01 | Diagnostic error code | Current status code from RAM |
| 22 | 0x2B | Sensor ID #3 | SENT_SENSOR_ID3 from NVRAM |
| 23 | 0x01 | Diagnostic error code | Current status code from RAM |
| 24 | 0x2C | Sensor ID #4 | SENT_SENSOR_ID4 from NVRAM |

Table 36 - SENT Enhanced Slow Channel Standard Data Sequence

| # | 8bit ID | Item | Source data |
|----|---------|-----------------------|------------------------------|
| 25 | 0x01 | Diagnostic error code | Current status code from RAM |
| 26 | 0x90 | OEM Code #1 | SENT_OEM_CODE1 from NVRAM |
| 27 | 0x01 | Diagnostic error code | Current status code from RAM |
| 28 | 0x91 | OEM Code #2 | SENT_OEM_CODE2 from NVRAM |
| 29 | 0x01 | Diagnostic error code | Current status code from RAM |
| 30 | 0x92 | OEM Code #3 | SENT_OEM_CODE3 from NVRAM |
| 31 | 0x01 | Diagnostic error code | Current status code from RAM |
| 32 | 0x93 | OEM Code #4 | SENT_OEM_CODE4 from NVRAM |
| 33 | 0x01 | Diagnostic error code | Current status code from RAM |
| 34 | 0x94 | OEM Code #5 | SENT_OEM_CODE5 from NVRAM |
| 35 | 0x01 | Diagnostic error code | Current status code from RAM |
| 36 | 0x95 | OEM Code #6 | SENT_OEM_CODE6 from NVRAM |
| 37 | 0x01 | Diagnostic error code | Current status code from RAM |
| 38 | 0x96 | OEM Code #7 | SENT_OEM_CODE7 from NVRAM |
| 39 | 0x01 | Diagnostic error code | Current status code from RAM |
| 40 | 0x97 | OEM Code #8 | SENT_OEM_CODE8 from NVRAM |

Table 37 - SENT Slow Channel Extended Data Sequence

| # | 8bit ID | Item | source data |
|----|---------|----------------|--|
| 25 | 0x80 | Field Strength | SENT_SLOW_EXTENSION 0: Disabled. No further extension 1: NTC value 2: field-strength temperature compensated, before magnet sensitivity compensation 3: field-strength temperature and gain independent (standard sequence) |
| 41 | 0x80 | Field Strength | SENT_SLOW_EXTENSION 0: Disabled. No further extension 1: NTC value 2: field-strength temperature compensated, before magnet sensitivity compensation 3: field-strength temperature and gain independent (extended sequence) |

Table 38 - SENT Slow Channel Magnetic Field Norm ID and position

For Field Strength encoding, see section 14.5.4.

11.3.7.2. Short Serial Message (SSM)

Short serial message with 8-bit data and 4-bit message ID is used (SAE J2716 APR2016 5.2.4.1, Figure 5.2.4.1-1). According to the standard, it contains a 8-bit data followed by a 4-bit CRC.

| # | 4-bit ID | Item | Source data |
|----|----------|---------------------------|------------------------------|
| 1 | 0x01 | Diagnostic error code | Current status code from RAM |
| 2 | 0x06 | SENT standard revision | SENT_REV from NVRAM |
| 3 | 0x01 | Diagnostic error code | Current status code from RAM |
| 4 | 0x05 | Manufacturer code | SENT_MAN_CODE from NVRAM |
| 5 | 0x01 | Diagnostic error code | Current status code from RAM |
| 6 | 0x03 | Channel 1 / 2 Sensor type | SENT_SENSOR_TYPE from NVRAM |
| 7 | 0x01 | Diagnostic error code | Current status code from RAM |
| 8 | 0x07 | Fast channel 1: X1 | SENT_CHANNEL_X1 from NVRAM |
| 9 | 0x01 | Diagnostic error code | Current status code from RAM |
| 10 | 0x08 | Fast channel 1: X2 | SENT_CHANNEL_X2 from NVRAM |
| 11 | 0x01 | Diagnostic error code | Current status code from RAM |
| 12 | 0x09 | Fast channel 1: Y1 | SENT_CHANNEL_Y1 from NVRAM |
| 13 | 0x01 | Diagnostic error code | Current status code from RAM |
| 14 | 0x0A | Fast channel 1: Y2 | SENT_CHANNEL_Y2 from NVRAM |
| 15 | 0x01 | Diagnostic error code | Current status code from RAM |
| 16 | 0x02 | (Internal) temperature | Current temperature from RAM |
| 17 | 0x01 | Diagnostic error code | Current status code from RAM |
| 18 | 0x0B | Sensor ID #1 | SENT_SENSOR_ID1 from NVRAM |

| # | 4-bit ID | Item | Source data |
|----|----------|-----------------------|----------------------------------|
| 19 | 0x01 | Diagnostic error code | Current status code from RAM |
| 20 | 0x0C | Sensor ID #2 | SENT_SENSOR_ID2 from NVRAM |
| 21 | 0x01 | Diagnostic error code | Current status code from RAM |
| 22 | 0x0D | Sensor ID #3 | SENT_SENSOR_ID3 from NVRAM |
| 23 | 0x01 | Diagnostic error code | Current status code from RAM |
| 24 | 0x0E | Sensor ID #4 | SENT_SENSOR_ID4 from NVRAM |
| 25 | 0x01 | Diagnostic error code | Current status code from RAM |
| 26 | 0x04 | Field Strength | Field Strength (mapped to 8 bit) |

Table 39 - SENT Short Serial Slow Channel Standard Data Sequence

11.3.8. Serial Message Error Code

11.3.8.1. Enhanced Serial Message (ESM)

The list of error and status messages transmitted in the 12-bit Serial Message data field when Serial Message 8-bit ID is 0x01, is given in the Table 40. The error is one-hot encoded and therefore each bit is linked to one or several diagnostics. Only the first error detected during a diagnostics cycle is reported, and serial message error code will be updated at every diagnostics cycle. The serial message error code will only be cleared out once all the errors disappear. This mechanism ensures only one error at a time takes control of the error debouncing counter (see 14.5.2).

The MSB acts as an error Flag when SENT_DIAG_STRICT is set. This bit will be high only when an error is present. This bit can be kept high even if no error is present (SENT_DIAG_STRICT = 0).

| Bit Nb | 12 Bit Data (hex) | Diagnostic | Comments |
|--------|-------------------|---|---|
| - | 0x000 / 0x800 | No error | Programmable (see SENT_DIAG_STRICT in Table 49) |
| 0 | 0x801 | GAINOOS | Magnetic Signal Conditioning Gain Clamping |
| 1 | 0x802 | FieldTooLow | Fieldstrength is below defined low threshold (see 14.5.4) |
| 2 | 0x804 | FieldTooHigh | Fieldstrength is above defined high threshold (see 14.5.4) |
| 3 | 0x808 | ADCclip | ADC is saturated, either low or high |
| 4 | 0x810 | ADC_test | ADC made wrong conversion |
| 5 | 0x820 | Analog Supply Monitors | Detects VDDA (VDEC) over and under voltage or VDD under voltage |
| 6 | 0x840 | Digital Supply Monitors | Detects VDDD (1.8V internal digital supply) overvoltage |
| 7 | 0x880 | RoughOffset | Hall Element offset monitor |
| 8 | 0x900 | Over/Under Temp / ADCdrop | Temperature sensor monitor (see 14.5.3) / ADC conversion interruption |
| 9 | 0xA00 | DSP overflow / HE_Bias / Analog Front End | DSP overflow / Hall Element biasing issue / Analog front end self-test |
| 10 | 0xC00 | Supply Bias Current / Stress Range | Current biasing system monitor / Stress Sensor Readout |
| 11 | 0x800 | Extra Error Flag | set to one if any error present (only when SENT_DIAG_STRICT = 1). Otherwise, always high. |

Table 40 - SENT Serial Message Error Code for Enhanced Serial Message

11.3.8.2. Short Serial Message (SSM)

The list of error and status messages transmitted in the 8-bit Serial Message data field when Serial Message 8-bit ID is 0x01, is given in the Table 41. The error is one-hot encoded and therefore each bit is linked to one or several diagnostics. Only the first error detected is reported and serial message error code will not be updated until all the errors have disappeared. This mechanism ensures only one error at a time takes control of the error debouncing counter (see 14.5.2).

| Bit Nb | 8 Bit Data (hex) | Diagnostic | Comments |
|--------|------------------|---|--|
| - | 0x00 | No error | |
| 0 | 0x01 | GAINOOS | Magnetic Signal Conditioning Gain Clamping |
| 1 | 0x02 | FieldTooLow | Fieldstrength is below defined low threshold (see 14.5.4) |
| 2 | 0x04 | FieldTooHigh | Fieldstrength is above defined high threshold (see 14.5.4) |
| 3 | 0x08 | ADCclip / ADC_test | ADC is saturated, either low or high / ADC made wrong conversion |
| 4 | 0x10 | Analog Supply Monitors / Digital Supply Monitors / Supply Bias Current / Stress Range | Detects VDDA (VDEC) over and under voltage or VDD under voltage / Detects VDDD (1.8V internal digital supply) overvoltage / Current biasing system monitor / Stress Sensor Readout |
| 5 | 0x20 | RoughOffset | Hall Element offset monitor |
| 6 | 0x40 | Over/Under Temp / ADCdrop | Temperature sensor monitor (see 14.5.3) / ADC conversion interruption |

Table 41 - SENT Serial Message Error Code for Short Serial Message

11.3.9. SENT configuration shorthand definition

| Shorthand description | Format | Req | 90377 programmable setting |
|------------------------|----------------------------------|--|--|
| SENT SAE J2716 Rev | SENT xxxx | 2007 2008 2010 2016 | CRC_2007 0 > 2007 1 2007 |
| Clock Tick length [μs] | XX.X μs | 1<xx<12 | SENT_TICK_TIME 0 SENT 6μs 1 SENT 1μs 2 SENT 2μs 3 SENT 3μs 4 SENT 4μs 5 SENT 5μs |
| SENT format option | H.x | SENT format | SENT_FC_FORMAT 0: format H.1 1: format H.1 2: format H.2 3: format H.3 4: format H.4 5: format H.5 6: format H.6 7: format H.7 |
| Pause Pulse Option | npp ppc (xxx.0) xxx | No pause Pulse Pause Pulse with const. frame length Frame Length (in clock ticks) | PROTOCOL 0 = npp 2 = ppc T_FRAME xxx > 282...922 |
| Use of Serial protocol | nsp ssp esp | No serial protocol Short serial protocol Enhanced serial protocol | SERIAL_CONFIG 1 nsp 2 ssp 3 esp |

Table 42 - SENT Shorthand Description

11.4. SPC (Short PWM Code) Output

11.4.1. General Definition

The MLX90377 can output the signal in SPC (Short PWM Code) frame selection according to the SPC2015 rev1.0. It is an extension of the SAE J2716 SENT protocol. It allows on-demand transmission of one data frame where the master triggers the transfer of data. The transmission of the SPC frame is always triggered by the master in every mode (synchronous and with ID selection).

The SPC output of the MLX90377 transmits a sequence of data nibbles, according to the following configurations:

| Description | Symbol | Min | Typ | Max | Unit | Description |
|---------------------------------|----------|-----|---------------------|-----|-------|---|
| Clock tick time | tickTime | 0.5 | 1.5 ⁽⁴⁰⁾ | 12 | µs | Main use cases : SPC, 1.5µs tick time (default) Normal SENT, 3µs tick time Slow SENT, 6µs tick time (see section 7.6) |
| Number of data nibbles | Xdn | 3 | | 6 | | |
| Frame duration (no pause pulse) | Npp | 154 | | 270 | ticks | 6 data nibbles |
| Frame duration with pause pulse | Ppc | 282 | 320 | 922 | ticks | |
| Sensor type | A.3 | | | | | Single Secure sensors |

Table 43 - SENT Protocol Frame Definition

11.4.2. Tick Time requirements

The basic SPC protocol unit time is defined as 3 µs. The protocol standard requires a tick time variation of no more than 20% for legacy applications and 10% for general applications. The lowest tick time that can be programmed in the MLX90377 is 0.5 µs. Improved Emission Mode will only be possible for tick time equal or greater to 1.5 µs.

11.4.3. Output Configuration

In SPC mode, the MLX90377 can be configured in open drain mode, normal push-pull mode, as well as an enhanced emission mode, see also section 14.1.1.

The default output driver configuration is to send the SENT frame in push-pull mode.

Users can select three options (see parameter SPC_PP_Option, in Table 49 in chapter 12):

1. IC is always in open drain
2. IC sends the SENT frame from a pre-defined timer (SPC_OUT_ON_TH) to CRC in push-pull. It then returns to high-Z
3. IC sends the SENT frame from a pre-defined timer (SPC_OUT_ON_TH) until end of a pre-defined timer (SPC_OUT_OFF_TH). It then returns to high-Z.

⁴⁰ Please contact our Direct Sales Team if other options than main use cases is required.

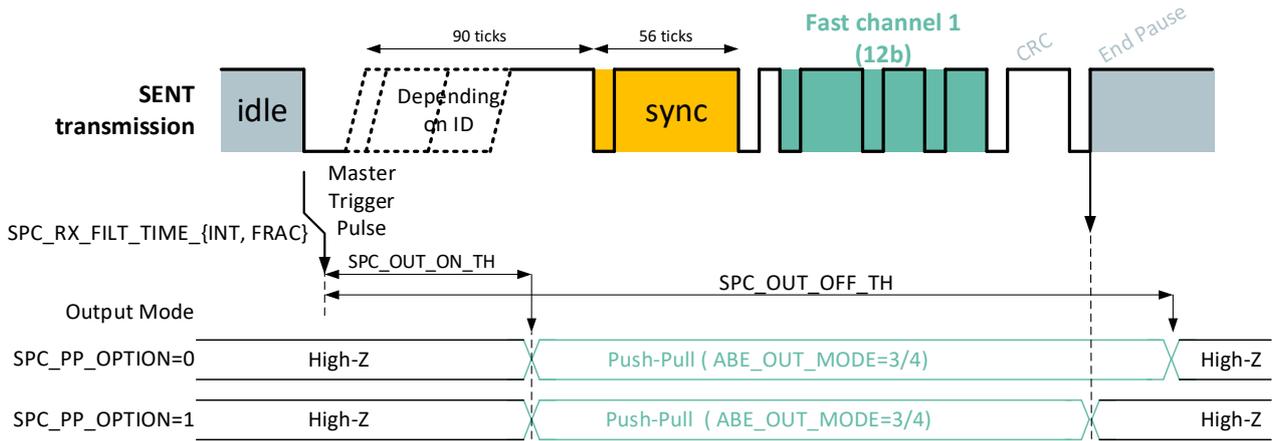


Figure 17 – SPC output driver state switching

| Parameter | Symbol | Min | Typ | Max | Unit | Condition |
|--------------------|------------------------|-----|-----|-----|---------------|--|
| Push-Pull Duration | $T_{\text{push-pull}}$ | 280 | | 550 | μs | Configurable Duration how long IC stays in push-pull once sync pulse of SENT frame starts. |

Table 44 – Push-pull duration in SPC mode

The sensor output driver can also be configured as open-drain to enable the master to drive the line. Initially the line is at high level, the chip is waiting for the master to pull it low. This state is called ‘Idle state’, meaning the sensor is not transmitting any data. At a given time, the master will send a trigger low pulse along the line. The embedded CPU of the sensor will measure this master pulse length. If it is recognized as a valid trigger, the chip will answer by sending back an SPC frame containing the current data. Several configurations of the SPC frame are available. If the trigger pulse is not valid, the chip will not transmit any data, keeping the line free. In order to set up the permanent open-drain mode (NV_ABE_OUT_MODE=1), it is important to configure the parameters NV_SPC_OUT_ON_TH and NV_SPC_OUT_OFF_TH thresholds according to the FRAME settings.

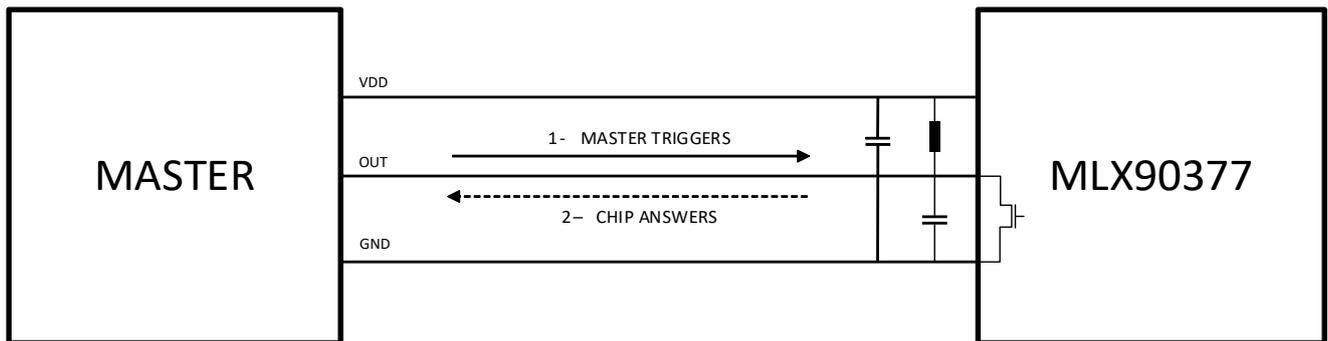


Figure 18 – SPC standard master-slave configuration

Like in SENT, the tick time is the unit reference for SPC. It can be as low as 0.5 μs enabling fast transmission rate and short frames. However, the default tick time value is 1.5 μs in SPC mode, to benefit from the fast acquisition and magnetic processing of the device.

For the MLX90377, the output resistance, e.g. the external pull-up or pull-down resistor should be carefully selected, because the MLX90377 has a built-in high order low pass filter, too heavy resistive load will

deteriorate the generated SPC signal, and could make the output signal not comply to the SPC specifications, such as the fall times and the minimum output voltages, i.e. parameters $V_{\text{satD_lopp}}/V_{\text{satD_hipp}}$ and $V_{\text{satD_loie}}/V_{\text{satD_hiie}}$ in Table 10 – Electrical Specifications in chapter 6. In principle, the values in Table 10 – Electrical Specifications in chapter 6 should be considered, which means it is not recommended to have a resistive load value smaller than 10k Ω , and a resistive load value smaller than 3k Ω should be avoided. The maximum output resistive load value should be less than 55k Ω to avoid unexpected impact from leakage current.

Furthermore, the output capacitance should also be properly chosen together with the output resistive load to correspondingly match the application, e.g. tick time, SPC ID, etc. to allow appropriate time window for the trigger pulse reception. The worst case scenario occurs for SPC ID0, where the trigger pulse should be recognized within only 4 ticks (refer to Table 48 and Figure 23 in section 11.4), resulting in 6 μs time window for 1.5 μs tick time SPC protocol. Therefore, RC time constant should be less than 6 μs in such application. When the time window for the trigger pulse reception changes, the resistive and capacitive load on output can be scaled correspondingly. It is also important to note that the mentioned capacitive load refers to the total load on the bus, meaning the value should be equally divided if there are multiple sensors connected, also the load on the ECU side should be taken into account and scaled properly.

11.4.4. SPC Output Timing configuration

Different SPC output timing configurations are listed in the following table:

| SENT_TICK_TIME | Tick time configuration | Description |
|----------------|-------------------------|-------------|
| 0 | 3 μs | |
| 1 | 0.5 μs | |
| 2 | 1 μs | |
| 3 | 1.5 μs | Default SPC |
| 4 | 2 μs | |
| 5 | 2.5 μs | |
| 6 | 6 μs | |
| 7 | 12 μs | |

Table 45 - SPC Tick Time Configuration (-x8x code)

11.4.5. Frame Content

11.4.5.1. Global Definition

A message frame consists of the following sequence:

1. Trigger pulse
2. Synchronization pulse
3. Status and communication nibble
4. Data nibbles

5. Checksum nibble
6. End Lone pulse

The overall scheme of the SPC frame transmission is described in the following figure.

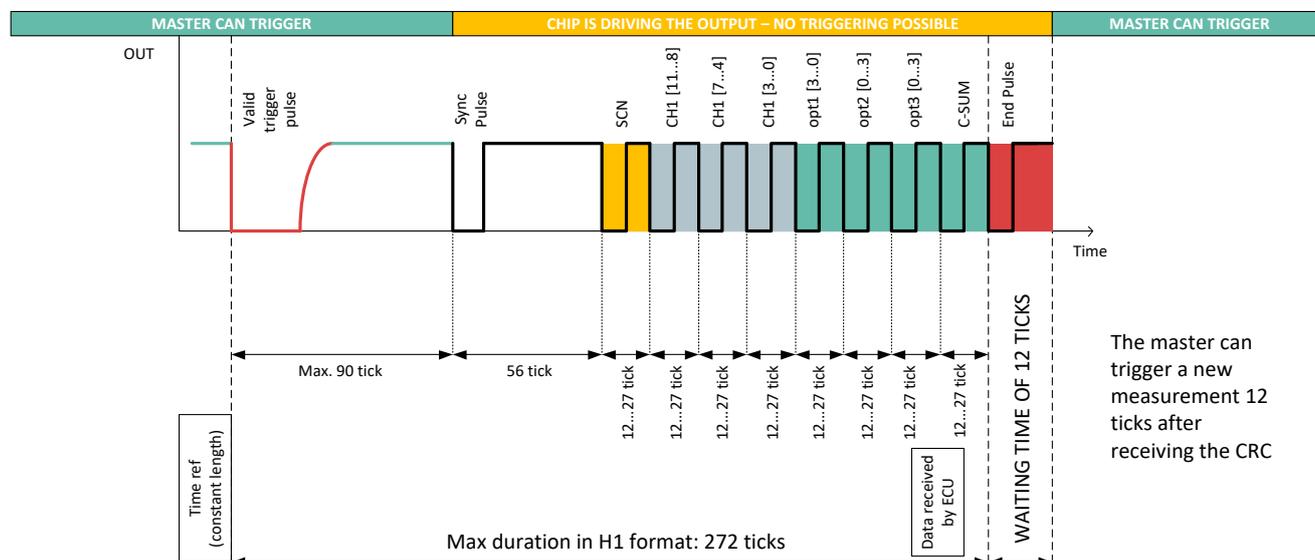


Figure 19 - SPC Frame description

11.4.5.2. Trigger Pulse

The trigger pulse is considered to be part of the SPC frame. The timing related to this trigger pulse is critical as it is initializing the data transfer. The sensor is monitoring the line and computes the duration the line is kept low by the master. This time is called “master low time”.

| Parameter | Symbol | Min | Typ | Max | Unit |
|------------------------|--------------|------|------|-----|------|
| Falling edge threshold | V_{thfSPC} | 1.1 | 1.32 | 1.7 | V |
| Rising Edge Threshold | V_{thrSPC} | 1.25 | 1.52 | 1.8 | V |

Table 46 – SPC Master Trigger Pulse Specifications

The falling edge of the trigger pulse is considered to be the reference of the protocol. The master low time is quantized using clock ticks.

After a non-valid trigger detected, a blanking time of five unit times (counted from the detected rising edge of the trigger pulse) is used by the sensor, before the next trigger measurement is performed. This concept improves EMC robustness and avoids unwanted answer from the chip if the line is subject to spikes.

After detection of a valid trigger and expiration of the trigger pulse time according to the selected transmission mode the sensor starts with the remaining protocol frame.

11.4.5.3. Status and communication nibble (SCN)

The status nibble contains information for error reporting and some optional information like the chip ID or slow channel data.

The position of the 2 status bits and 2 ID bits in SPC mode of MLX90377 complies with the one specified in the SPC standard, but can be reversed if the SENT-standard definition is preferred.

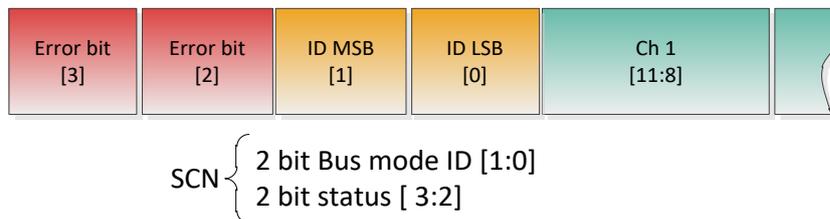


Figure 20 – SPC Bus Transmission Mode Multiplex Example (SPC_SCN_BIT_ORDER=1)

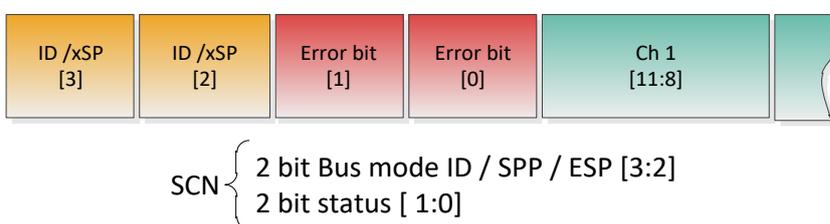


Figure 21 – SENT Transmission Mode Multiplex Example (SPC_SCN_BIT_ORDER=0)

11.4.5.4. Data Nibble

The definition of the data nibbles in SPC matches with SENT. The number of data nibbles transmitted is determined by frame format configured. The data content is programmable. By default, 12 bits of data are transmitted in 3 nibbles.

11.4.5.5. Checksum Nibble

SPC protocol allows transmission of error/warning flags in the status and communication nibble. Then it is highly recommended to take this nibble into the checksum calculation. This option is programmable in the MLX90377. Nevertheless, to be SENT compatible, it is possible to limit the checksum input data to data nibbles only.

MLX90377 also supports the SPC improved nibble checksum algorithms (for the SENT nibbles), method-E and method-O.

11.4.5.6. End Lone Pulse

The MLX90377 generates a pulse with a length of 12 clock ticks after transmission of checksum nibble. In that case the master cannot trigger the chip. During that period, MLX90377 is still holding the line. The blanking time is programmable, see parameters SPC_OUT_ON_TH, SPC_OUT_OFF_TH in chapter 12.

11.4.6. Synchronous Transmission Mode

In the SPC synchronous mode, the sensor responds to a low pulse of duration between 1.5 and 4 tick times. When using this mode, the time between the end of the trigger pulse and the start of the SENT frame shall

be set properly⁽⁴¹⁾, in order to ensure the synchronisation between the master and the chip, and the data gets acquired in the same frame, which is guaranteed with the condition of 3 μ s tick time.

| Parameter | Symbol | Min | Typ | Max | Unit |
|---------------------------------------|----------------------|-----|------|-----|-------|
| SPC Synchronous Mode Trigger Low Time | $t_{m\text{lowSPC}}$ | 1.5 | 2.75 | 4 | ticks |

Table 47 – SPC Synchronous Timings

11.4.7. Bus Transmission Mode

In the on-demand timing modes, the sensor responds to SPC trigger pulses from the master when its low pulse time corresponds to its pre-programmed ID. The selection of the ID is done with the parameter SPC_ID.

| Parameter | Symbol | SPC_CHIP_ID | Min | Typ | Max | Unit | Remark |
|--------------------------|----------------------|-------------|-----|-----|-----|-------|---|
| Bus Mode Master Low Time | $t_{m\text{lowSPC}}$ | 0 | 8 | | 15 | ticks | |
| | | 1 | 16 | | 28 | ticks | |
| | | 2 | 29 | | 49 | ticks | |
| | | 3 | 50 | | 82 | ticks | |
| Total trigger time | $t_{m\text{trSPC}}$ | - | - | 90 | - | ticks | master trigger pulse, independent of SPC_ID, not programmable |

Table 48 – Bus Transmission Mode Timings

The ECU is able to trigger only one chip on a bus. The maximum amount of chip on the bus line is 4. Each chip must have its output configured to open-drain with a different ID.

⁴¹ Please contact Melexis application service for detailed support.

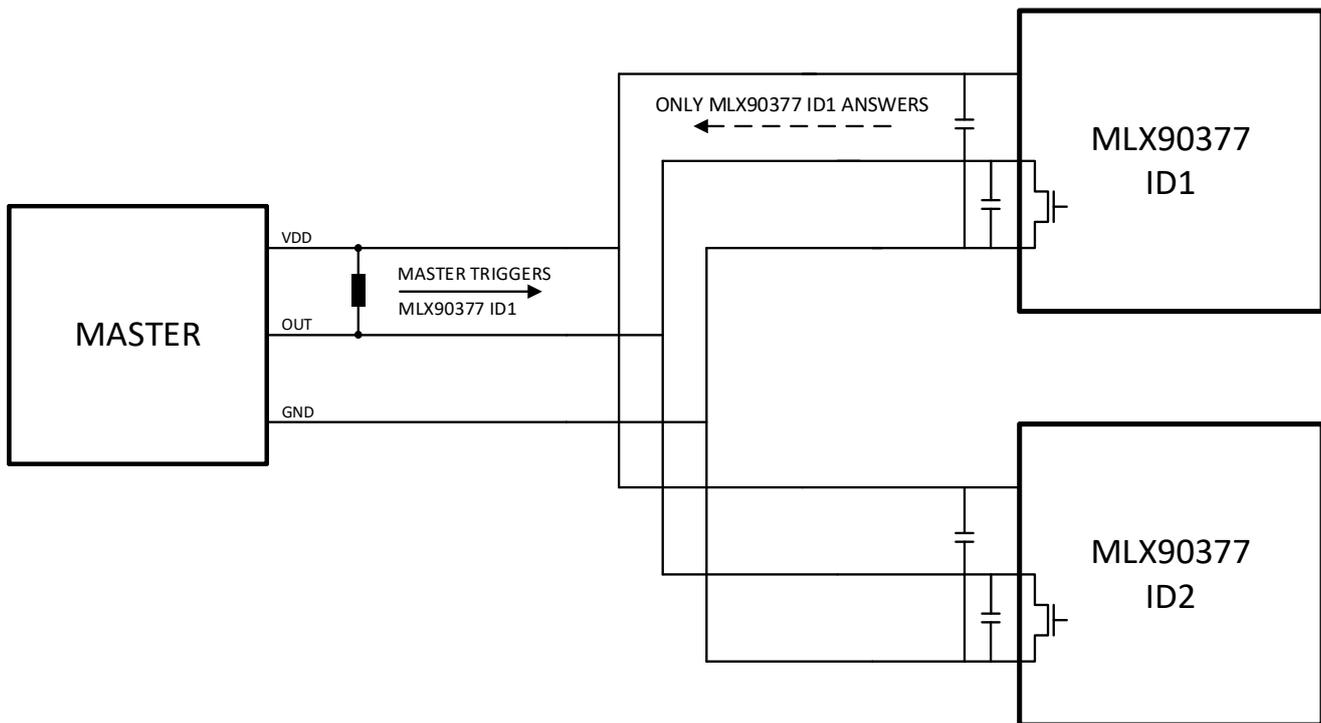


Figure 22 – SPC Bus Transmission Mode Multiplex Example

Figure 23 shows the timing diagram of the trigger pulse reception depending on the SPC ID.

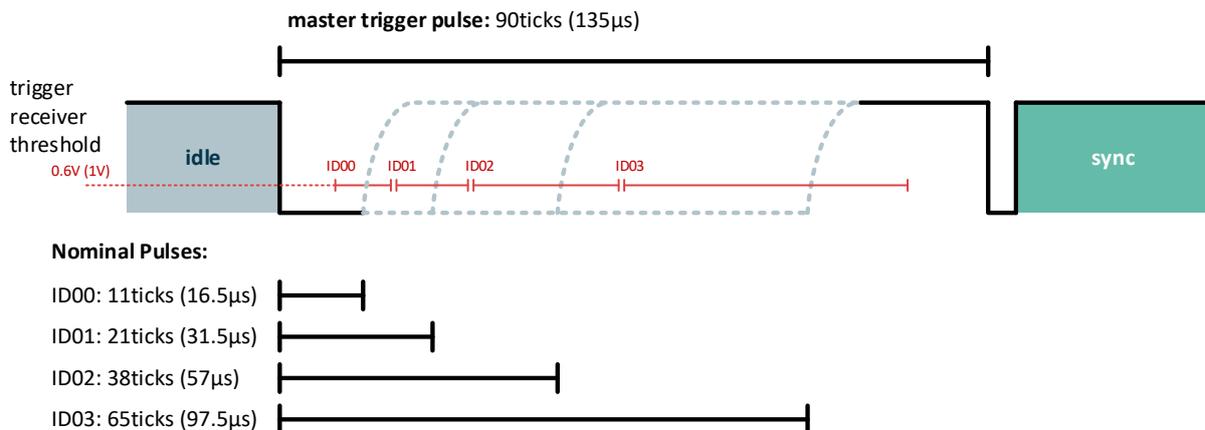


Figure 23 – Time Window for the Trigger Pulse Reception

Notes:

- ID01 and ID03 are advised when only 2 sensors are on the bus
- For correct trigger pulse reception, the clock tolerance of the sensor, the time constant of the rising edge and the receiver threshold shall be considered
- The ECU may adjust the low trigger time to account for the time constant of the rising edge
- Example timings are calculated for the default 1.5µs tick time

Bus Transmission Mode allows two configurations for acquiring data and sending data:

1. The IC for which SPC ID matches acquires its data and sends in the same frame the data.
2. All IC's on the bus acquire the data when a specific SPC ID is sent. Each IC buffers and transmits when called upon, see Figure 24.

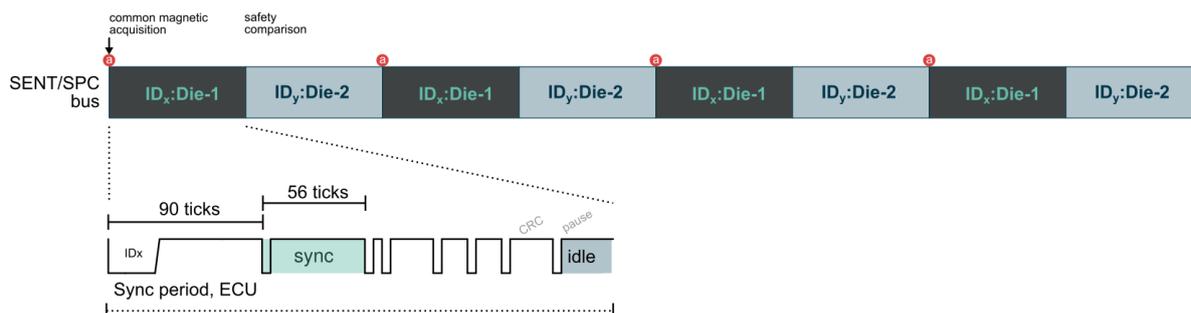


Figure 24 – Acquire Data during one SPC ID on all ICs on the same bus. Buffer and send when SPC ID matches.

11.4.8. Programming in Bus Mode

The EoL programming can be executed even if 4 chips are connected to the same bus. The protocol is capable of selective communication with one chip connected in bus configuration together with other chips (up to 4 on the same line). It uses the MUPET_ADDRESS stored in NVRAM to select the chip to be programmed. When programming 4 chips in parallel, external serial resistors shall be avoided, and are in fact not needed, as the improved emission mode is addressing the additional filtering via a resistance implemented on chip. It enables EoL programming even if 4 chips are connected to the same bus. The SPC ID must be programmed up front. Discuss with your local sales representative for this option.

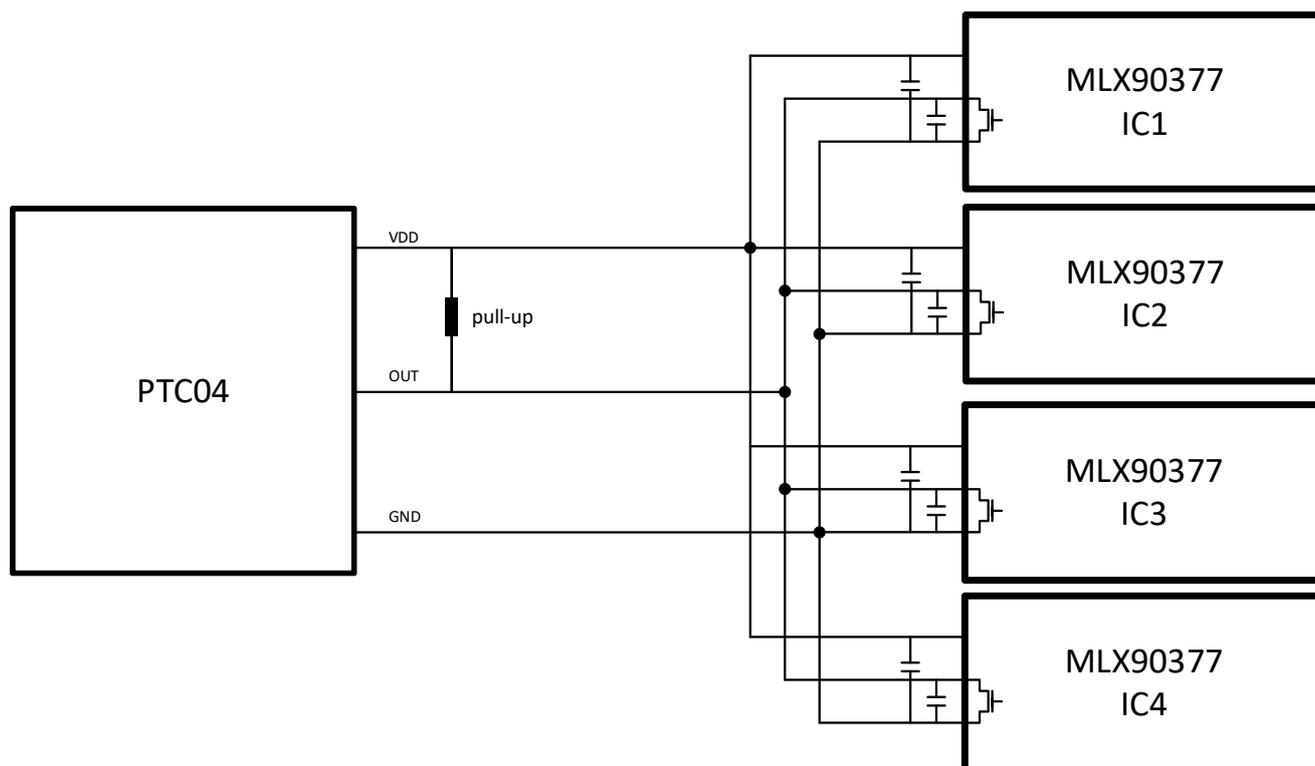


Figure 25 – Example of multiple units programming

12. End User Programmable Items

| Parameter | Description | Default Values | | | |
|--|--|----------------------------|--------|-------------|--------|
| | | SENT | SPC | Analog /PWM | #bits |
| GENERAL CONFIGURATION | | | | | |
| USER_ID[0..5] | User Id. Reference, details see Table 50 in chapter 13 | see Table 50 in chapter 13 | | | 8 |
| WARM_TRIGGER_LONG | Add delay for PTC Entry Level | 0 | 0 | 0 | 1 |
| WARM_ACT_HIGHV | Select PTC Entry Level Default 0 = V_{PROV0} | 0 | 0 | 0 | 1 |
| ROUT_LOW | Select output impedance for PTC communication | 1 | 1 | 1 | 1 |
| MUPET_ADDRESS_MODE | Configuration for PTC address. Do not modify! | 0 | 0 | 0 | 1 |
| MUPET_ADDRESS | PTC address for which the slave will communicate | 0/1 | 0/1 | 0/1 | 2 |
| MEMLOCK | Enable NVRAM write protection | 0 | 0 | 0 | 2 |
| SENSOR FRONT END | | | | | |
| GAINMIN | Minimum virtual gain code [0..63] | 0 | 0 | 0 | 8 |
| GAINMAX | Maximum virtual gain code [0..63] | 63 | 63 | 64 | |
| SENSING_MODE | Rotary position stray-field Immune | 0 | 0 | 0 | 3 |
| | X-Y Angular Rotary (standard / legacy) | 1 | 1 | 1 | |
| | Y-Z Angular Rotary (standard / legacy) | 2 | 2 | 2 | |
| | X-Z Angular Rotary (standard / legacy) | 3 | 3 | 3 | |
| | Linear position, stray-field Immune | 4 | 4 | 4 | |
| FILTERING | | | | | |
| FILTER | FIR filter bandwidth selection | 0 | 0 | 0 | 2 |
| HYST | Hysteresis of the denoising filter (hides the small variations but lowers the resolution) Denoising and FIR filtering are applied before the linearization step. | 0 | 0 | 0 | 8 |
| DENOISING_FILTER_ALPHA_SEL | Select the alpha parameter of the EMA (IIR) filter | 0 | 0 | 0 | 2 |
| SPC_RX_FILT_TIME_INT SPC_RX_FILT_TIME_FRACT | Filtered SPC trigger pulse: Min pulse tick time = $(8 * SPC_RX_FILT_TIME_INT + SPC_RX_FILT_TIME_FRACT)/8$ | N/A | 6 0 | N/A | 4 3 |
| SPC_BLANKING_TIME | SPC blanking time (= $5+2*SPC_BLANKING_TIME$ tick) in case of invalid SPC -ID pulse detected, counted from the rising edge of the trigger pulse. Any other (parasitic) pulse occurring during that time will be ignored. | N/A | 0 | N/A | 2 |

LINEAR TRANSFER CHARACTERISTIC

| Parameter | Description | Default Values | | | |
|------------------------------------|---|----------------|--------|-------------|-------|
| | | SENT | SPC | Analog /PWM | #bits |
| DP | DSP discontinuity point: New Angle = Angle - DP | 0 | 0 | 0 | 16 |
| CW | Magnet rotation direction, see section 14.2.2 | 0 | 0 | 0 | 1 |
| 4POINTS | Enable 4 points PWL linearization | 0 | 0 | 0 | 1 |
| DSP_LNR_RESX2 | Enable a doubled LNR method 0: 4-points or 16-segments 1: 8-points or 32-segments | 0 | 0 | 0 | 1 |
| WORK_RANGE_GAIN | Post DSP Gain stage, applied just before linearization, details see section 14.2.9 | 16 | 16 | 16 | 8 |
| GAIN_ANCHOR_MID | Select 50 % position as anchor point Post DSP Gain stage | 1 | 1 | 1 | 1 |
| USEROPTION_SCALING | Enables output scaling x2 0 = [0..100%] 1 = [-50%..150%] | 1 | 1 | 1 | 1 |
| LNR_SO | 4-pts - Slope coefficient before reference point A | N/A | N/A | N/A | 16 |
| LNR_A_X, LNR_B_X, LNR_C_X, LNR_D_X | 4-pts - X Coordinate for reference points A,B,C,D | N/A | N/A | N/A | 16 |
| LNR_A_Y, LNR_B_Y, LNR_C_Y, LNR_D_Y | 4-pts - Y Coordinate for reference points A,B,C,D | N/A | N/A | N/A | 16 |
| LNR_A_S, LNR_B_S, LNR_C_S, LNR_D_S | 4-pts - Slope coefficient for reference points A,B,C,D | N/A | N/A | N/A | 16 |
| LNR_Y00 | 17-pts / 16 segments - Y coordinate point 0 | 0x4009 | 0x4009 | 0x4CCD | 16 |
| LNR_Y01 | 17-pts / 16 segments - Y coordinate point 1 | 0x4804 | 0x4804 | 0x5334 | 16 |
| LNR_Y02 | 17-pts / 16 segments - Y coordinate point 2 | 0x5000 | 0x5000 | 0x599A | 16 |
| LNR_Y03 | 17-pts / 16 segments - Y coordinate point 3 | 0x57FC | 0x57FC | 0x6000 | 16 |
| LNR_Y04 | 17-pts / 16 segments - Y coordinate point 4 | 0x5FF8 | 0x5FF8 | 0x6667 | 16 |
| LNR_Y05 | 17-pts / 16 segments - Y coordinate point 5 | 0x67F4 | 0x67F4 | 0x6CCD | 16 |
| LNR_Y06 | 17-pts / 16 segments - Y coordinate point 6 | 0x6FF0 | 0x6FF0 | 0x7334 | 16 |
| LNR_Y07 | 17-pts / 16 segments - Y coordinate point 7 | 0x77EC | 0x77EC | 0x799A | 16 |
| LNR_Y08 | 17-pts / 16 segments - Y coordinate point 8 | 0x7FE8 | 0x7FE8 | 0x8000 | 16 |
| LNR_Y09 | 17-pts / 16 segments - Y coordinate point 9 | 0x87E4 | 0x87E4 | 0x8667 | 16 |
| LNR_Y10 | 17-pts / 16 segments - Y coordinate point 10 | 0x8FE0 | 0x8FE0 | 0x8CCD | 16 |

| Parameter | Description | Default Values | | | |
|-------------------------|---|----------------|--------|-------------|-------|
| | | SENT | SPC | Analog /PWM | #bits |
| LNR_Y11 | 17-pts / 16 segments - Y coordinate point 11 | 0x97DC | 0x97DC | 0x9333 | 16 |
| LNR_Y12 | 17-pts / 16 segments - Y coordinate point 12 | 0x9FD8 | 0x9FD8 | 0x999A | 16 |
| LNR_Y13 | 17-pts / 16 segments - Y coordinate point 13 | 0xA7D4 | 0xA7D4 | 0xA000 | 16 |
| LNR_Y14 | 17-pts / 16 segments - Y coordinate point 14 | 0xAFD0 | 0xAFD0 | 0xA667 | 16 |
| LNR_Y15 | 17-pts / 16 segments - Y coordinate point 15 | 0xB7CC | 0xB7CC | 0xACCD | 16 |
| LNR_Y16 | 17-pts / 16 segments - Y coordinate point 16 | 0xBFC8 | 0xBFC8 | 0xB333 | 16 |
| LNR_X[00..07] | 8-pts - X coordinate point 0..7 | N/A | N/A | N/A | 16 |
| LNR_DELTA_Y[00..32] | 33-pts / 32 segments - Delta Y coordinate point 0..32 | N/A | N/A | N/A | 8 |
| LNR_DELTA_Y_EXPAND_LOG2 | Adjust the span of 32 segments NV_LNR_DELTA_Y{n} offsets | 0 | 0 | 0 | 2 |
| CLAMPLOW | Low clamping value of angle output data | 16 | 16 | 32767 | 16 |
| CLAMPHIGH | High clamping value of angle output data | 65408 | 65408 | 32767 | 16 |
| OUTSLOPE_SEL | Possible output offset correction within the post-DSP chain: 0: disabled (default) 1: enabled, applied after discontinuity point stage 2: enabled, applied before clamping stage 3: disabled (unused) | 0 | 0 | 0 | 2 |
| OUTSLOPE_COLD | Slope coefficient at cold of the programmable temperature-dependent offset. Used only if NV_OUTSLOPE_SEL enabled | 0 | 0 | 0 | 8 |
| OUTSLOPE_HOT | Slope coefficient at hot of the programmable temperature-dependent offset. Used only if NV_OUTSLOPE_SEL enabled | 0 | 0 | 0 | 8 |
| DIAGNOSTICS | | | | | |
| DIAG_GLOBAL_EN | Diagnostics global enable. Do not modify! (see section 15.2 Safety Mechanisms) | 1 | 1 | 1 | 1 |
| DIAG_TEMP_THR_LOW | Temperature threshold & limiter for lower-temperature diagnostic. Default=08 -> -57 deg.C | 8 | 8 | 8 | 8 |
| DIAG_TEMP_THR_HIGH | Temperature threshold & limiter for over-temperature diagnostic. Default=128 -> 183 deg.C | 128 | 128 | 128 | 8 |

| Parameter | Description | Default Values | | | |
|-------------------------|--|----------------|-----|-------------|-------|
| | | SENT | SPC | Analog /PWM | #bits |
| DIAG_FIELDTOOLOWTH RES | Field strength limit under which a fault is reported. (see 14.5.4) | 6 | 6 | 0 | 8 |
| DIAG_FIELDTOOHIGHTH RES | Field strength limit over which a fault is reported. (see 14.5.4) | 255 | 255 | 255 | 8 |
| DIAG_DEBOUNCE_STEP DOWN | Diagnostic debouncing stepdown time used for recovery time setting | 1 | 1 | 1 | 4 |
| DIAG_DEBOUNCE_STEP UP | Diagnostic debouncing stepup time used for hold time setting | 1 | 1 | 1 | 4 |
| DIAG_DEBOUNCE_THRE SH | Diagnostic debouncing threshold | 1 | 1 | 1 | 6 |
| COLD_SAFE_STARTUP_E N | Normal (0) or safe start-up (1) after power-on reset | 0 | 0 | 0 | 1 |
| OUT_ALWAYS_HIGHZ | Forces the OUT pin in high-Z mode (for test mode, Do not modify!) | 0 | 0 | 0 | 1 |
| SENT_REPORT_MODE_ ANA | 1: Enables the error message within SENT/SPC frame in ANA diagnostic mode. See DIAG_FAULT_CODE | 0 | 0 | N/A | 2 |
| DIAG_FAULT_CODE | Defines the error message = DIAG_FAULT_CODE + FF9 | 6 | 6 | N/A | 3 |
| DAC_REPORT_MODE_A NA | Defines the DAC state in analog-fault report mode | N/A | N/A | 0 | 2 |
| SENT_INIT_GM | Initialization frame definition until first valid data 0: 000 1 : DIAG_FAULT_CODE + FF9 | 0 | 0 | N/A | 1 |
| SPC_SCN_INIT | SCN nibble Initialization contents until first valid data | N/A | 0 | N/A | 0 |
| OUT_DIAG_HIZ_TIME | Output Digital failure (HiZ) recovery time. = 6 + OUT_DIAG_HIZ_TIME ms | 0 | 0 | 0 | 5 |

OUTPUT CONFIGURATIONS

| | | | | | |
|--------------|---|---|---|---|---|
| PROTOCOL | 0: SENT without pause 1: PWM 2: SENT with pause 3: SPC (with or without ID) 4: Analog output 5: SENT-B (SENT rev 5) - reserved for future use | 2 | 3 | 4 | 3 |
| ABE_OUT_MODE | HW backend output-amplifier mode selection 0: Analog output (12 bit) 1: Digital output with open-drain-Nmos 2: Digital output with open-drain-Pmos 3: Digital output with Push-Pull 4: Digital output with improved emission | 4 | 4 | 0 | 3 |

| Parameter | Description | Default Values | | | |
|-----------------------------|--|----------------|------|-------------|-------|
| | | SENT | SPC | Analog /PWM | #bits |
| ABE_DAC_SEQ_BYPASS | DAC sequencer bypass option | 0 | 0 | 1 | 1 |
| DAC_SEQ_LUT_SEL | DAC Sequence LUT, default for 3μs tick time. Do not modify! | 6 | 1 | N/A | 4 |
| NV_DAC_SEQ_CLK_DIV | DAC Sequence LUT, default for 3μs tick time. Do not modify! | 0 | 0 | N/A | 3 |
| NIBBLE_PULSE_CONFIG | Sets the SENT nibble high/low-time configuration 2: 5 fixed ticks low time (Default) 3: 6 fixed ticks high time | 2 | 2 | N/A | 2 |
| SPC_PP_OPTION | SPC push-pull and high-Z switching option 0: switch on Push-pull mode @ SPC_OUT_ON_TH , switch it off @ SPC_OUT_OFF_TH (PROTOCOL = 3) 1: switch on Push-pull mode @ SPC_OUT_ON_TH, switch it off right after the last rising edge of the pause pulse. (PROTOCOL = 3) 2: Permanent High-Z (do not use) 3: Permanent state defined by NV_ABE_OUT_MODE (to be used for analog , PWM & SENT) | 0 | 0 | 3 | 2 |
| SPC_OUT_ON_TH | Tick delay between the SPC trigger falling edge and the output mode switch: open-drain -> push-pull. Note: tmlowSPC - SPC_RX_FILT_TIME_INT < SPC_OUT_ON_TH < 90 | 0x32 | 0x32 | N/A | 9 |
| SPC_OUT_OFF_TH | Tick delay between the SPC trigger falling edge (filtered) and the output mode switch: push-pull -> open-drain. | 450 | 450 | N/A | 9 |
| PWM PROTOCOL OPTIONS | | | | | |
| PWM_POL | Invert the PWM polarity 0: PWM Period start = falling edge , % DC active low | 2 | N/A | N/A | 2 |
| PWM_REPORT_MODE_ANA | Error message within PWM frame 0x0: PWM DC_FAULT 0x1: HiZ 0x2: V _{SS} 0x3: V _{DD} | N/A | N/A | 0 | 2 |
| PWM_DC_FAULT | PWM Duty Cycle in case of Fault | N/A | N/A | N/A | 8 |
| PWM_DC_FIELDTOOLow | PWM Duty Cycle in case of Field Strength Too Low | N/A | N/A | N/A | 8 |
| SPC PROTOCOL OPTIONS | | | | | |
| SENT_TICK_TIME | SPC mode | N/A | 3 | N/A | 3 |

| Parameter | Description | Default Values | | | |
|----------------------|--|----------------|--------------------|-------------|-------|
| | | SENT | SPC | Analog /PWM | #bits |
| SPC_ID_EN | 0: SPC without ID selection 1: SPC with ID selection. See SPC_CHIP_ID (default) | N/A | 1 | N/A | 1 |
| SPC_CHIP_ID | SPC trigger ID upon which the SENT response will be transmitted on the bus. | N/A | 0: Die1 1: Die2 | N/A | 2 |
| SPC_MEAS_ID | SPC trigger ID upon which the ADC measurement will be done | N/A | 1 | N/A | 2 |
| SERIAL_CONFIG | 1: NSP 2: SSP 3: ESP | 3 | 1 ⁽⁴²⁾ | N/A | 2 |
| ID_IN_STATUS | Enable SPC ID in SCN (status and communication nibble) | N/A | 1 | N/A | 1 |
| SPC_SCN_BIT_ORDER | Enables SCN bit order according to SPC 1: [b3:2] = Error indication , [b1:0] = SPC_CHIP_ID | N/A | 1 | N/A | 1 |
| SPC_TRIGGER_MODE | SPC trigger pulse treatment mode: 0: Synchronous Transmission Mode 1: Bus mode with constant length trigger pulse (default) | N/A | 1 | N/A | 2 |
| SPC_FORMAT | SPC frame sequence: 8: see SENT_FC_FORMAT[2:0] | 8 | 0 | N/A | 4 |
| SPC_FC_CSUM_CFG | Fast-channel checksum calculation method 0: checksum in line with SAE J2716 (default) 1: Method "O" in line with SPC2014 Specification 2: Method "E" in line with SPC2014 Specification | N/A | 0 | N/A | 2 |
| SENT_LEGACY_CRC | Enable legacy CRC calculation 0: The SAE J2716 recommended CRC is calculated | 0 | 0 | N/A | 1 |
| STATUS_IN_CRC | 1: Add SCN in checksum calculation | 0 | 1 | N/A | 1 |
| SPC_CSUM_MODE | Checksum nibble calculation mode in SPC: 0: checksum only (default) 1: Add ID in checksum 2: Add ID & RC in checksum | N/A | 0 | N/A | 2 |
| SPC_BLANK_SENT_FRAME | reserve | N/A | 0 | N/A | 1 |
| SPC_TRIG_TH | Delay between the filtered SPC trigger falling edge and the ADC start of sequence, given in number of tick. | N/A | 0 | N/A | 9 |

⁴² Used for SCN configuration, **do not modify!** See parameter "ID_IN_STATUS". 2 bits are overlaid with parameter "SPC_CHIP_ID".

| Parameter | Description | Default Values | | | |
|-------------------------------------|---|----------------|-------|-------------|-------|
| | | SENT | SPC | Analog /PWM | #bits |
| SENT PROTOCOL OPTIONS | | | | | |
| SENT_TICK_TIME | Normal SENT low power mode | 3 | N/A | N/A | 3 |
| SENT_FC_FORMAT | SENT format option (only if SPC_FORMAT = 8) 1: Format H.1 (A.1, Two 12-bit Fast Channels) 2: Format H.2 (One 12-bit Fast Channel, 3 x 4 bit) 3: Format H.3 (One 12-bit fast channel, 4 x 3 bit) 4: Format H.4 (A.3, 12-bit Fast Channel & Single secure) 5: Format H.5 (12-bit Fast Channel & zero values) 6: Format H.6 (14-bit Fast Channel1 & 10-bit Fast Channel2) 7: Format H.7 (16-bit Fast Channel & 8-bit Fast Channel2) | 4 | N/A | N/A | 3 |
| SENT_FC_FORMAT_EN | SENT format enable | 1 | 0 | N/A | 1 |
| TWO_ANGLES_FRAME | Enable 2 fields measure per output frame (SENT with pulse) | 1 | N/A | N/A | 1 |
| T_FRAME | SENT w/ PP frame length PWM period (4µs/LSB) | 320 | N/A | N/A | 12 |
| T_SYNC_DELAY | SENT - ADC synchronization delay - used in SENT-with-pause only | 27 | N/A | N/A | 12 |
| SENT_FAST_CHANNEL_2 | Definition of data transmitted in the SENT fast channel 2 in case SENT_FC_FORMAT=1, 6 or 7 0 Temperature sensor (SP ID 0x23) 1 0xFF9(d4089) - CH1 2 RAM data (RAMPROBE_PTR) 3 0xFFF(d4095) - CH1 | 0 | N/A | N/A | 2 |
| SENT_SLOW_EXTENDED | ESM length configuration | 0 | N/A | N/A | 1 |
| SENT_SLOW_EXTENSION | ESM length configuration | 0 | N/A | N/A | 2 |
| SENT_DIAG_STRICT | ESM length configuration | 0 | N/A | N/A | 1 |
| SERIAL MESSAGE CONFIGURATION | | | | | |
| SENT_SENSOR_TYPE | Part of SSM/ESM : Channel 1/2 sensor type | 0x050 | 0x050 | N/A | 12 |
| SENT_MAN_CODE | Part of SSM/ESM: Manufacturer code | 0x000 | 0x000 | N/A | 12 |
| SENT_REV | Part of SSM/ESM: SENT standard revision | 0x4 | 0x4 | N/A | 4 |
| SENT_CHANNEL_X1 | Part of SSM/ESM: Fast-channel 1 - X1 | 0 | 0 | N/A | 12 |
| SENT_CHANNEL_X2 | Part of SSM/ESM: Fast-channel 1 - X2 | 0 | N/A | N/A | 12 |
| SENT_CHANNEL_Y1 | Part of SSM/ESM: Fast-channel 1 - Y1 | 0 | N/A | N/A | 12 |

| Parameter | Description | Default Values | | | |
|-----------------|--------------------------------------|----------------|-----|-------------|-------|
| | | SENT | SPC | Analog /PWM | #bits |
| SENT_CHANNEL_Y2 | Part of SSM/ESM: Fast-channel 1 - Y2 | 0 | N/A | N/A | 12 |
| SENT_SENSOR_ID1 | Part of SSM/ESM : Sensor ID-1 | 0 | N/A | N/A | 12 |
| SENT_SENSOR_ID2 | Part of SSM/ESM : Sensor ID-2 | 0 | N/A | N/A | 12 |
| SENT_SENSOR_ID3 | Part of SSM/ESM : Sensor ID-3 | 0 | N/A | N/A | 12 |
| SENT_SENSOR_ID4 | Part of SSM/ESM : Sensor ID-4 | 0 | N/A | N/A | 12 |
| SENT_OEM_CODE1 | Part of SSM/ESM : OEM code 1 | 0 | N/A | N/A | 12 |
| SENT_OEM_CODE2 | Part of SSM/ESM : OEM code 2 | 0 | N/A | N/A | 12 |
| SENT_OEM_CODE3 | Part of SSM/ESM : OEM code 3 | 0 | N/A | N/A | 12 |
| SENT_OEM_CODE4 | Part of SSM/ESM : OEM code 4 | 0 | N/A | N/A | 12 |
| SENT_OEM_CODE5 | Part of SSM/ESM : OEM code 5 | 0 | N/A | N/A | 12 |
| SENT_OEM_CODE6 | Part of SSM/ESM : OEM code 6 | 0 | N/A | N/A | 12 |
| SENT_OEM_CODE7 | Part of SSM/ESM : OEM code 7 | 0 | N/A | N/A | 12 |
| SENT_OEM_CODE8 | Part of SSM/ESM : OEM code 8 | 0 | N/A | N/A | 12 |

Table 49 - MLX90377 End-User Programmable Items Table

Performances described in this document are only achieved by adequate programming of the device. To ensure desired functionality, Melexis recommends following its programming guide and to contact its technical or application service.

13. End User Identification Items

| Parameter | Description | Default Values | | |
|-------------|--|--|------|-------|
| | | Analog | SENT | #bits |
| USER_ID0 | Bin1 from production test, can also be reserved for end-user to program information to keep traceability | 1 | | 8 |
| USER_ID1 | Reserve for end-user to program information to keep traceability | N/A | | 8 |
| USER_ID2 | Reserve for end-user to program information to keep traceability | 0x20 | | 8 |
| USER_ID3 | Reserve for end-user to program information to keep traceability | N/A | | 8 |
| USER_ID4 | Reserve for end-user to program information to keep traceability | N/A | | 8 |
| USER_ID5 | Reserve for end-user to program information to keep traceability | N/A | | 8 |
| IMC_VERSION | IMC / application type: 0: Sun (rotary stray field robust low field variant) 1: Clover (standard/legacy) 2: Dual-disk (linear stray field robust) 4: Bubble (rotary stray field robust high field variant) | 0 for -1xx order code 1 for -3xx order code 2 for -2xx order code 4 for -5xx order code | | 6 |
| TEST_STATUS | Bin 1 verification bit | 1 | | 1 |
| MLX_ID0 | X-Y position on the wafer (8 bit each) | MLX | | 16 |
| MLX_ID1 | Wafer ID [b15..11] Lot ID [b10..0] | MLX | | 16 |
| MLX_ID2 | Fab ID (4 bits) Test Database ID (6 bits) | MLX | | 16 |

Table 50 - Melexis and Customer ID fields description

User identification numbers (48 bits) are freely usable by customers for traceability purpose. Other IDs are read only.

14. Description of End-User Programmable Items

14.1. Output Modes

14.1.1. OUT mode

Output Stage mode (outside fail-safe state) in application can be defined by the parameters “ABE_OUT_MODE” and “ABE_DAC_SEQ_BYPASS”:

| ABE_OUT_MODE | ABE_DAC_SEQ_BYPASS | Description | Comments |
|--------------|--------------------|---------------------------------------|--|
| 0 | 1 | Analog output (12-bit DAC) | |
| 1 | 1 | Digital output with open-drain-Nmos | Requires a pull-up resistor on output. |
| 2 | 1 | Digital output with open-drain-Pmos | Requires a pull-down resistor on output. |
| 3 | 1 | Digital output with Push-Pull | |
| 4 | 0 | Digital output with improved emission | |

Table 51 - Output Mode Selection

14.1.2. Digital OUT protocol

The measurement timing mode and the corresponding output protocol can be selected by the parameter “PROTOCOL”:

| PROTOCOL | Descriptions |
|----------|--|
| 0 | Continuous asynchronous angle acquisition, SENT without pause |
| 1 | Continuous asynchronous angle acquisition, PWM |
| 2 | Continuous synchronous angle acquisition, SENT with pause |
| 3 | master-triggered angle transmission, SPC (with or without ID) |
| 4 | continuous asynchronous angle acquisition, analog output (DAC) |

Table 52 - Protocol Selection

14.1.3. Serial Channel Configuration

Slow Channel configuration can be done by the parameter “SERIAL_CONFIG”:

| SERIAL_CONFIG | Descriptions |
|---------------|--|
| 0 | Slow Channel isn't present, the whole nibble is skipped => violating the protocol. Do not use. |
| 1 | No serial message |
| 2 | Short serial message |
| 3 | Enhanced serial message |

Table 53 - SENT Serial channel Configuration

14.1.4. PWM Output Mode

If PWM output mode is selected, the output signal is a digital signal with Pulse Width Modulation (PWM). The PWM polarity is selected by the parameter “PWM_POL”:

- PWM_POL = 0 for a low level at 100%
- PWM_POL = 1 for a high level at 100%

The PWM frequency is selected in the range [100, 2000] Hz by the T_FRAME parameter (12bits), defining the period time in the range [0.5; 10] ms. Minimum allowed value for T_FRAME is therefore 125 (0x7d).

$$T_{PWM} = \frac{4}{10^6} \times T_{FRAME}$$

The PWM period is subject to the same tolerances as the main clock (see ΔT_{ck}).

14.1.5. SPC Frame Formats

The SPC frame formats (see section 11.4.1) can be defined by the parameter “SPC_FORMATS”, with the following overview:

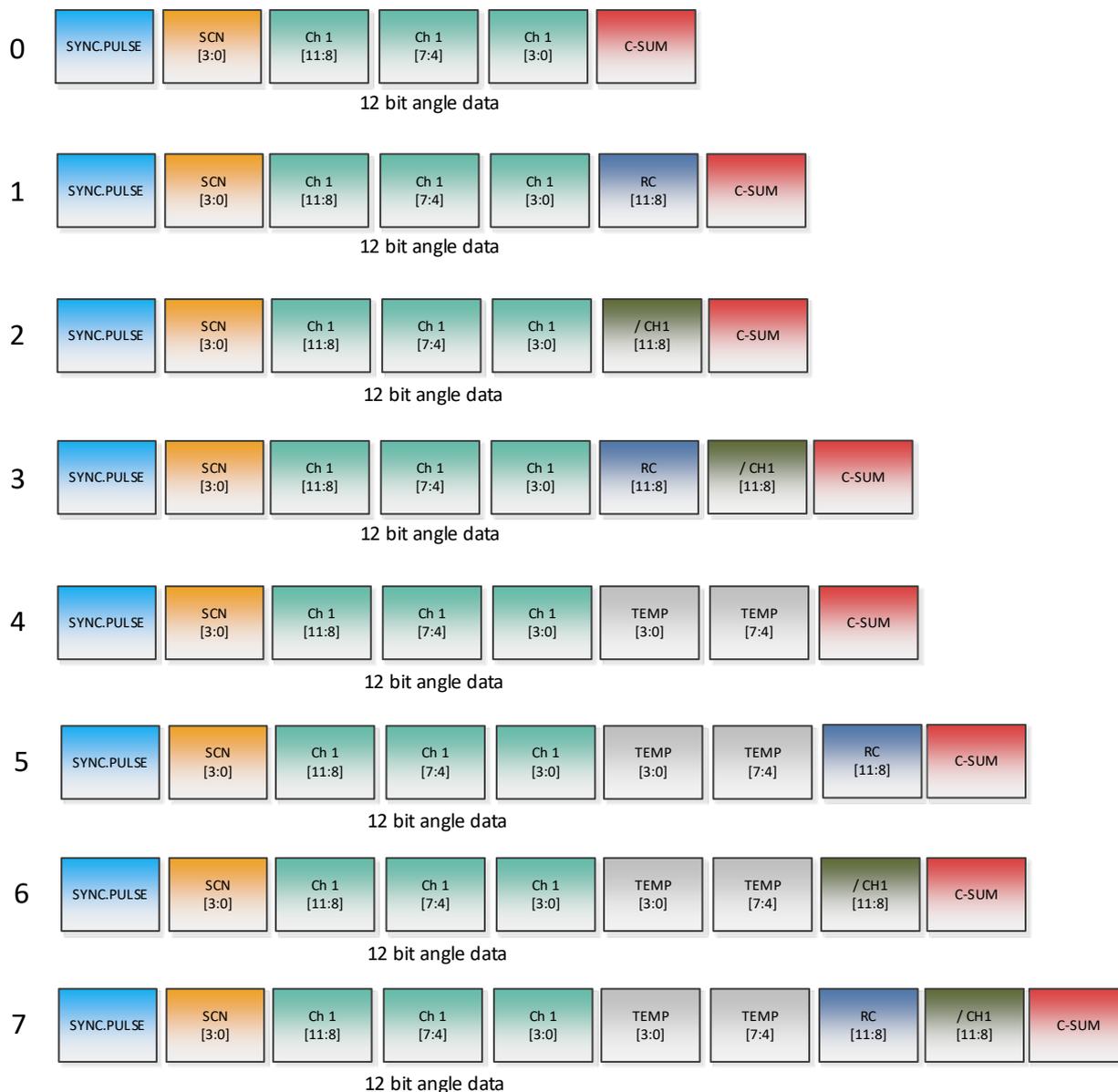


Figure 26 – SPC Frame Format details

14.2. Output Transfer Characteristic

Figure 27 gives the simplified digital signal processing chain from the position after ADC to the output. This section explains the compensation capability of the IC. The remainder of this chapter explains every parameter in more detail.

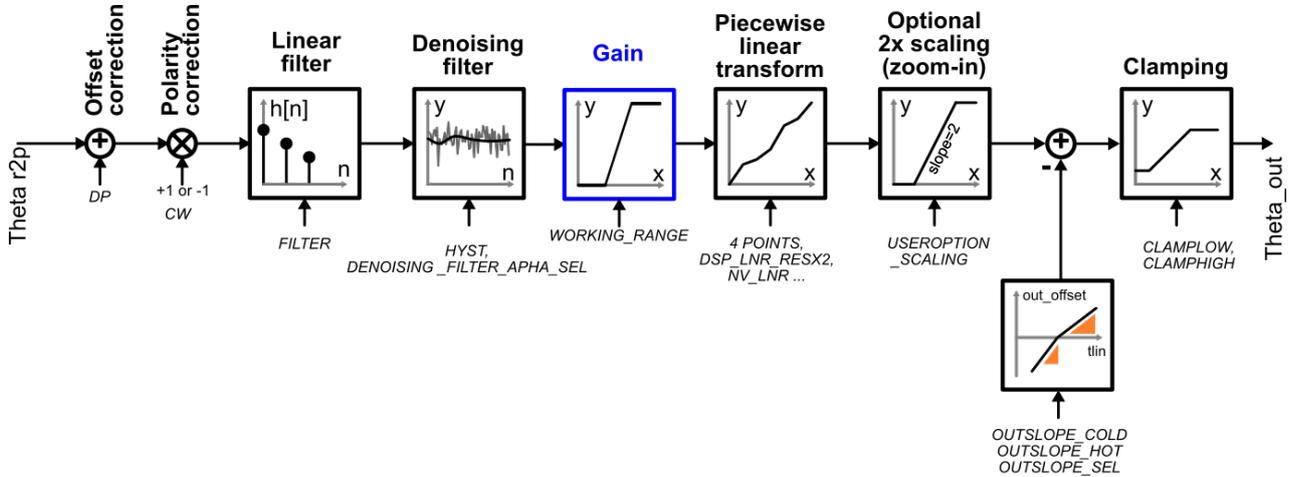


Figure 27 – A Digital Signal Process Chain from ADC to the Output of MLX90377

There are 4 different possibilities to define the transfer function (LNR) as specified in the Table 54.

- With 4 arbitrary points (defined by X and Y coordinates) and 5 slopes
- With 8 arbitrary points (defined by X and Y coordinates)
- With 17 equidistant points for which only the Y coordinates are defined
- With 32 equidistant points for which only offset of Y compared to the average value is defined

| Output Transfer Characteristic | 4 Points | DSP_LNR_RESX2 |
|--------------------------------|----------|---------------|
| 4 Arbitrary Points | 1 | 0 |
| 8 Arbitrary Points | 1 | 1 |
| 17 Equidistant Points | 0 | 0 |
| 32 Equidistant Points | 0 | 1 |

Table 54 - Output Transfer Characteristic Selection Table

| Parameter | LNR type | Value | Unit |
|--|----------------------|---------------------------------------|------|
| CW | All | 0 → counterclockwise 1 → clockwise | LSB |
| DP | All | 0 ... 359.9999 | deg |
| LNR_A_X LNR_B_X LNR_C_X LNR_D_X | 4 pts, X coordinates | 0 ... 359.9999 | deg |

| Parameter | LNR type | Value | Unit |
|--|-------------------------|---|-------|
| LNR_A_Y LNR_B_Y LNR_C_Y LNR_D_Y | 4 pts, Y coordinates | 0 ... 100 -50 ... + 150 | % |
| LNR_S0 LNR_A_S LNR_B_S LNR_C_S LNR_D_S | 4 pts, slopes | -17 ... 0 ... 17 | %/deg |
| LNR_X00..X07 | 8 pts, X coordinates | 0 ... 359.9999 | deg |
| LNR_Y00..Y16 | 8,17 pts, Y coordinates | 0..100 -50 ... + 150 | % |
| LNN_DELTA_Y00..Y32 | 33 pts offsets | +/-3.125% +/-6.25% +/-12.5% +/-25% | % |
| WORKING RANGE | 17/33 pts | 65.5 ... 360 32.75 ... 180 | deg |
| CLAMPLOW | All | 0 ... 100 | % |
| CLAMPHIGH | All | 0 ... 100 | % |

Table 55 - Output linearization and clamping parameters

14.2.1. Enable scaling Parameter

This parameter enables to double the scale of Y coordinates linearisation parameters from [0 .. 100]% to [-50 .. 150]% according to the following table (Table 56). This is valid for all linearisation schemes except the 32 points.

| USEROPTION_SCALING | LNR_Y min value | LNR_Y max value |
|--------------------|-----------------|-----------------|
| 0 | 0% | 100% |
| 1 | -50% | 150% |

Table 56 - USEROPTION_SCALING parameter

14.2.2. CW (Clockwise) Parameter

The CW parameter defines the magnet rotation direction.

- 0 or counter clockwise is defined by 1-4-5-8 pin order direction for the SOIC-8 package, 1-8-9-16 pin order direction for the TSSOP-16 package, 1-2-3-4 pin order direction for the DMP-4 package, 1-2-3 pin order direction for the SMP-3 package;
- 1 or clockwise is defined by the reverse direction: 8-5-4-1 pin order direction for the SOIC-8 package, 16-9-8-1 pin order direction for the TSSOP-16 package, 4-3-2-1 pin order direction for the DMP-4 package, 3-2-1 pin order direction for the SMP-3 package.

Refer to the drawing in the sensitive spot positioning section (19.1.4, 19.2.4, 19.3.6).

14.2.3. Discontinuity Point (or Zero Degree Point)

The Discontinuity Point defines the 0° point on the circle. The discontinuity point places the origin at any location of the trigonometric circle. The DP is used as reference for all the angular measurements.

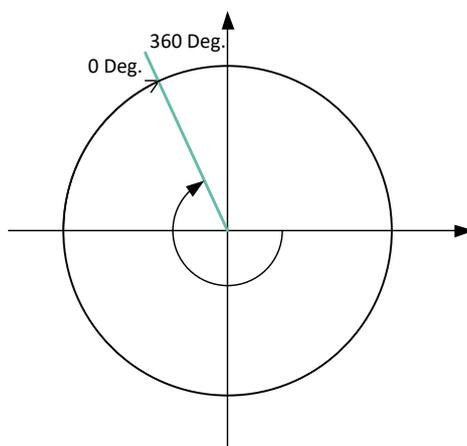


Figure 28 - Discontinuity Point Positioning

14.2.4. 4-pts LNR Parameters

The LNR parameters, together with the clamping values, fully define the relation (the transfer function) between the digital angle and the output signal.

The shape of the MLX90377 4-pts transfer function from the digital angle value to the digital output is described in Figure 29. Seven segments can be programmed but the clamping levels are always flat.

Two to six calibration points are available, reducing the overall non-linearity of the IC by almost an order of magnitude each time. Three or more calibration points will be preferred by customers looking for excellent non-linearity figures. Two-point calibrations will be preferred by customers looking for a lower cost calibration set-up and shorter calibration time.

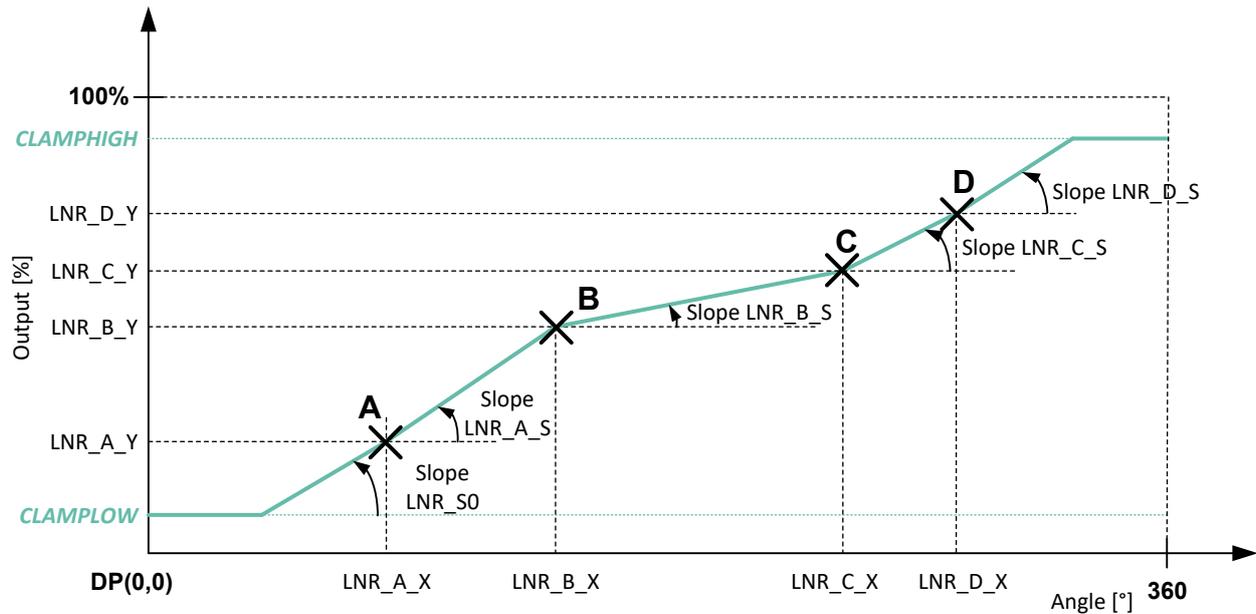


Figure 29 – 4-pts Linearization Parameters Description

14.2.5. 8-pts LNR Parameters

The LNR parameters, together with the clamping values, fully define the relation (the transfer function) between the digital angle and the output signal.

The shape of the MLX90377 8-pts transfer function from the digital angle value to the output voltage is described in Figure 30⁽⁴³⁾. Eight calibration points [LNR_X0...7, LNR_Y0...7] together with 2 fixed points at the extremity of the range ([0°, 0%] ; [360°, 100%]) divides the transfer curve into 9 segments. Each segment is defined by 2 points and the values in between is calculated by linear interpolation.

⁴³ Applications with maximal operation range of 360°, i.e. ordering code of -2xx / -3xx, shown as example. Applications with maximal operation range of 180°, i.e. ordering code of -1xx / -5xx, can be scaled accordingly.

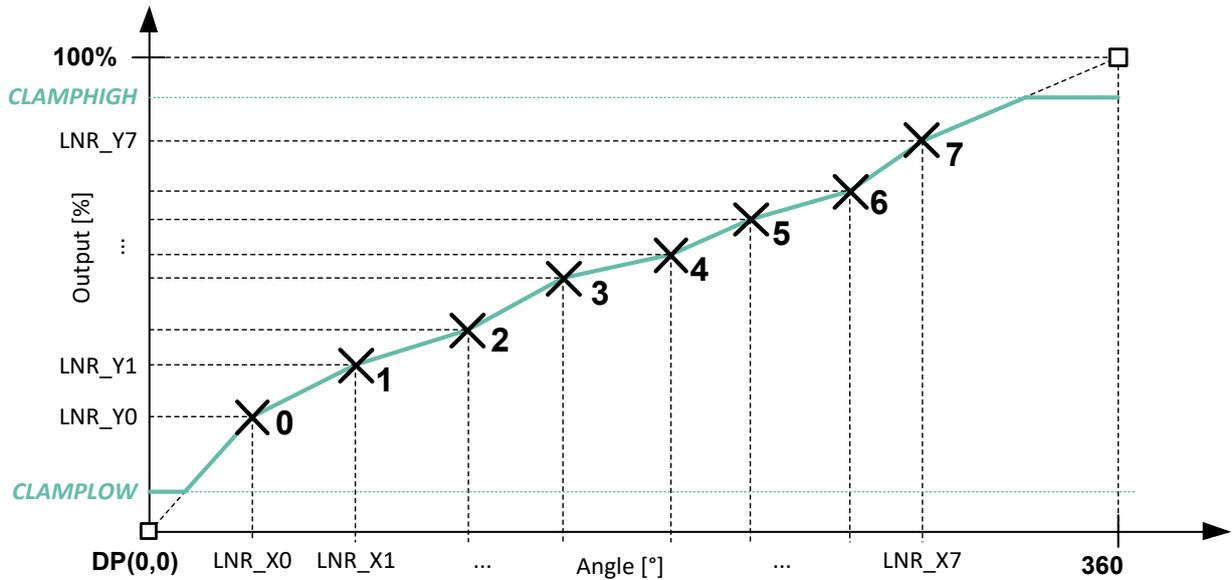


Figure 30 – 8-pts Linearization Parameters Description

14.2.6. 17-pts LNR Parameters

The LNR parameters, together with the clamping values, fully define the relation (the transfer function) between the digital angle and the output signal.

The shape of the MLX90377 17-pts transfer function from the digital angle value to the output voltage is described in Figure 31⁽⁴³⁾. In the 17-pts mode, the output transfer characteristic is Piece-Wise-Linear (PWL).

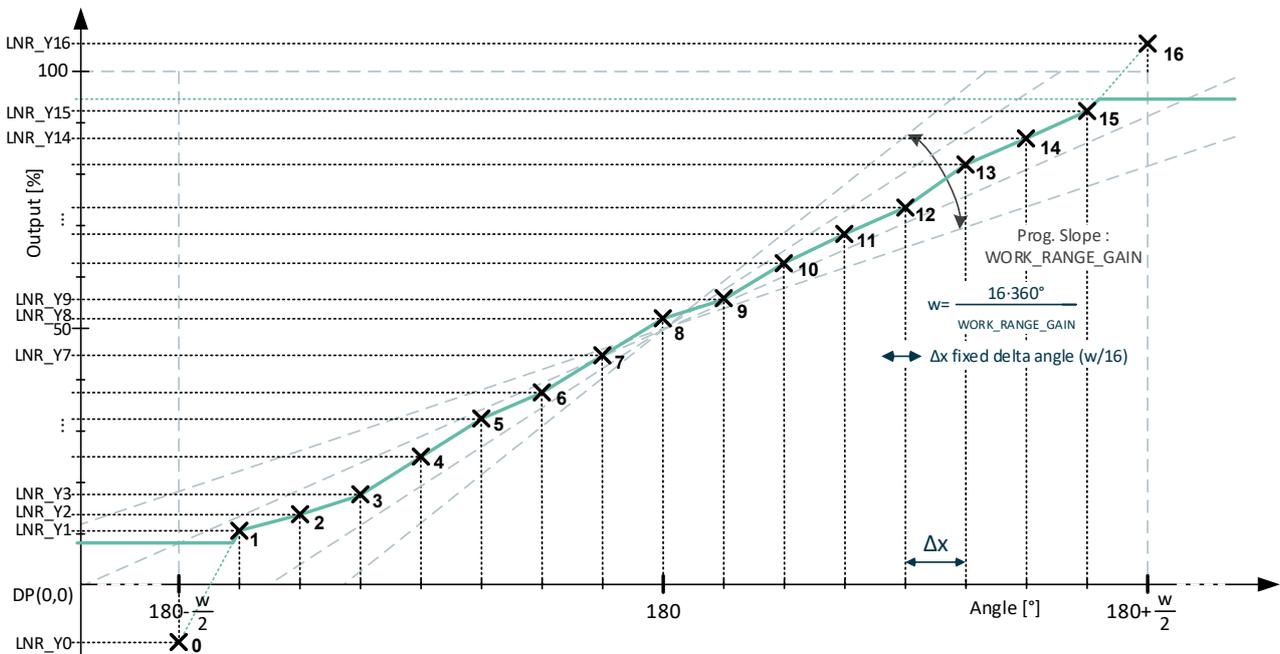


Figure 31 – 17-pts Linearization Parameters Description

All the Y-coordinates can be programmed from -50% up to +150% to allow clamping in the middle of one segment (like on the figure), but the output value is limited to CLAMPLOW and CLAMPHIGH values.

Between two consecutive points, the output characteristic is interpolated.

14.2.7. 33-pts LNR parameters

The LNR parameters, together with the clamping values, fully define the relation (the transfer function) between the digital angle and the output signal.

The shape of the MLX90377 33-pts transfer function from the digital angle value to the output voltage is described in Figure 32 ⁽⁴³⁾ below. In the 33-pts mode, the output transfer characteristic is Piece-Wise-Linear (PWL).

The points are spread evenly across the working range (see sections 14.2.8 and 14.2.9 for working range selection). The Y-coordinates can be offset from the ideal characteristic within an adjustable range defined by LNR_DELTA_Y_EXPAND_LOG2. The available values are summarized in Table 57. All LNR_delta_Y## parameters are encoded in a fractional signed 8-bit value.

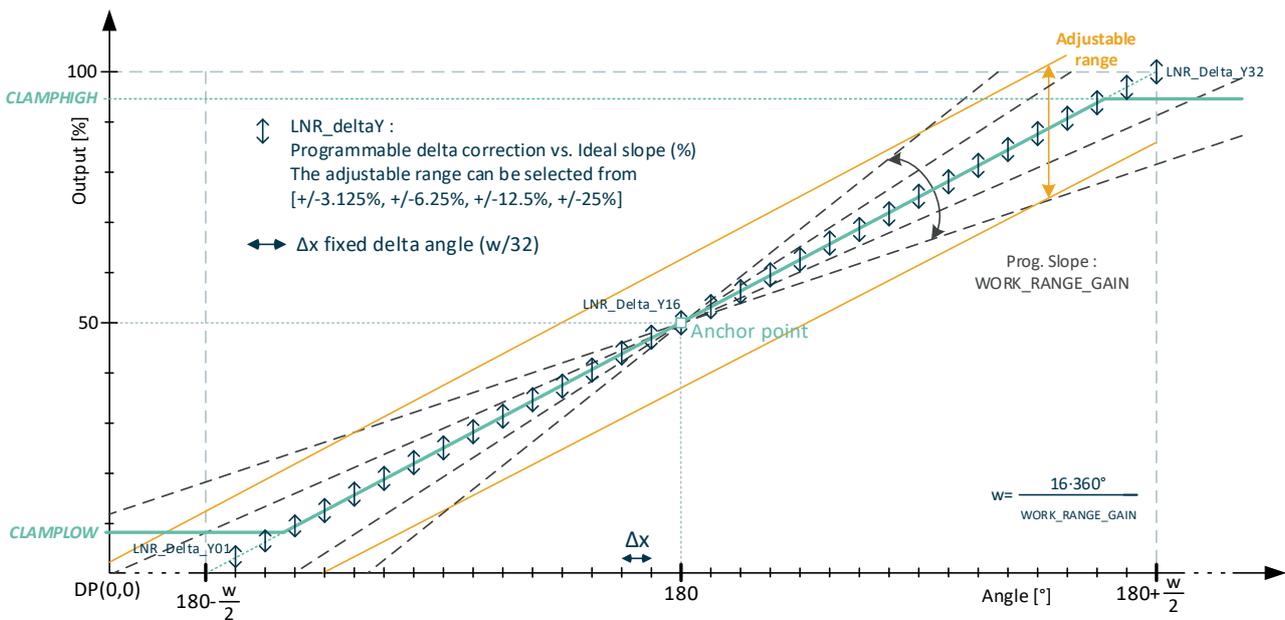


Figure 32 – 33-pts Linearization Parameters Description

| LNR_DELTA_Y_EXP AND_LOG2 | Adjustable Range | Correction resolution |
|--------------------------|------------------|-----------------------|
| 0 | ±3.125% | 0.024% |
| 1 | ±6.25% | 0.049% |
| 2 | ±12.5% | 0.098% |
| 3 | ±25% | 0.20% |

Table 57 - LNR_DELTA_Y_EXPAND_LOG2 values and correction resolution

14.2.8. WORK_RANGE Parameter for Angle Range Selection

The parameter WORK_RANGE determines the input range on which the 16 or 32 segments are uniformly spread. This parameter is provided for compatibility with former versions of Melexis Triaxis sensors. For full featured working range selection, see 14.2.9. For WORK_RANGE parameter, following table applies.

| WORK_RANGE | Range | Δx 17pts | Δx 32pts | WORK_RANGE | Range | Δx 17pts | Δx 32pts |
|------------|--------|----------|----------|------------|--------|----------|----------|
| 0 | 180.0° | 11.3° | 5.6° | 8 | 90.0° | 5.6° | 2.8° |
| 1 | 160.0° | 10.0° | 5.0° | 9 | 72.0° | 4.5° | 2.3° |
| 2 | 144.0° | 9.0° | 4.5° | 10 | 60.0° | 3.8° | 1.9° |
| 3 | 131° | 8.2° | 4.1° | 11 | 51.45° | 3.2° | 1.6° |
| 4 | 120.0° | 7.5° | 3.8° | 12 | 45.0° | 2.8° | 1.4° |
| 5 | 221.5° | 6.9° | 3.5° | 13 | 40.0° | 2.5° | 1.3° |
| 6 | 103° | 6.4° | 3.2° | 14 | 36.0° | 2.3° | 1.1° |
| 7 | 96° | 6.0° | 3.0° | 15 | 32.75° | 2.0° | 1.0° |

Table 58 - Work range for 180° periodicity (ordering code -1xx, -5xx)

| WORK_RANGE | Range | Δx 17pts | Δx 32pts | WORK_RANGE | Range | Δx 17pts | Δx 32pts |
|------------|--------|----------|----------|------------|--------|----------|----------|
| 0 | 360.0° | 22.5° | 11.3° | 8 | 180.0° | 11.3° | 5.6° |
| 1 | 320.0° | 20.0° | 10.0° | 9 | 144.0° | 9.0° | 4.5° |
| 2 | 288.0° | 18.0° | 9.0° | 10 | 120.0° | 7.5° | 3.8° |
| 3 | 261.8° | 16.4° | 8.2° | 11 | 102.9° | 6.4° | 3.2° |
| 4 | 240.0° | 15.0° | 7.5° | 12 | 90.0° | 5.6° | 2.8° |
| 5 | 221.5° | 13.8° | 6.9° | 13 | 80.0° | 5.0° | 2.5° |
| 6 | 205.7° | 12.9° | 6.4° | 14 | 72.0° | 4.5° | 2.3° |
| 7 | 192.0° | 12.0° | 6.0° | 15 | 65.5° | 4.1° | 2.0° |

Table 59 - Work range for 360° periodicity (ordering code -2xx, -3xx)

Outside of the selected range, the output will remain at clamping levels.

14.2.9. WORK_RANGE_GAIN Parameter for Angle Range Selection

Alternatively, the range for the angle can be selected using the WORK_RANGE_GAIN parameter, which applies a fixed gain to the transfer characteristics. WORK_RANGE_GAIN is coded on 8 bits where the 4 MSB defines the integer part and the 4 LSB the fractional part (in power of twos). Therefore, the following equation applies to define the angle range w:

$$w = \frac{16 * MaxRange}{WORK_RANGE_GAIN}$$

MaxRange depends on the application. It is 360° for ordering code -2xx / -3xx, and 180° for ordering code -1xx / -5xx (rotary stray-field immune). Both minimal and maximal angles are then defined by:

$$\theta_{min} = \frac{MaxRange - w}{2} ; \theta_{max} = \frac{MaxiRange + w}{2}$$

where θ_{min} corresponds to the angle yielding 0% output and θ_{max} the angle giving a 100% output.

Using WORK_RANGE_GAIN parameter, the anchor point is kept at $MaxRange/2$ and the range is symmetrically set around this value. It creates a zoom-in of the angle around this point.

Following tables give some values as example:

| WORK_RANGE_GAIN | Factor | Range (w) | θ_{min} | θ_{max} | Δx 17pts | Δx 32pts |
|-----------------|--------|-----------|----------------|----------------|------------------|------------------|
| 0x10 | 1 | 180° | 0° | 180° | 11.25° | 5.63° |
| 0x20 | 2 | 90° | 45° | 135° | 5.63° | 2.81° |
| 0x40 | 4 | 45° | 67.5° | 112.5° | 2.81° | 1.41° |
| 0xFF | 15.94 | 11.3° | 78.7° | 101.3° | 0.71° | 0.35° |

Table 60 - Working range defined by WORK_RANGE_GAIN parameter (ordering code -1xx, -5xx)

| WORK_RANGE_GAIN | Factor | Range (w) | θ_{min} | θ_{max} | Δx 17pts | Δx 32pts |
|-----------------|--------|-----------|----------------|----------------|------------------|------------------|
| 0x10 | 1 | 360° | 0° | 360° | 22.5° | 11.3° |
| 0x20 | 2 | 180° | 90° | 270° | 11.3° | 5.6° |
| 0x40 | 4 | 90° | 135° | 225° | 5.6° | 2.8° |
| 0xFF | 15.94 | 22.6° | 168.7° | 191.3° | 1.41° | 0.71° |

Table 61 - Working range defined by WORK_RANGE_GAIN parameter (ordering code -2xx, -3xx)

Outside of the working range, the output will remain at clamping levels.

14.2.10. Thermal OUTSLOPE offset correction

Two parameters, OUTSLOPEHOT and OUTSLOPECOLD, are used to add a temperature dependent offset. This feature is enabled by the parameter OUTSLOPE_SEL that apply this modification either directly to the angle or after the linearisation function. The MLX90377 uses its internal linearized temperature to compute the offset shift as depicted in the figure below (Figure 33).

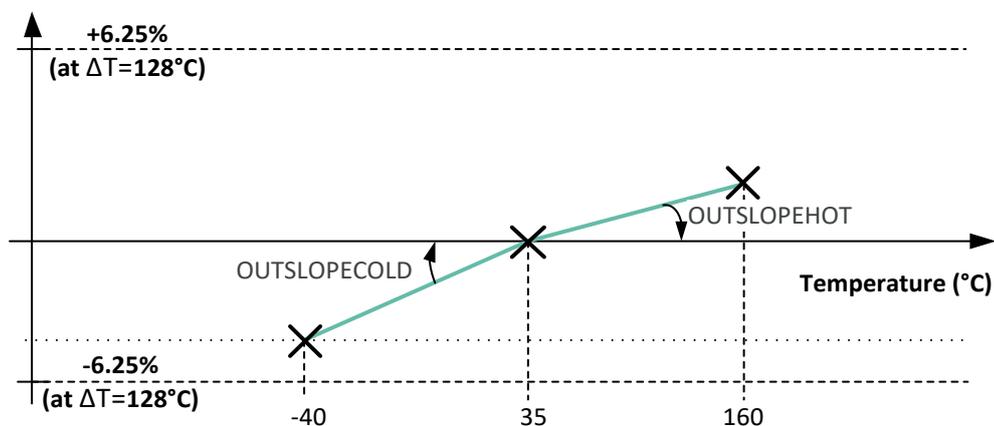


Figure 33 - Temperature compensated offset

The thermal offset can be added or subtracted to the output, before the clamping. The span of this offset is $\pm 6.25\%$ of the full output scale for a temperature difference of 128°C . The added thermal offset varies with temperature following the equations below. The two thermal coefficients are encoded in signed two's complement 8bit format ($-128..127$) and defined separately below 35°C (OUTSLOPECOLD) and above 35°C (OUTSLOPEHOT).

| OUTSLOPE_SEL | Description |
|--------------|--|
| 0 | No thermal offset correction |
| 1 | Thermal offset enabled, applied after angle calculation, i.e. after discontinuity point (θ_{r2p}) |
| 2 | Enabled, applied after output calculation and before clamping (θ_{out}) |

Table 62 - Temperature compensated offset selection parameter

If IC internal temperature is higher than 35°C then:

$$\theta_{Tcomp} = \theta_{in} - \Delta T \cdot \text{OUTSLOPEHOT}$$

If IC internal temperature is lower than 35°C then:

$$\theta_{Tcomp} = \theta_{in} - \Delta T \cdot \text{OUTSLOPECOLD}$$

where θ_{in} is either θ_{r2p} or θ_{out} depending on OUSLOPE_SEL value.

14.2.11. CLAMPING Parameters

The clamping levels are two independent values to limit the output voltage range. The CLAMPLOW parameter adjusts the minimum output level. The CLAMPHIGH parameter sets the maximum output. Both parameters have 16 bits of adjustment and are available for all four LNR modes. As output data resolution is limited to 12 bits, both in SENT and in PWM, the 4 LSB of this parameter will have no significant effect on the output. The value is encoded in fractional code, from 0% to 100%

14.3. Sensor Front-End

| Parameter | Value |
|----------------|---------|
| SENSING MODE | [0..7] |
| GAINMIN | [0..63] |
| GAINMAX | [0..63] |
| GAINSATURATION | [0, 1] |

Table 63 - Sensing Mode and Front-End Configuration

14.3.1. SENSING MODE

The SENSING_MODE parameter defines which sensing mode and fields are used to calculate the angle. The different possibilities are described in the tables below. This 3-bit value selects the first (B1) and second (B2) field components according to the Table 64 content.

| MAPXYZ | B1 | B2 | Angular |
|--------|------------|------------|--|
| 0 | X | Y | Rotary position stray-field Immune, order code -1xx and -5xx |
| 1 | X | Y | X-Y Angular Rotary, order code -3xx |
| 2 | Y | Z | Y-Z Angular Rotary, order code -3xx |
| 3 | X | Z | X-Z Angular Rotary, order code -3xx |
| 4 | ΔX | ΔZ | Linear position, stray-field Immune, order code -2xx |

Table 64 - Sensing Mode Description

14.3.2. GAINMIN and GAINMAX Parameters

GAINMIN and GAINMAX define the thresholds on the gain code outside which the fault “GAIN out of Spec.” is reported. If GAINSATURATION is set, then the virtual gain code is clamped at GAINMIN and GAINMAX, and no diagnostic fault will be set since the clamping will prevent the gain from exceeding GAINMIN and GAINMAX.

14.4. Filtering

The MLX90377 includes 2 types of filters:

- Exponential moving average (EMA) Filter: programmable by the HYST parameter
- Low Pass FIR Filters controlled with the FILTER parameter

| Parameter | Value |
|------------|-----------|
| DSP_FILTER | 0 ... 2 |
| HYST | 0 ... 255 |

Table 65 - Filtering configuration

14.4.1. Exponential Moving Average (IIR) Filter

The HYST parameter is a hysteresis threshold to activate / de-activate the exponential moving average filter. The output value of the IC is updated with the applied filter when the digital step is smaller than the programmed HYST parameter value. The output value is updated without applying the filter when the increment is bigger than the hysteresis. The filter reduces therefore the noise but still allows a fast step response for bigger angle changes. The hysteresis must be programmed to a value close to the internal magnetic angle noise level ($1\text{LSB} = 8 \cdot \text{MaxRange} / 2^{16}$)⁽⁴⁴⁾.

⁴⁴ For the definition of MaxRange, see section 14.2.9.

$$y_n = a * x_n + (1-a) * y_{n-1} \quad \begin{matrix} x_n = \text{Angle} \\ y_n = \text{Output} \end{matrix}$$

The filters characteristic is given in the following table (Table 66):

| DENOISING_FILTER_ALPHA_SEL | 0 | 1 | 2 | 3 |
|----------------------------|------|-----|------|-------|
| Coefficients a | 0.75 | 0.5 | 0.25 | 0.125 |
| Efficiency RMS (dB) | | 2.4 | 4.2 | |

Table 66 - IIR Filter characteristics

14.4.2. FIR Filters

The MLX90377 features 2 FIR filter modes controlled with Filter = 1...2. Filter = 0 corresponds to no filtering. The transfer function is described by:

$$y_n = \frac{1}{\sum_{i=0}^j a_i} \sum_{i=0}^j a_i x_{n-i}$$

This filter characteristic is given in the Table 67.

| FILTER value | 0 | 1 | 2 |
|-----------------------------|-----------|-------------------------------|-------|
| Type | Disable | Finite Impulse Response (FIR) | |
| Coefficients a _i | 1 | 11 | 1111 |
| Title | No filter | Extra light | Light |
| DSP cycles (#taps) | 1 | 2 | 4 |
| Efficiency RMS (dB) | 0 | 3.0 | 6.0 |

Table 67 - FIR Filter Characteristics

14.5. Programmable Diagnostics Settings

14.5.1. Diagnostics Global Enable

DIAG_EN must be kept to its default value (1) to retain all functional safety abilities of the MLX90377. This feature shall not be disabled.

14.5.2. Diagnostic Debouncer

A debouncing algorithm is available for analog diagnostic reporting. Enabling this debouncer will increase the FHTI of the device. Therefore, Melexis recommends keeping the debouncing of analog faults off, by not modifying below described values. The factory default settings mentioned in chapter 12 should be used.

| NVRAM Parameter | Description |
|-----------------------|--|
| DIAGDEBOUNCE_STEPDOWN | Decrement values for debouncer counter |
| DIAGDEBOUNCE_STEPUP | Increment value for debouncer counter |
| DIAG_DEBOUNCE_THRESH | Threshold for debouncer counter to enter diagnostic mode |

Table 68 - Diagnostic debouncing parameters

Once an analog monitor detects an error, it takes control of the debouncing counter. This counter will be incremented by STEPUP value each time this specific monitor is evaluated and the error is still present. When the debouncing counter reaches the value defined by DEBOUNCE_THRESH, an error is reported on the error channel, and the debouncing counter stays clamped to this DEBOUNCE_THRESH value (see section 11.3.8 for SENT error message codes, and 14.5.5 for PWM error reporting). Once the error disappears, each time its monitor is evaluated, the debouncing counter is decremented by STEPDOWN value. When the debouncing counter reaches zero, the error disappears from the reporting channel and the debouncing counter is released. To implement proper reporting times, one should refer to the FHTI, see chapter 15.3. The reporting and recovery time are defined in the table below (valid for THRESH≠0).

| Parameter | Min | Max |
|----------------|---|---|
| Reporting Time | $DCT \cdot \left(\left\lceil \frac{THRESH}{STEPUP} \right\rceil - 1 \right)$ | $DCT \cdot \left(\left\lceil \frac{THRESH}{STEPUP} \right\rceil \right)$ |
| Recovery Time | $DCT \cdot \left(\left\lceil \frac{THRESH}{STEPDOWN} \right\rceil \right)$ | $DCT \cdot \left(\left\lceil \frac{THRESH}{STEPDOWN} \right\rceil + 1 \right)$ |
| | $\left\lceil \frac{x}{y} \right\rceil$ | is the ceiling function of x divided by y |

Table 69 - Diagnostic Reporting and Recovery times

14.5.3. Over/Under Temperature Diagnostic

DIAG_TEMP_THR_HIGH defines the threshold for over temperature detection and is compared to the linearized value of the temperature sensor T_{LIN}. DIAG_TEMP_THR_LOW defines the threshold for under temperature detection and is compared to the linearized value of the temperature sensor T_{LIN}.

T_{LIN} is encoded using the SENT standard for temperature sensor. One can get the physical temperature of the die using following formula:

$$T_{PHY}[^{\circ}C] = \frac{T_{LIN}}{8} - 73.15$$

DIAG_TEMP_THR_LOW/HIGH are encoded on 8-bit unsigned values with the following relationship towards T_{Lin}

$$DIAG_TEMP_THR_(LOW/HIGH) = \frac{T_{LIN}}{16}$$

The value of T_{LIN} is clamped between the thresholds defined by DIAG_TEMP_THR_LOW and DIAG_TEMP_THR_HIGH.

Following table summarizes the characteristics of the linearized temperature sensor and the encoding of the temperature monitor thresholds.

| Parameter | Symbol | Min | Typ | Max | Unit | Condition |
|---|---------------------|-----|-------|-----|--------|---|
| T_{LIN} resolution | Res _{TLIN} | - | 0.125 | - | °C/LSB | |
| T_{LIN} refresh rate | F _{S,TLIN} | - | 200 | - | Hz | |
| T_{LIN} linearity error | T _{LinErr} | -8 | - | 8 | °C | from -40 to 160°C |
| T_{LIN} linearity error | T _{LinErr} | -2 | - | 6 | °C | from 35 to 125°C |
| High temperature threshold | DIAG_TEMP_THR_LOW | - | 8 | - | LSB | Recommended value, corresponds to -57°C |
| Low temperature threshold | DIAG_TEMP_THR_HIGH | - | 128 | - | LSB | Recommended value, corresponds to 183°C |
| High/low temperature threshold resolution | Res _{Tthr} | | 2 | | °C/LSB | |

Table 70 - Linearized Temperature Sensor characteristics

14.5.4. Field Strength and Field Monitoring Diagnostics

Field Strength is compensated over the circuit operating temperature range and represents a reliable image of the field intensity generated by the magnet. Field Strength value is optionally available in SENT secondary fast channel.

14.5.5. PWM Diagnostic

DC_FAULT

This parameter defines the duty-cycle that is present on the PWM output in case of diagnostic reporting.

DC_FIELDTOOLOW

This parameter defines the duty-cycle that is output in case of Field Too Low, from 0% till 100 % by steps of (100/256)%.

15. Functional Safety

15.1. Safety Manual

The safety manual, available upon request, contains the necessary information to integrate the MLX90377 component in a safety related item, as Safety Element Out-of-Context (SEoC).

In particular, it includes:

- The description of the Product Development lifecycle tailored for the Safety Element.
- An extract of the Technical Safety concept.
- The description of Assumptions-of-Use (AoU) of the element with respect to its intended use, including:
 - assumption on the device safe state;
 - assumptions on fault tolerant time interval and multiple-point faults detection interval;
 - assumptions on the context, including its external interfaces;
- The description of safety analysis results (at the device level, to be used for the system integration), HW architectural metrics and description of dependent failures initiators.
- The description and the result of the functional safety assessment process; list of confirmation measures and description of the independency level.

15.2. Safety Mechanisms

The MLX90377 provides numerous self-diagnostic features (safety mechanisms). Those features increase the robustness of the IC functionality either by preventing the IC from providing an erroneous output signal or by reporting the failure according to the SENT protocol definition.

| Legend |
|--|
| ● High coverage |
| ○ Medium coverage |
| ANA : Analog hardware failure reporting, described in the safety manual |
| High-Z : Special reporting, output is set in high impedance mode (no HW fail-safe mode/timeout, no SW safe startup) |
| DIG : Digital hardware failure reporting, described in the safety manual |
| At Startup : HW fault present at time zero is detected before a first frame is transmitted. |
| DIAG_EN : This safety mechanism can be disabled by setting DIAG_EN = 0 (see chapter 12 End User Programmable Items). This option should not be used in application mode! |

Table 71 - Self Diagnostic Legend

| Category and safety mechanism name | Front-end | ADC | DSP | Back-end | Support. Func. | Module & Package | Reporting mode | At startup | DIAG EN |
|--|-----------|-----|-----|----------|----------------|------------------|----------------|------------|---------|
| Signal-conditioning (AFE, External Sensor) Diagnostic | | | | | | | | | |
| Magnetic Signal Conditioning Voltage Test Pattern | ● | ○ | ○ | | | | ANA | NO | ● |
| Magnetic Signal Conditioning Rough Offset Clipping check | ● | | ○ | | | | ANA | NO | ● |
| Magnetic Signal Conditioning Gain Monitor | ● | | ○ | | | ● | ANA | YES | ● |
| Magnetic Signal Conditioning Gain Clamping | ● | | ○ | | | ● | ANA | YES | ● |
| Mag. Sig. Cond. Failure control by the chopping technique | ● | | | | | | n/a | YES | |
| External Sensor Sig. Cond. Voltage Valid Range Check | ● | | | | | ● | ANA | YES | ● |
| External Sensor Sig. Cond. Frequency Valid Range Check | ● | | | | | ● | ANA | YES | ● |
| A/D Converter Test Pattern | | ● | | | | | ANA | NO | ● |
| ADC Conversion errors & Overflow Errors | | ● | | | | | ANA | YES | ● |
| Flux Monitor (Specific to Rotary mode) | ● | ○ | | | | ● | ANA | NO | ● |
| HE Switching Circuitry Symmetry Check | ● | | | | | | ANA | NO | ● |
| HE Bias Current Supply Monitor | ● | | | | | | ANA | NO | ● |
| Stress Sensor Valid Range Check | ● | | | | | | ANA | YES | ● |
| Digital-circuit Diagnostic | | | | | | | | | |
| RAM Parity, 1 bit per 16 bits word, ISO D.2.5.2 | | | ● | | | | DIG | YES | ● |
| ROM Parity, 1 bit per 32 bits word, ISO D.2.5.2 | | | ● | | | | DIG | YES | ● |

| Category and safety mechanism name | Front-end | ADC | DSP | Back-end | Support. Func. | Module & Package | Reporting mode | At startup | DIAG EN |
|--|-----------|-----|-----|----------|----------------|------------------|----------------|------------|---------|
| "NVRAM 16 bits signature (run-time) ISO D.2.4.3, by means of SW CRC-CCITT16" | | | ● | | | | DIG | NO | ● |
| NVRAM Double Error Detection ECC ISO | | | ● | | | | DIG | YES | |
| Logical Monitoring of program sequence ISO D.2.9.3 via Watchdog "IWD" (cpu clock) ISO D2.9.2 | | | ● | | ○ | | DIG | NO | ● |
| Watchdog "AWD" (separate clock) ISO D2.9.1 | | | ● | | ○ | | DIG | YES | |
| CPU Errors "Invalid Address", "Wrong opcode" | | | ● | | ○ | | DIG | YES | |
| ADC Interface Checksum | | ● | ○ | | | | DIG | NO | ● |
| DSP Test Pattern (atan2) | | | ● | | ○ | | DIG | NO | ● |
| Critical ports monitoring | | | ● | | | | DIG | NO | ● |
| DSP Overflow | ○ | ○ | ● | | | | DIG | YES | ● |
| ADC data adder test | | ○ | ● | | | | DIG | YES | ● |
| SENT Fall Collision detection (SENT pulse generator) | | | ● | | | | DIG | NO | ● |
| DAC Overflow | | | ● | | | | DIG | NO | ● |
| Communication Interface Diagnostic | | | | | | | | | |
| SENT parity check over Configuration registers | | | | | ● | | DIG | NO | ● |
| SENT block: Protection against re-configuration at run-time | | | | | ● | | DIG | NO | ● |
| SENT Frame Counter & Redundant Nibble | | | | | ● | | n/a | n/a | |
| System-level diagnostic | | | | | | | | | |

| Category and safety mechanism name | Front-end | ADC | DSP | Back-end | Support. Func. | Module & Package | Reporting mode | At startup | DIAG EN |
|---|-----------|-----|-----|----------|----------------|------------------|----------------|------------|---------|
| Supply Voltage Monitors (all supply domains) except VS_OV & POR | | | | | ● | ● | ANA | YES | ● |
| External Supply Overvoltage Monitor VS_OV | | | | | ● | ● | High-Z | YES | |
| Digital Supply under-voltage monitor (Power-on reset) | | | | | ● | ● | High-Z | YES | |
| Supply Bias Current Monitor | | | | | ● | | ANA | NO | ● |
| Overheating monitor | ○ | ○ | ○ | ○ | ○ | ● | ANA | YES | ● |
| Warning/Reporting Mechanisms | | | | | | | | | |
| HW Error Controller | | | ● | ● | ● | | DIG | YES | |
| HW Fail-safe mode with timeout | | | ● | ● | ● | | DIG | YES | |
| Analog-type Error management | ● | ● | | | ● | | ANA | NO | ● |
| Safe start-up mode | | | ● | | ● | | DIG | n/a | |
| Mechanisms executed at start-up only | | | | | | | | | |
| RAM March-C HW Test at start-up | | | ● | | ● | | DIG | YES | |

Table 72 - MLX90377 List of Self Diagnostics with Characteristics

15.3. Fault Handling Time Interval

| Name | Definition |
|------------------------|--|
| DTI _{ANA} | Time to run one full analog diagnostic cycle (programmable value) |
| DTI _{DIG} | Time to run one full nvram BIST cycle (programmable value) |
| τ _R | Output refresh period in analog output mode (see “Timings definition” chapter) |
| T _{pwm_input} | the period of the incoming PWM signal on gateway pin |
| T _{trigger} | the time between 2 SPC trigger pulses |
| DTI | Diagnostic Time Interval: worst-case time between 2 consecutive runs of a specific diagnostic |
| EXE | Execution Time: worst-case time between the measurement start of a given diagnostic and the availability of the result of this measurement |
| REP | Reporting Time: worst-case time needed between an internal diagnostic error event and a switch of the output to fail-safe state |
| FHTI | Fault Handling Time Interval: the time interval between the start of the first frame with invalid position value without notice, and the end of the last frame preceding a fail-safe state of the IC. It is calculated as the sum of DTI, EXE, and REP figures. The following equation is valid for any diagnostic: $FHTI = DTI + EXE + REP$ |

Table 73 – Glossary of Terms for the Definition Regarding Fault Handling Time Interval

The following table details the cycle time, execution time and reporting time for all monitors included in MLX90377. In all cases, the worst-case timing values are provided in the table below:

| Category and safety mechanism name | DTI | EXE | REP |
|--|----------------------------|--------------------|--------------------|
| Signal-conditioning (AFE, External Sensor) Diagnostic | | | |
| Magnetic Signal Conditioning Voltage Test Pattern | 1 * DTI _{ANA} | 1 * τ _R | 1 * τ _R |
| Magnetic Signal Cond. Rough Offset Clipping check | 2.5 ms | 1 * τ _R | 1 * τ _R |
| Magnetic Signal Conditioning Gain Monitor & clamp | 2.5 ms | 1 * τ _R | 1 * τ _R |
| External Sensor Sig. Cond. Voltage Valid Range Check | 9 * τ _R | 0 | 1 * τ _R |
| External Sensor Sig. Cond. Frequency Valid Range Check | 1 * T _{pwm_input} | 1 * τ _R | 1 * τ _R |
| A/D Converter Test Pattern | 1 * DTI _{ANA} | 0 | 1 * τ _R |
| ADC Conversion errors & Overflow Errors | 1 * τ _R | 0 | 1 * τ _R |
| Flux monitor | 2.5 ms | 1 * τ _R | 1 * τ _R |
| Hall-element & Mechanical-stress Tests | 1 * DTI _{ANA} | 0 | 1 * τ _R |
| Digital-circuit Diagnostic | | | |
| RAM Parity, 1 bit per 16 bits word, ISO D.2.5.2 | < 1 * DTI _{ANA} | 0 | 0 |
| ROM Parity, 1 bit per 16 bits word, ISO D.2.5.2 | < 1 * DTI _{ANA} | 0 | 0 |
| NVRAM Double Error Detection ECC ISO D.2.4.1 | < 10μs | 0 | 0 |

MLX90377 - Triaxis Position Processor

Datasheet

| Category and safety mechanism name | DTI | EXE | REP |
|--|--------------------------|--------------|--------------|
| NVRAM 16 bits signature (run-time) ISO D.2.4.3 | 1 * DTI _{DIG} | 0 | 0 |
| Logical Monitoring of program sequence ISO D.2.9.3 via Watchdog "IWD" (cpu clock) ISO D2.9.2 | 4 ms | 0 | 0 |
| Watch-dog "AWD" (separate clock) ISO D2.9.1 | 5ms | 0 | 0 |
| CPU error detection | 1 * τ_R | 0 | 0 |
| ADC Interface Checksum | 1 * DTI _{ANA} | 0 | 0 |
| DSP Test Pattern (atan2) | 1 * DTI _{ANA} | 0 | 0 |
| Critical ports monitoring | 1 * DTI _{ANA} | 0 | 0 |
| DSP overflow | 1 * τ_R | 0 | 0 |
| ADC data adder test | 1 * τ_R | 0 | 0 |
| SENT Fall Collision detection & DAC Overflow | 1 * τ_R | 0 | 0 |
| SENT H/W Interface Diagnostic | | | |
| SENT parity check over Configuration registers | 1 * T _{trigger} | 0 | 0 |
| SENT HW: Protection against re-configuration | 1 * T _{trigger} | 0 | 0 |
| System-level diagnostic | | | |
| Supply Voltage Monitors except VS_OV & POR | 1 * DTI _{ANA} | 0 | 1 * τ_R |
| External Supply Overvoltage Monitor VS_OV | 1 * DTI _{ANA} | 0 | 1 * τ_R |
| Supply Bias Current Monitor | 1 * DTI _{ANA} | 0 | 1 * τ_R |
| Overheating monitor | 9 * τ_R | 1 * τ_R | 1 * τ_R |

Table 74 - Timings of the cyclic detection mechanism

Note:

1. **The orange colour coding** shows the worst-case FHTI for analog and digital diagnostics
2. The ROM and RAM parity mechanism trigger a fail-safe state on the failing ROM or RAM word is accessed. (It is assumed that it will be accessed within one full diagnostic cycle in worst case)
3. EXE/REP time = 0 means that the execution or reporting time is negligible compared to the FHTI value

16. Recommended Application Diagrams

This chapter shows the typical application diagrams for the MLX90377 in different packages, which should not be considered as a specification but rather a recommendation. It is up to the end user to define the proper application circuit, based on the exact use-case, i.e. output protocol, power mode, etc.

16.1. Wiring with the MLX90377 in SOIC-8 Package

Figure 34 shows the typical recommended application circuit for the MLX90377 in SOIC-8 package, and Table 75 shows the corresponding value of the external components.

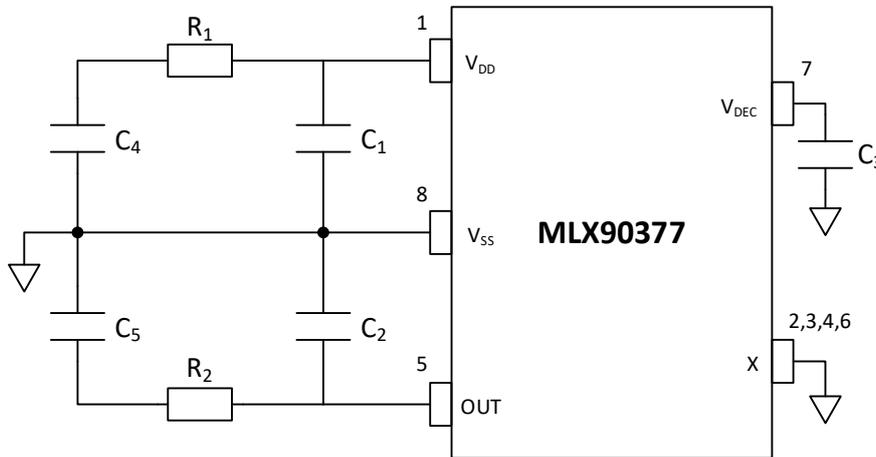


Figure 34 - Recommended wiring for the MLX90377 in SOIC-8 package

| Component | Value | Remark |
|----------------------------------|-------|------------------------------|
| C ₁ | 220nF | Place close to the IC pin |
| C ₂ (C _L) | 100nF | for analog output |
| | 10nF | for SENT and PWM output |
| | 2.2nF | for SPC output |
| C ₃ | 100nF | Place close to the IC pin |
| C ₄ | 0nF | Place close to the connector |
| C ₅ | 0nF | Place close to the connector |
| R ₁ | 0Ω | |
| R ₂ | 0Ω | |

Table 75 - Recommended Values for the MLX90377 in SOIC-8 Package

C₂ should consider the total capacitance on the bus. C₄, C₅, R₁, R₂ are not needed under typical conditions. They are only needed, when extremely high electromagnetic immunity (EMI) compliance is required, but the value of the resistors and capacitors should not exceed 10Ω and 1nF.

In SENT protocol, an external pi-filter to improve radiated emission performance is not needed. The MLX90377 has a built-in high order low pass filter. Therefore, any additional external filter will deteriorate the generated SENT signal, and could make the output signal not comply to the SENT specifications, such as the fall times and the minimum output voltages.

16.2. Wiring with the MLX90377 in TSSOP-16 Package

Figure 35 shows the typical recommended application circuit for the MLX90377 in TSSOP-16 package, and Table 76 shows the corresponding value of the external components.

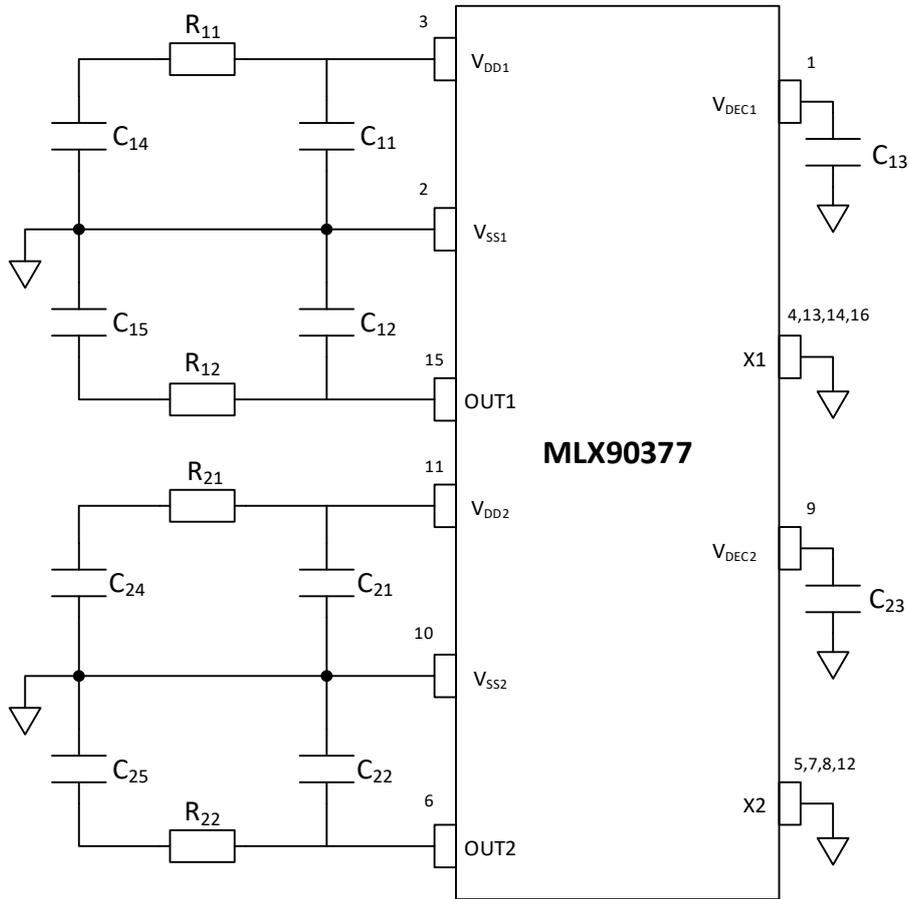


Figure 35 - Recommended wiring for the MLX90377 in TSSOP-16 dual-die package

| Component | Value | Remark |
|-----------------|-------|------------------------------|
| C _{x1} | 220nF | Place close to the IC pin |
| C _{x2} | 100nF | for analog output |
| | 10nF | for SENT and PWM output |
| | 2.2nF | for SPC output |
| C _{x3} | 100nF | Place close to the IC pin |
| C _{x4} | 0nF | Place close to the connector |
| C _{x5} | 0nF | Place close to the connector |
| R _{x1} | 0Ω | |
| R _{x2} | 0Ω | |

Table 76 - Recommended Values for the MLX90377 in TSSOP-16 dual-die Package

C_{x2} should consider the total capacitance on the bus. C_{x4} , C_{x5} , R_{x1} , R_{x2} are not needed under typical conditions. They are only needed, when extremely high electromagnetic immunity (EMI) compliance is required, but the value of the resistors and capacitors should not exceed 10Ω and $1nF$.

In SENT or SPC protocol, an external pi-filter to improve radiated emission performance is not needed. The MLX90377 has a built-in high order low pass filter. Therefore, any additional external filter will deteriorate the generated output signal, and could make the output signal not comply to the SENT or SPC specifications, such as the fall times and the minimum output voltages.

16.3. Wiring with the MLX90377 in DMP-4 Package (built-in capacitors)

Figure 36 shows the typical recommended application circuit for the MLX90377 in DMP-4 package, and Table 77 shows the corresponding value of the external components.

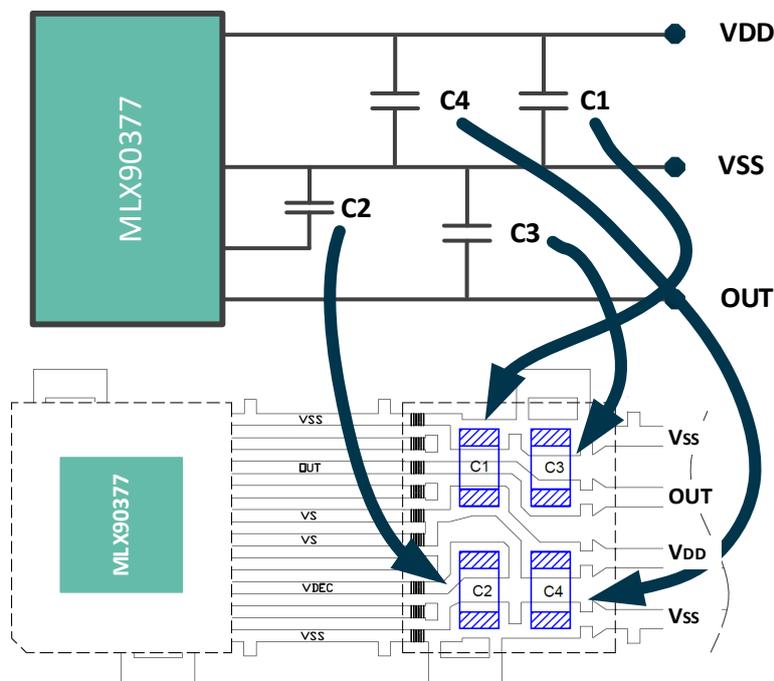


Figure 36 - Internal wiring of the MLX90377 in DMP-4

| Component | Value | Remark |
|-----------|-------|-------------|
| C1 | 220nF | |
| C2 | 100nF | |
| C3 | 10nF | |
| C4 | 0 | not mounted |

Table 77 - DMP-4 capacitors configuration

16.4. Wiring with the MLX90377 in SMP-3 Package (built-in capacitors)

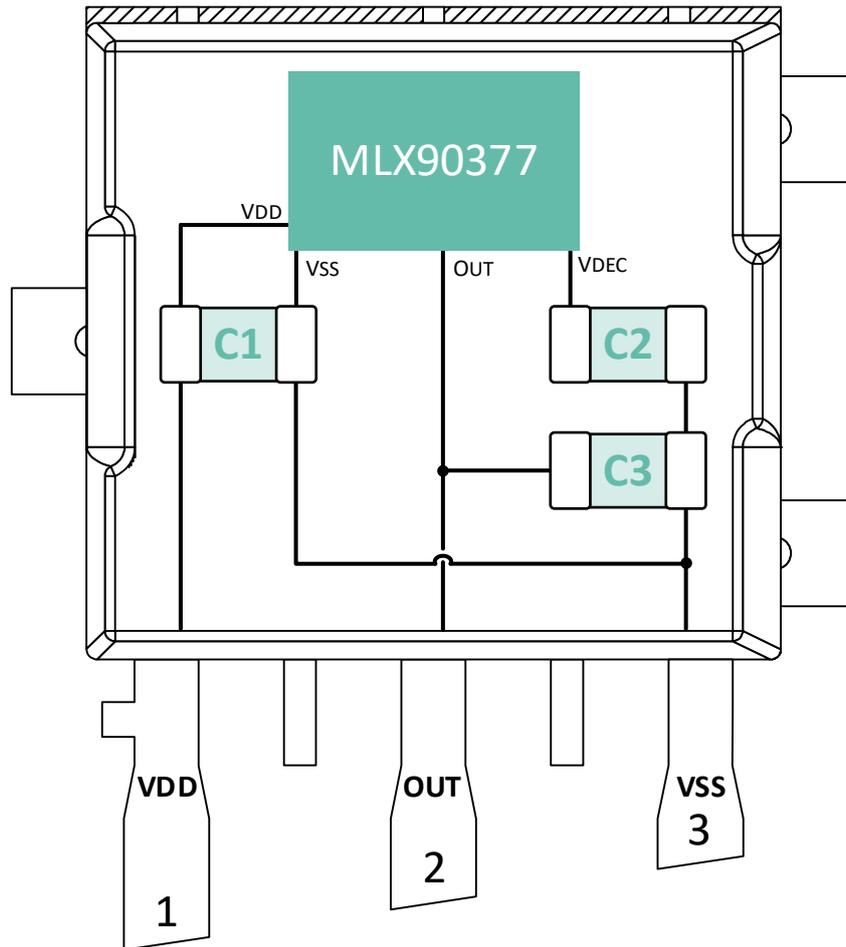


Figure 37 - Internal wiring of the MLX90377 in SMP-3

| Component | Value | Remark |
|-----------|--------|--------|
| C1 | 220 nF | |
| C2 | 100 nF | |
| C3 | 10 nF | |

Table 78 - SMP-3 capacitors configuration

17. Standard Information Regarding Manufacturability of Melexis Products with Different Soldering Processes

Our products are classified and qualified regarding soldering technology, solderability and moisture sensitivity level according to standards in place in Semiconductor industry.

For further details about test method references and for compliance verification of selected soldering method for product integration, Melexis recommends reviewing on our web site the General Guidelines soldering recommendation (<http://www.melexis.com/en/quality-environment/soldering>)

For all soldering technologies deviating from the one mentioned in above document (regarding peak temperature, temperature gradient, temperature profile etc), additional classification and qualification tests have to be agreed upon with Melexis.

For package technology embedding trim and form post-delivery capability, Melexis recommends consulting the dedicated trim & form recommendation application note : “Lead Trimming and Forming Recommendations” (<http://www.melexis.com/en/documents/documentation/application-notes/lead-trimming-and-forming-recommendations>).

Melexis is contributing to global environmental conservation by promoting lead free solutions. For more information on qualifications of RoHS compliant products (RoHS = European directive on the Restriction Of the use of certain Hazardous Substances) please visit the quality page on our website: <http://www.melexis.com/en/quality-environment>.

18. ESD Precautions

Electronic semiconductor products are sensitive to Electro Static Discharge (ESD).

Always observe Electro Static Discharge control procedures whenever handling semiconductor products.

19. Package Information

19.1. SOIC-8 Package

19.1.1. SOIC-8- Package Dimensions

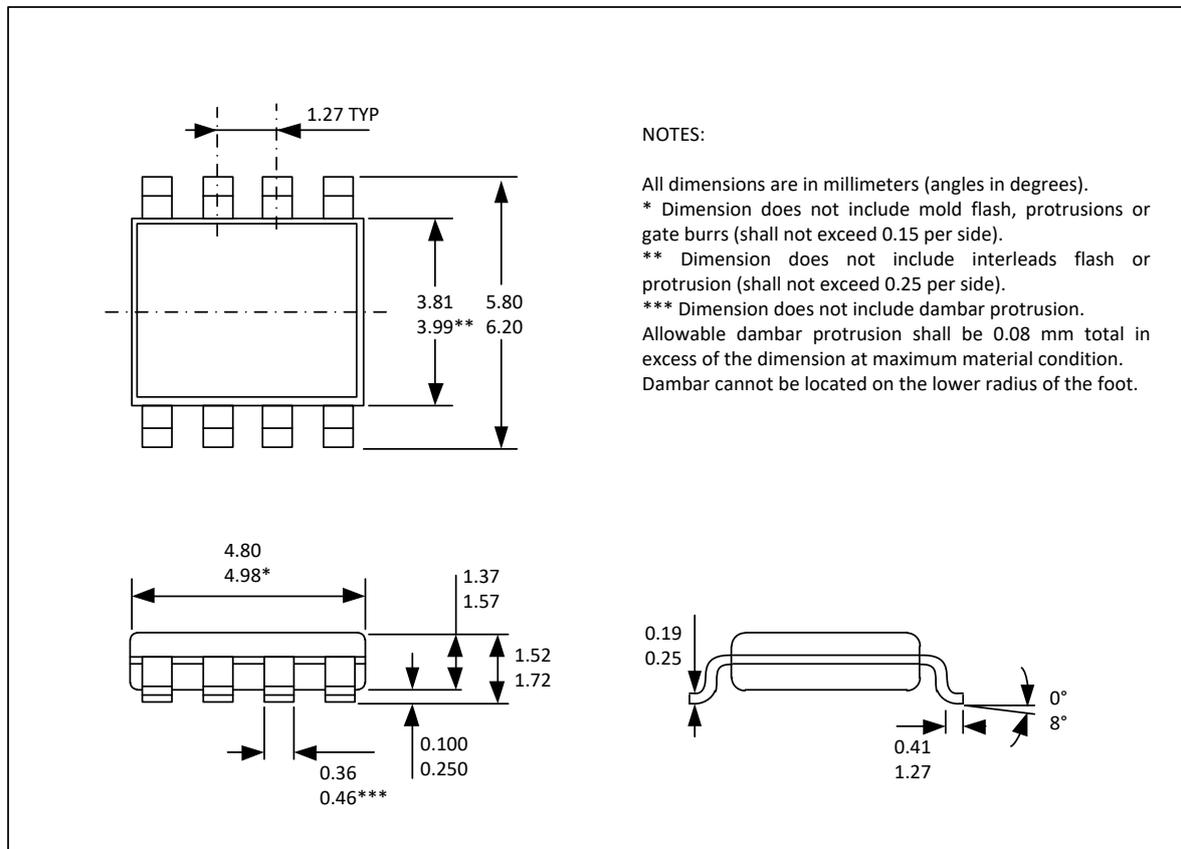


Figure 38 - SOIC-8 Package Outline Dimensions

19.1.2. SOIC-8- Pinout and Marking

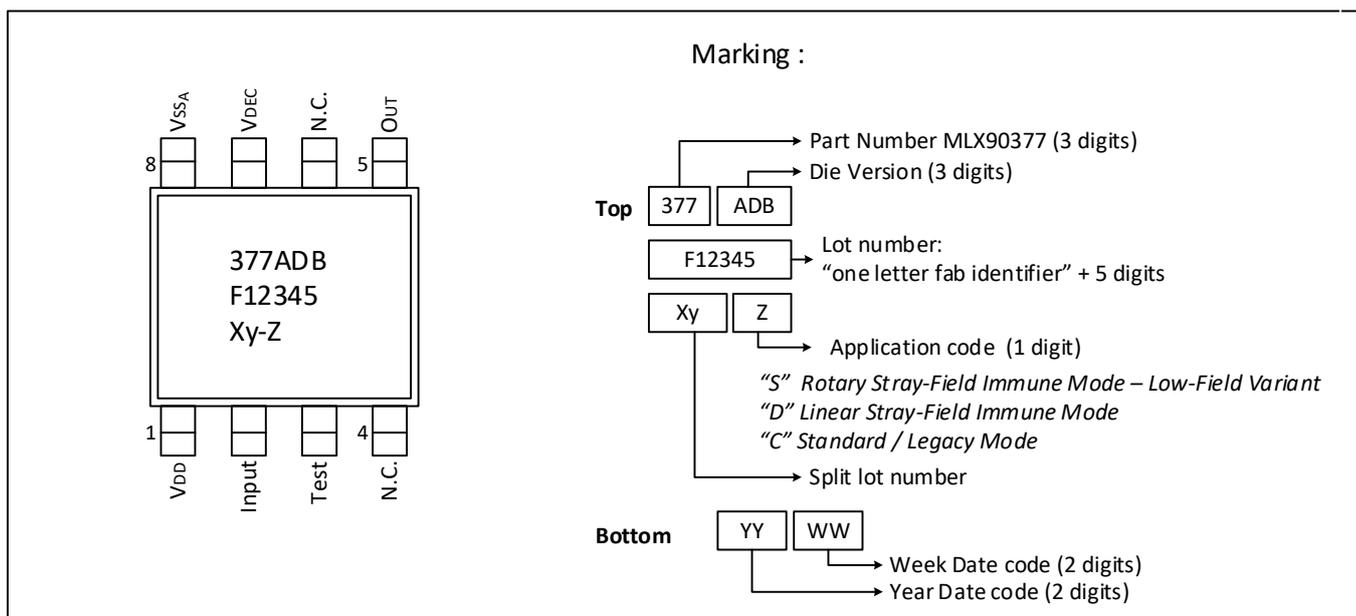


Figure 39 - SOIC-8 Pinout and Marking

19.1.3. SOIC-8- Sensitive spot positioning

19.1.3.1. Rotary Stray-field Immune and Standard Mode Applications

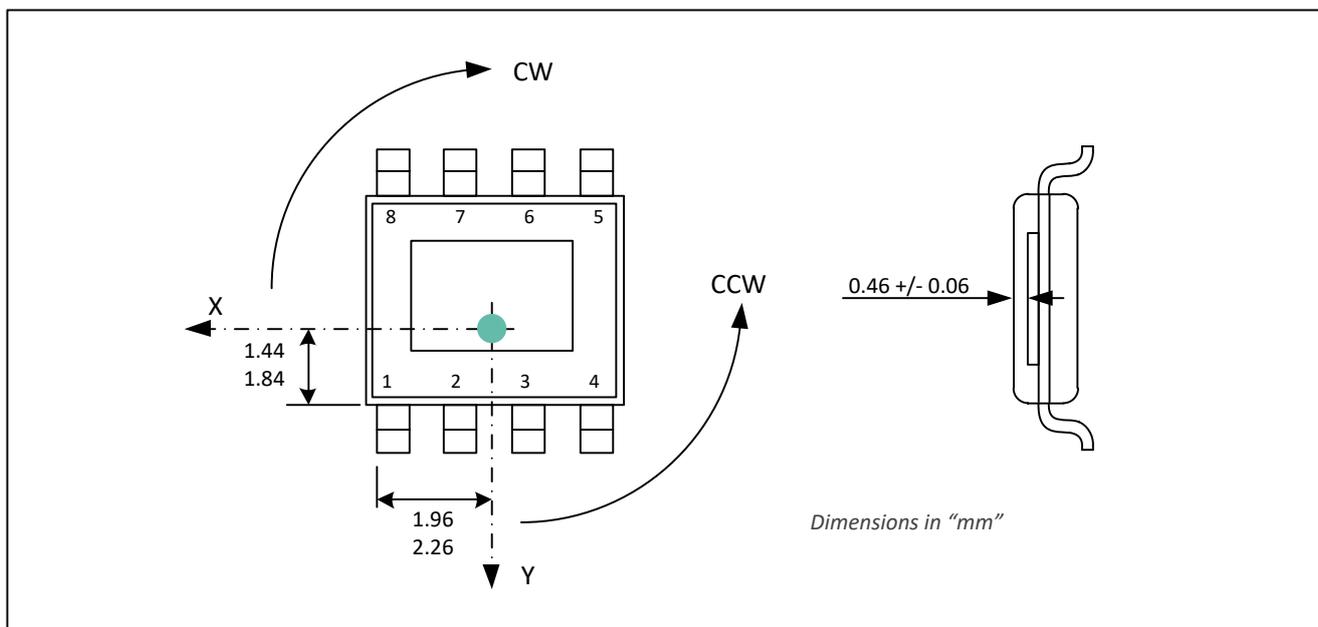


Figure 40 - SOIC-8 Sensitive Spot Position

19.1.3.2. Linear Stray-field Immune Applications

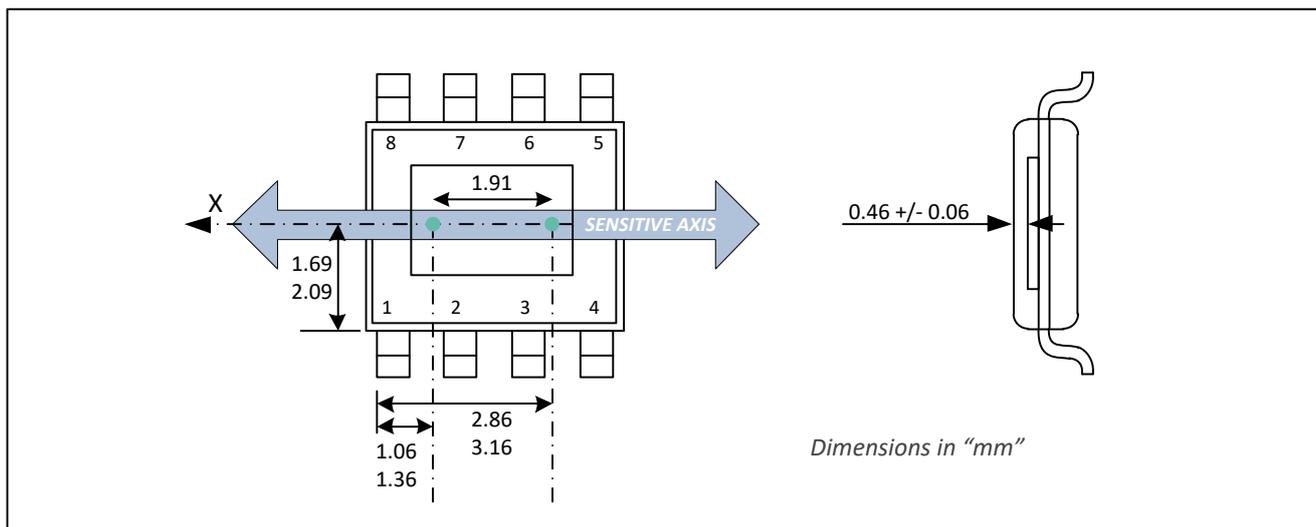


Figure 41 - SOIC-8 Sensitive Spot position for Linear Stray-Field Immune

19.1.4. SOIC-8- Angle detection

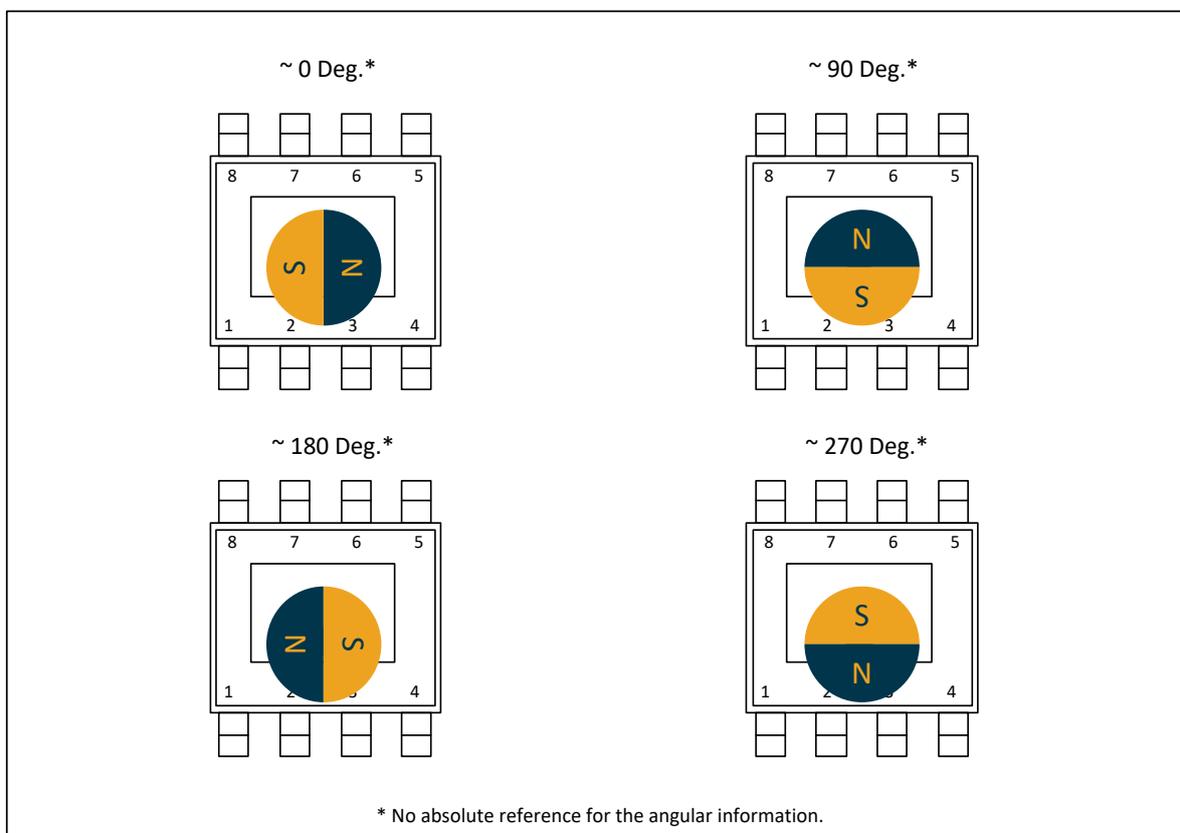


Figure 42 - SOIC-8 Angle Detection

The MLX90377 is an absolute angular position sensor but the linearity error (See section 9) does not include the error linked to the absolute reference 0 Deg (which can be fixed in the application through the discontinuity point).

19.2. TSSOP-16 Package

19.2.1. TSSOP-16- Package Dimensions

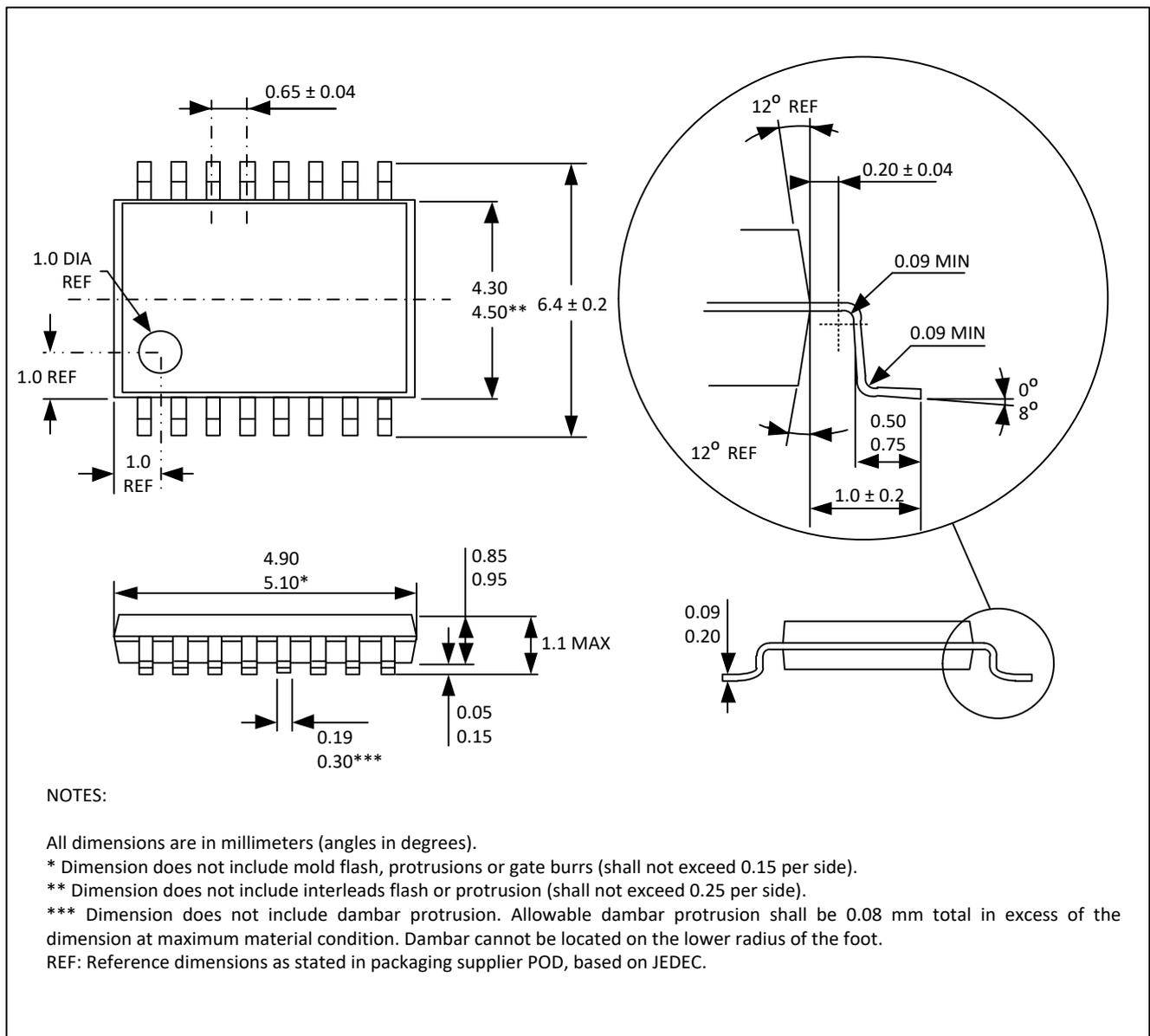


Figure 43 - TSSOP-16 Package Outline Dimensions

19.2.2. TSSOP-16 Dual-Die Package- Pinout and Marking

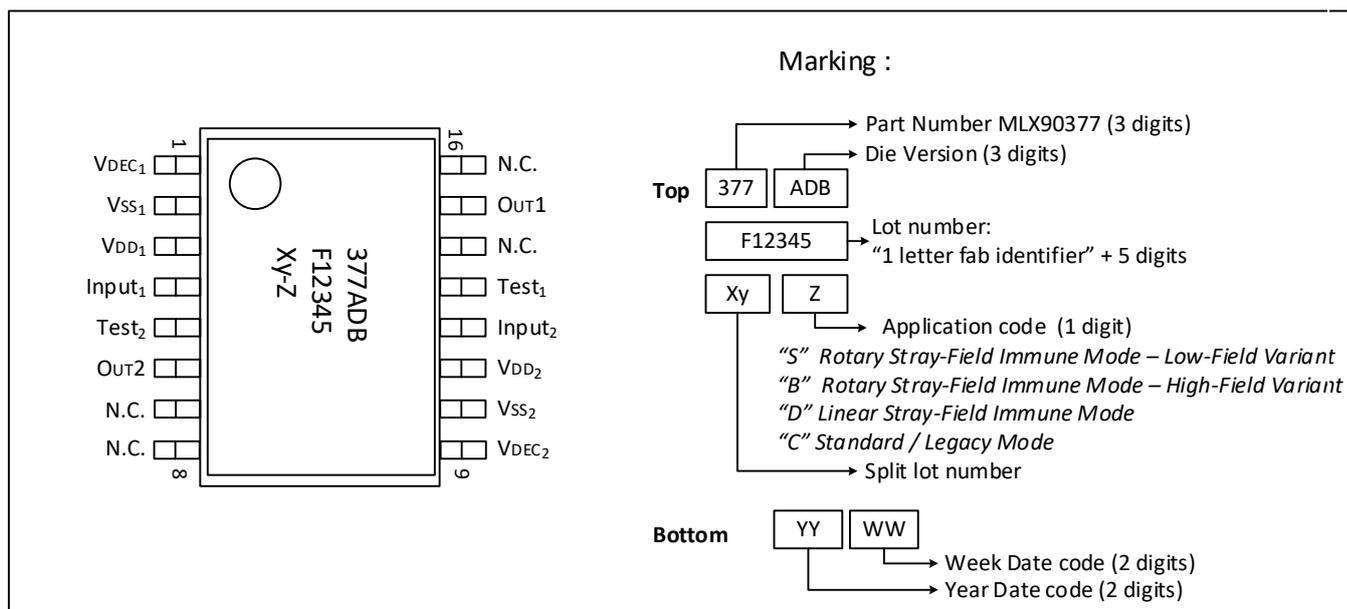


Figure 44 - TSSOP-16 Dual-Die Package Pinout and Marking

19.2.3. TSSOP-16 Dual-Die Package- Sensitive spot positioning

19.2.3.1. Rotary Stray-field Immune and Standard Mode applications

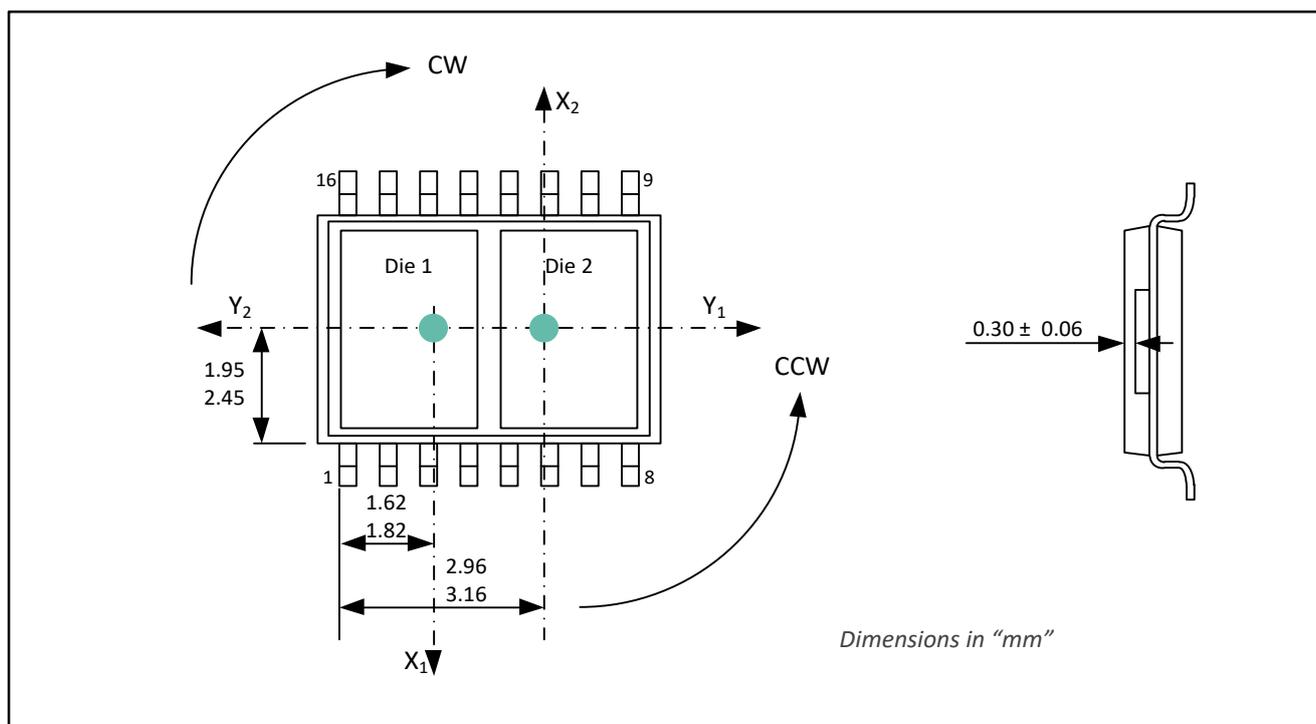


Figure 45 - TSSOP-16 dual-die package Sensitive Spot Position

19.2.3.2. Linear Stray-field Immune Applications

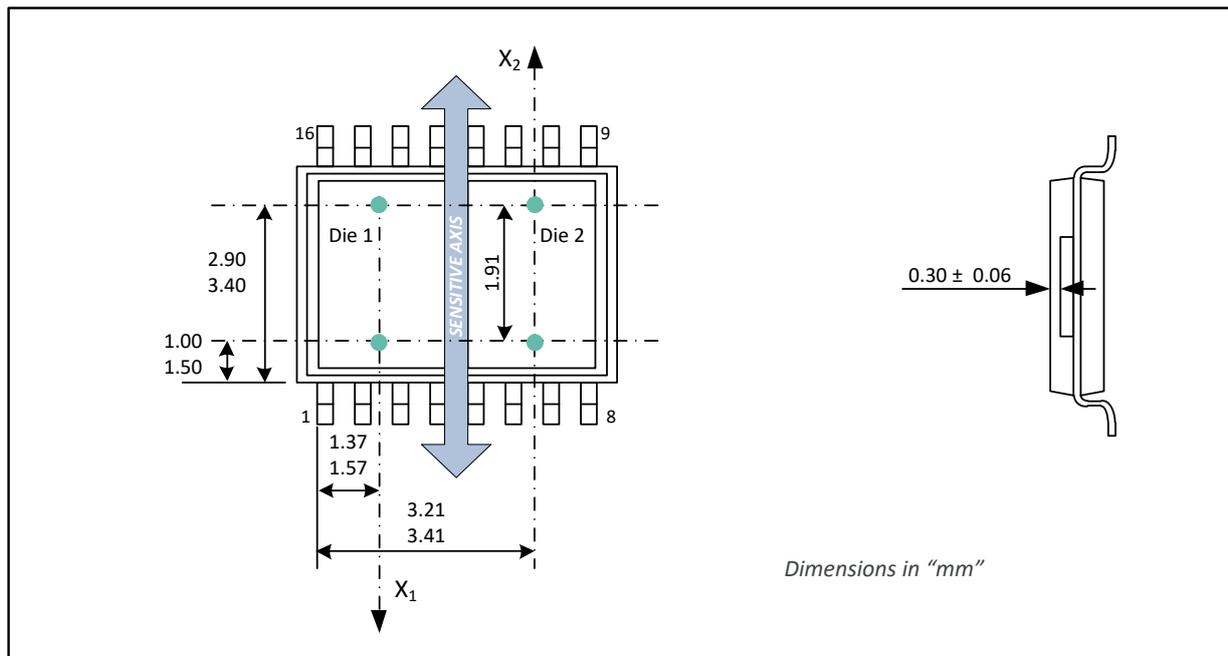


Figure 46 - TSSOP-16 dual-die package - Sensitive Spot Location for Linear Stray-field Immune

19.2.4. TSSOP-16- Angle Detection

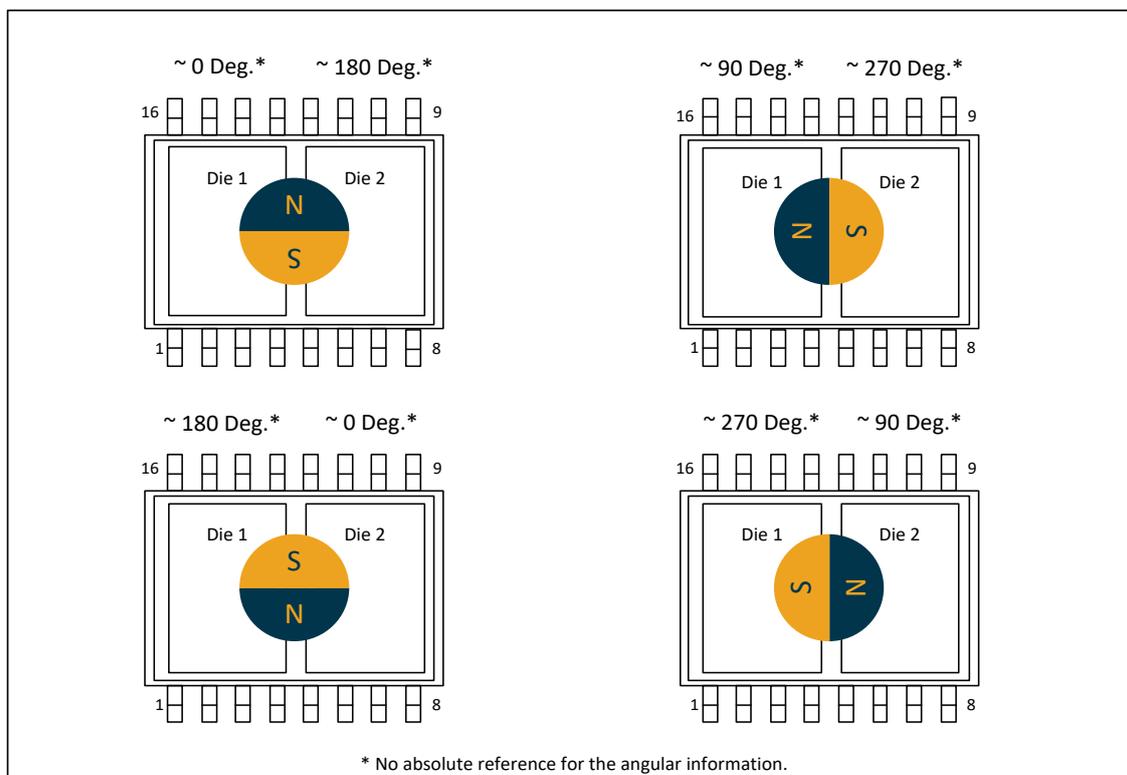


Figure 47 - TSSOP-16 Angle Detection

The MLX90377 is an absolute angular position sensor but the linearity error (see section 9) does not include the error linked to the absolute reference 0Deg (which can be fixed in the application through the discontinuity point).

19.3. DMP-4- Package

19.3.1. DMP-4- Package Outline Dimensions (POD)- Straight Leads

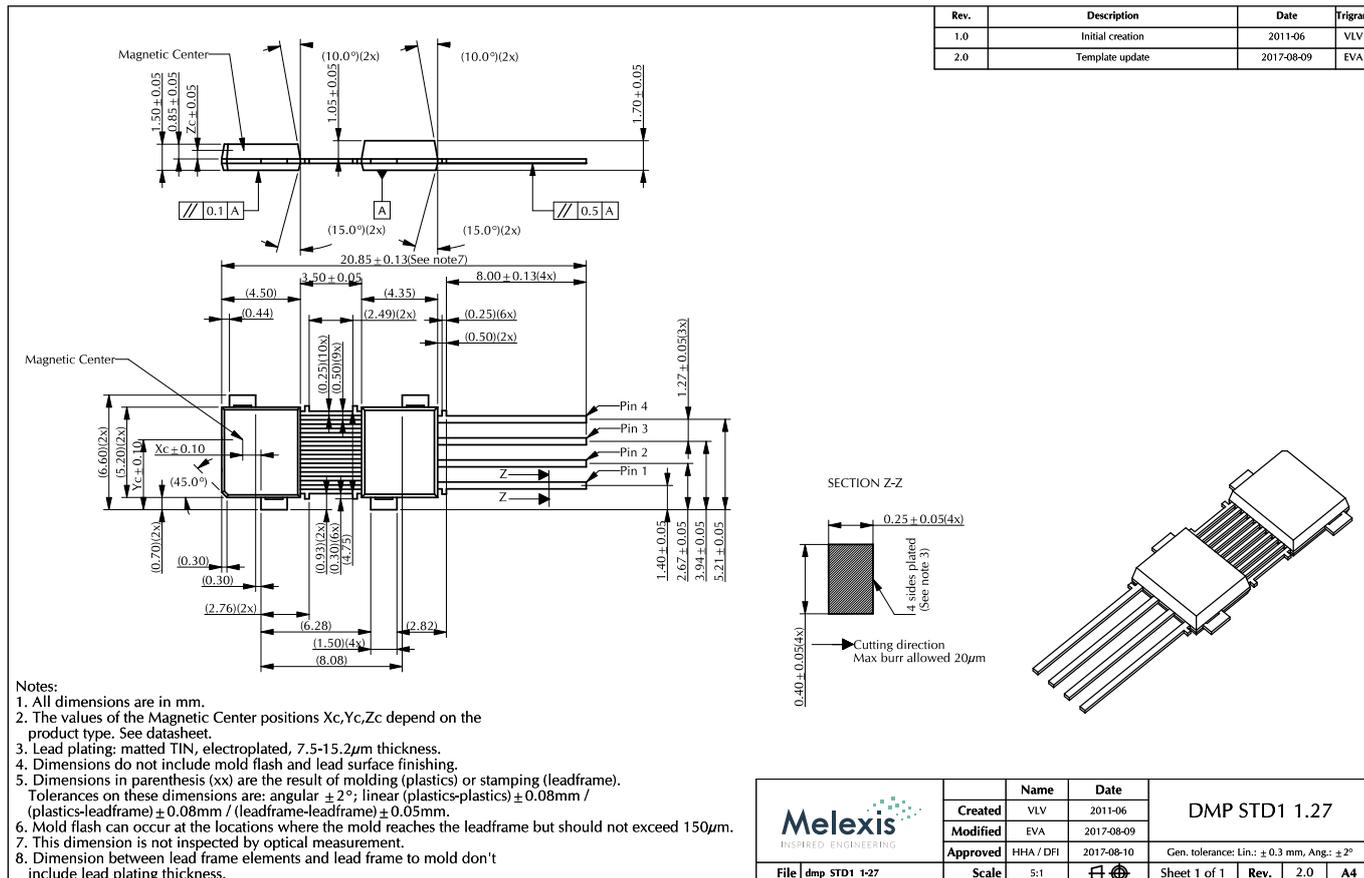


Figure 48 - DMP-4 Straight Leads Package Outline Drawing

19.3.2. DMP-4- Package Outline Dimensions (POD)- STD2 2.54

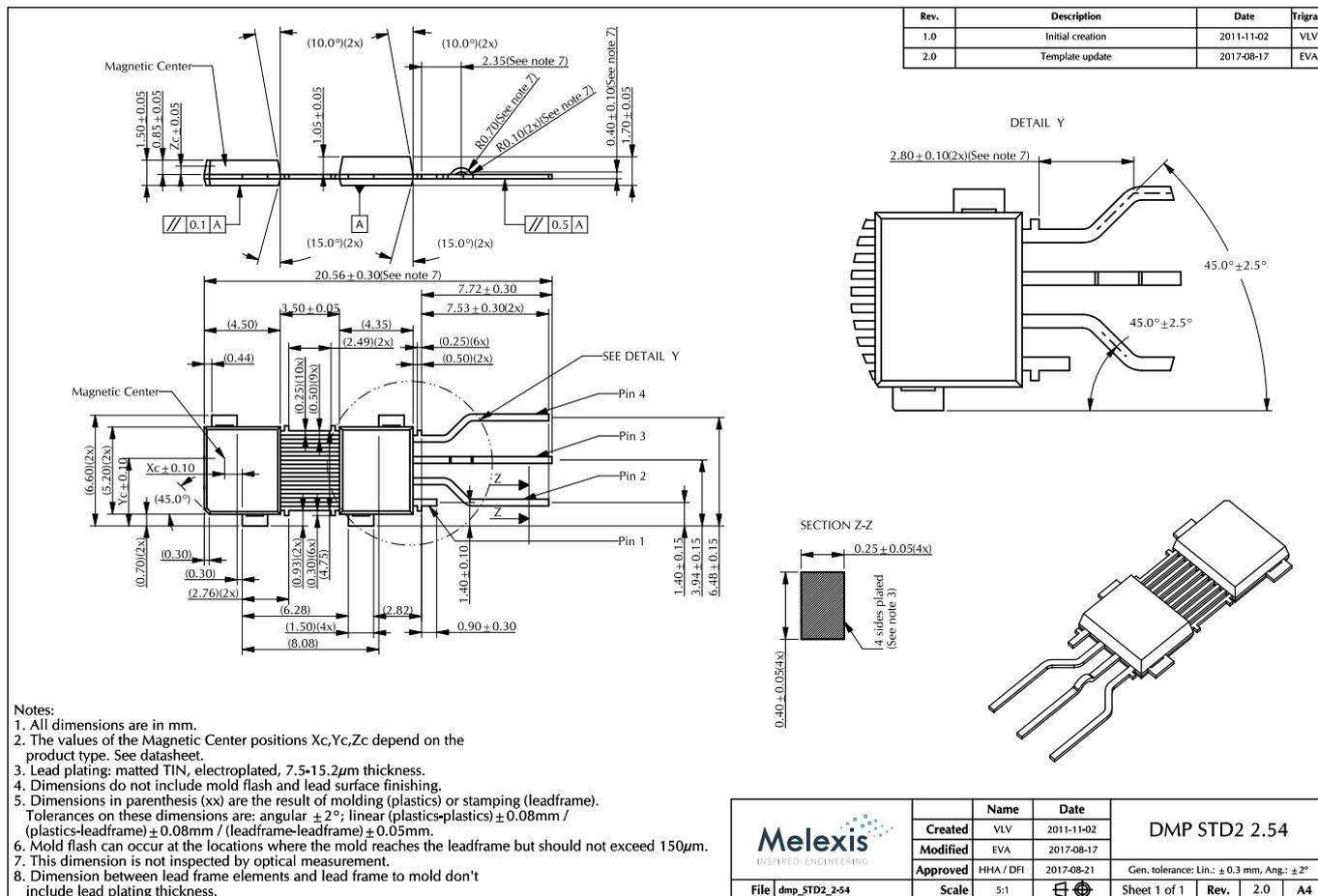


Figure 49 - DMP-4 STD2 2.54 Package Outline Drawing

19.3.3. DMP-4- Package Outline Dimensions (POD) STD4 2.54

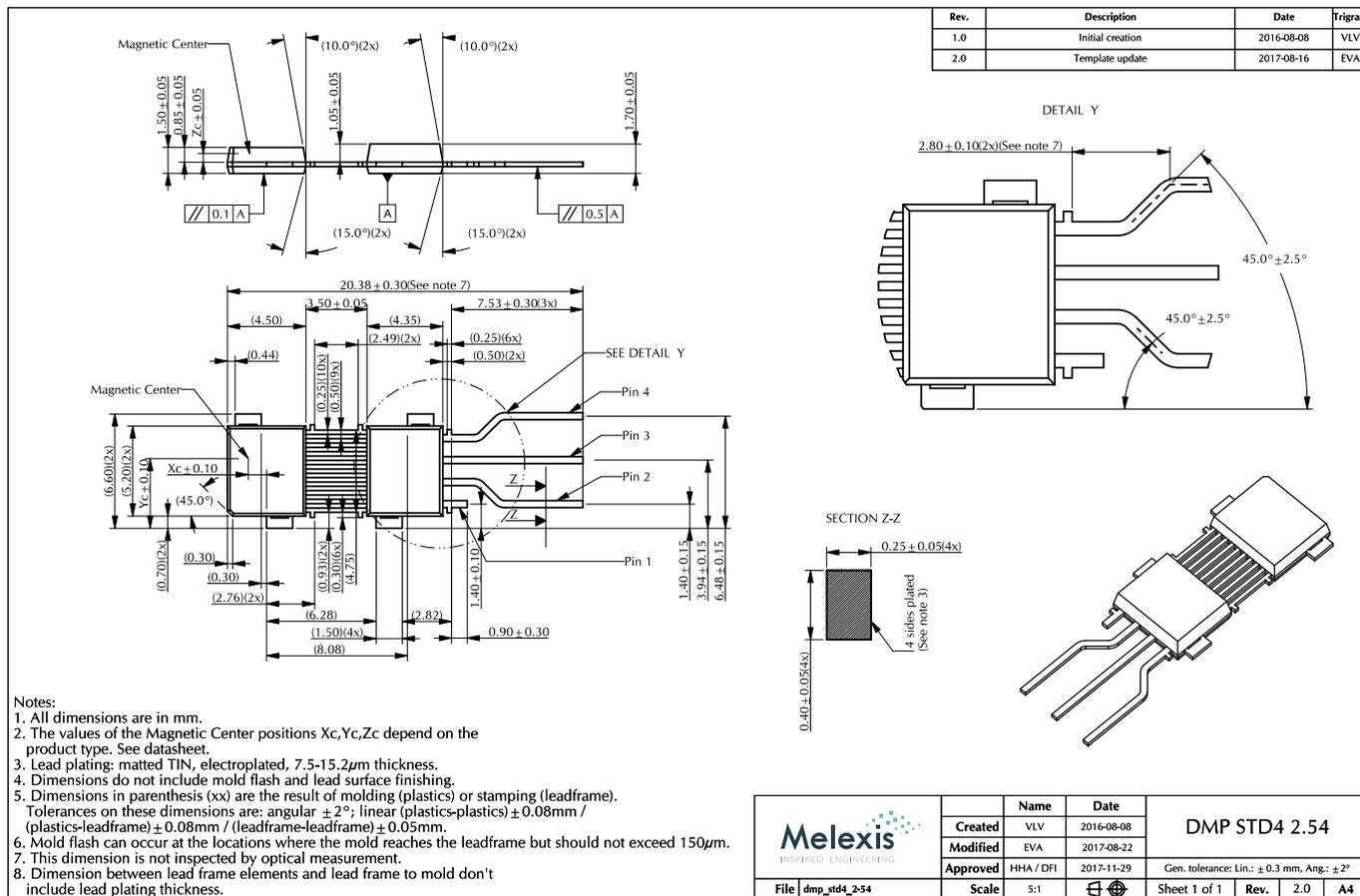


Figure 50 - DMP-4 STD4 2.54 Package Outline Drawing

19.3.4. DMP-4- Marking

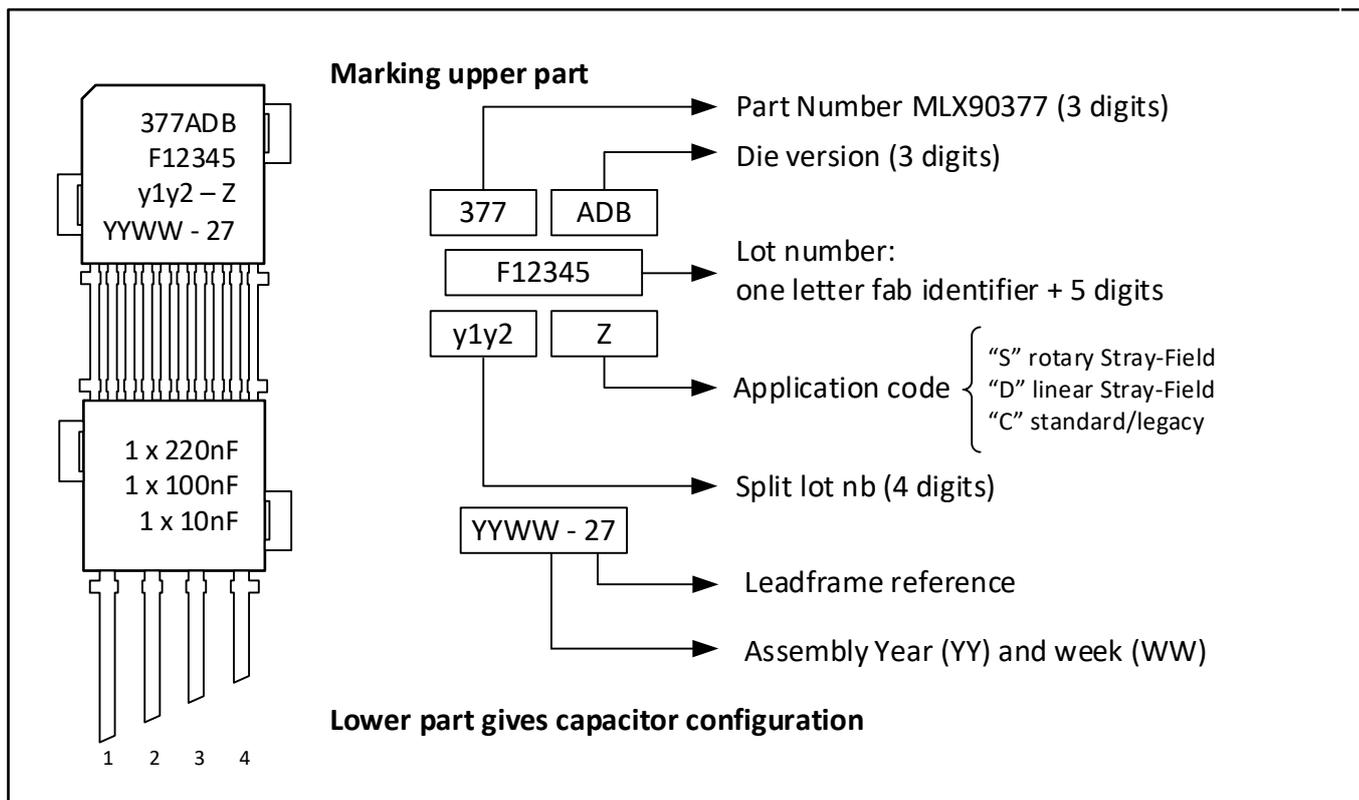


Figure 51 - DMP-4 Marking

19.3.5. DMP-4- Sensitive Spot Positioning

19.3.5.1. Rotary Stray-field Immune or Standard Mode Applications

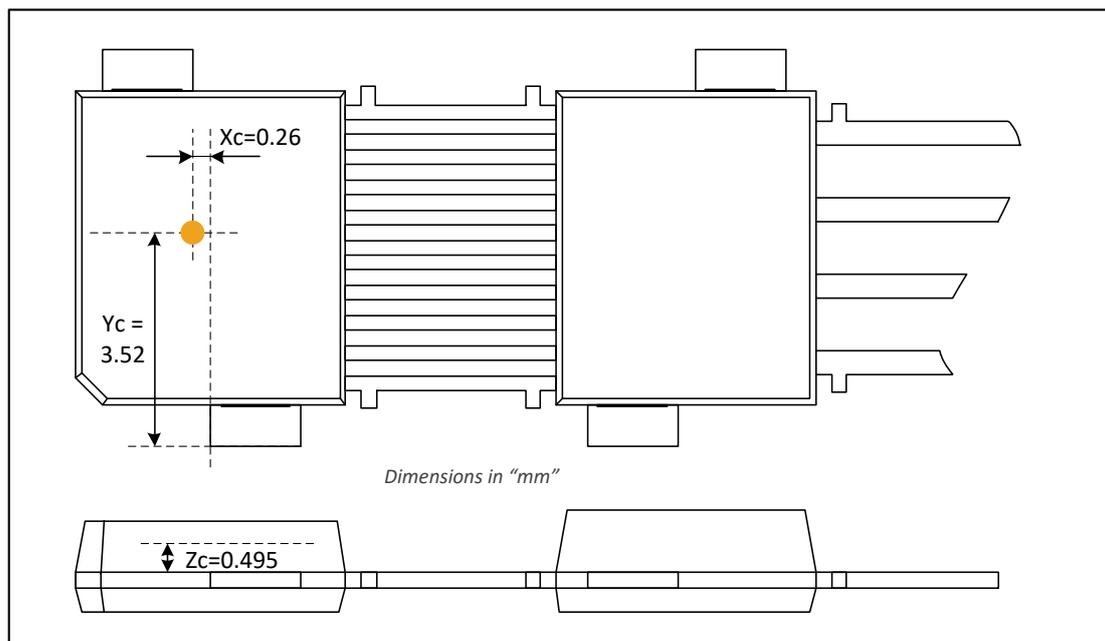


Figure 52 - DMP-4 Rotary Stray-field or legacy Sensitive Spot Position

19.3.5.2. Linear Stray-field Immune Applications

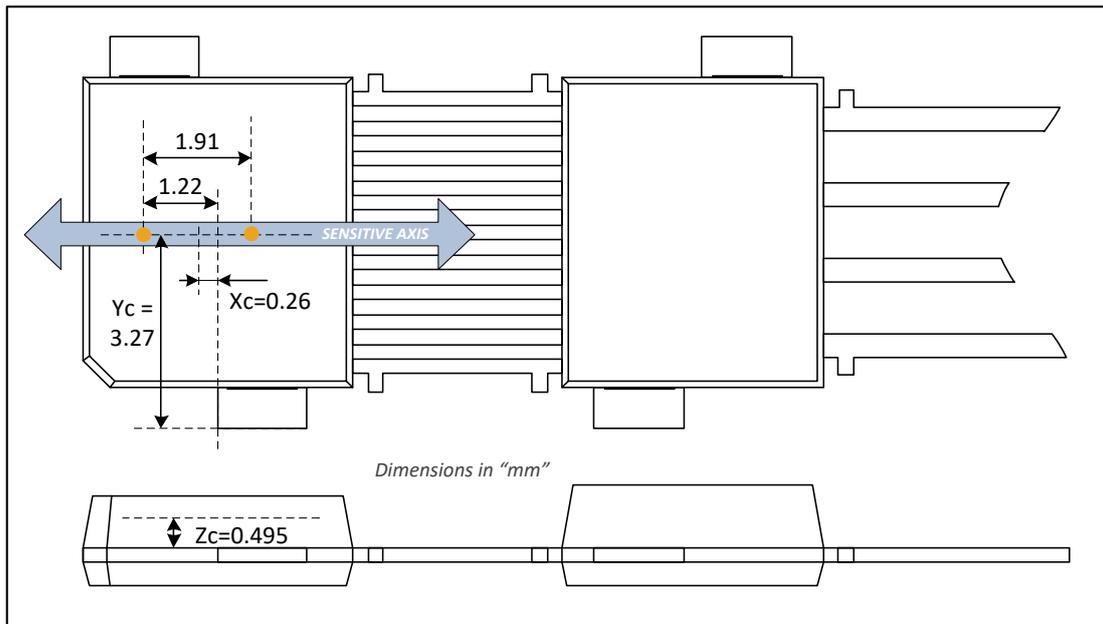


Figure 53 - DMP-4 Rotary Stray-field or legacy Sensitive Spot Position

19.3.6. DMP-4- Angle detection MLX90377 DMP-4

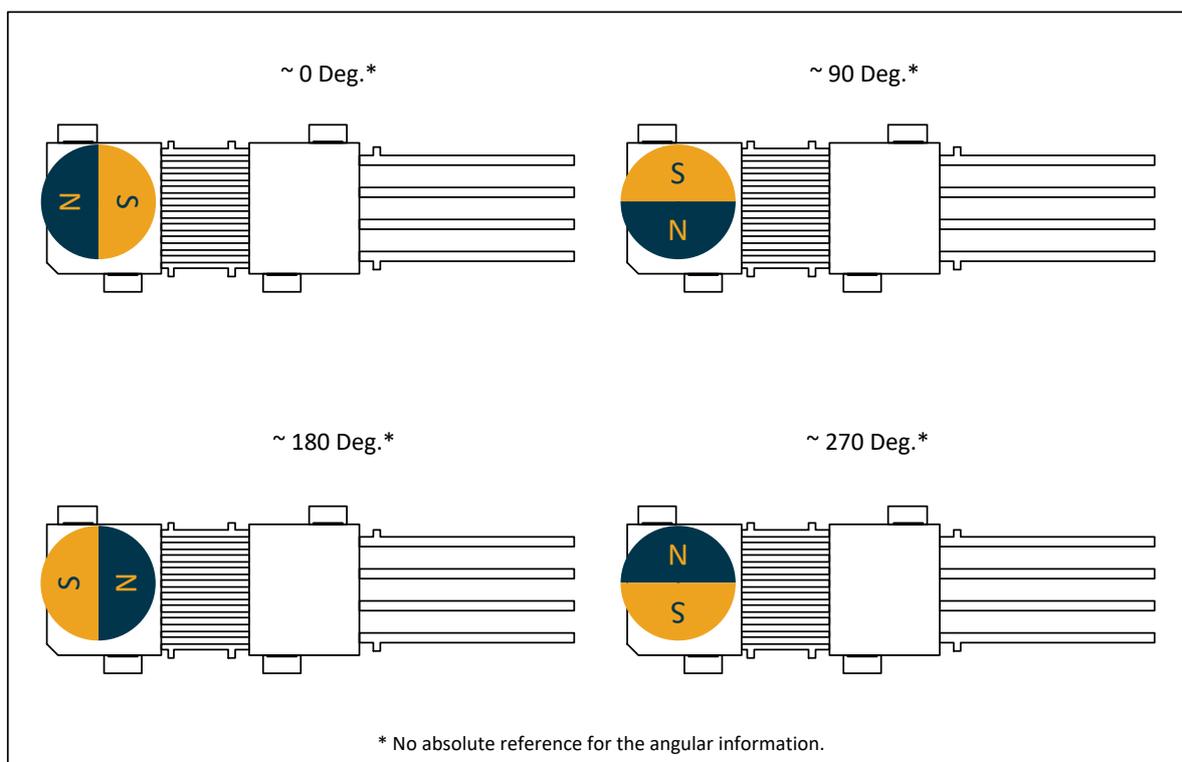
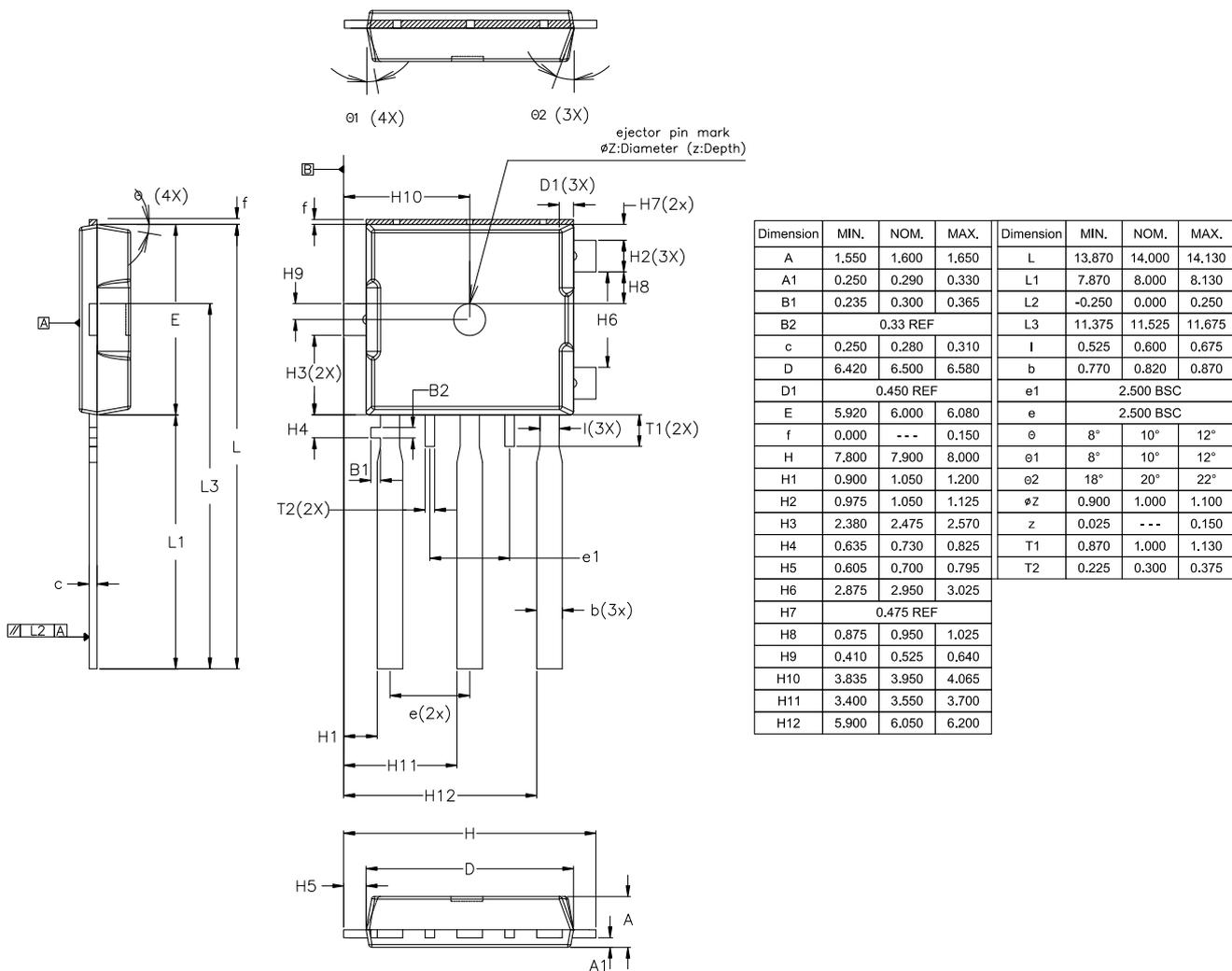


Figure 54 - DMP-4 Angle Detection

The MLX90377 is an absolute angular position sensor but the linearity error (see section 9) does not include the error linked to the absolute reference 0 Deg (which can be fixed in the application through the discontinuity point).

19.4. SMP-3- Package

19.4.1. SMP-3- Package Outline Dimensions (POD)



NOTES:

1. DIMENSIONS ARE IN MILLIMETER UNLESS NOTED OTHERWISE.

⚠ PACKAGE WIDTH DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS. MOLD FLASH, PROTRUSIONS OR GATE BURRS SHALL NOT EXCEED 0.15MM PER END. PACKAGE LENGTH DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSION. INTERLEAD FLASH OR PROTRUSION SHALL NOT EXCEED 0.25MM PER SIDE.

⚠ THE PACKAGE TOP MAY BE SMALLER THAN THE PACKAGE BOTTOM. PACKAGE WIDTH AND LENGTH ARE DETERMINED AT THE OUTERMOST EXTREMES OF THE PLASTIC BODY EXCLUSIVE OF MOLD FLASH, TIE BAR BURRS, GATE BURRS AND INTERLEAD FLASH.

4. PLATING SPECS: MATTED TIN, ELECTROPLATED, 12 ± 5 MICROMETER (µm) THICKNESS

5. ALL "EARS" ARE CONNECTED TO ELECTRIC GROUND.

Figure 55 - SMP-3 Package Outline Drawing

19.4.2. SMP-3- Marking

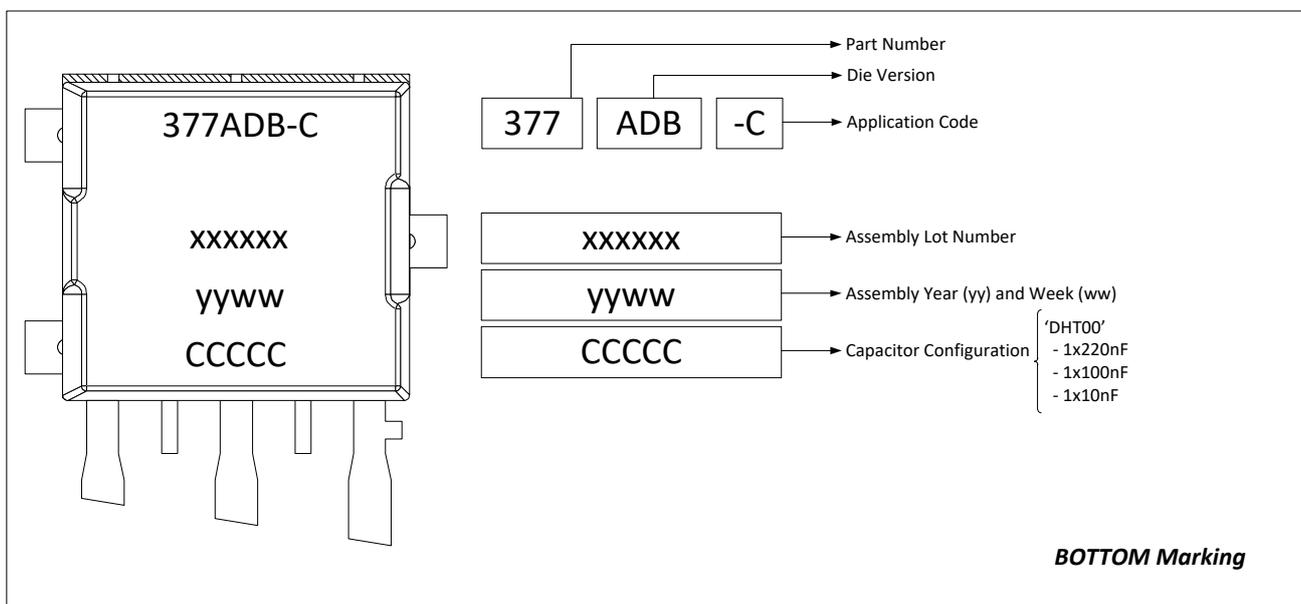
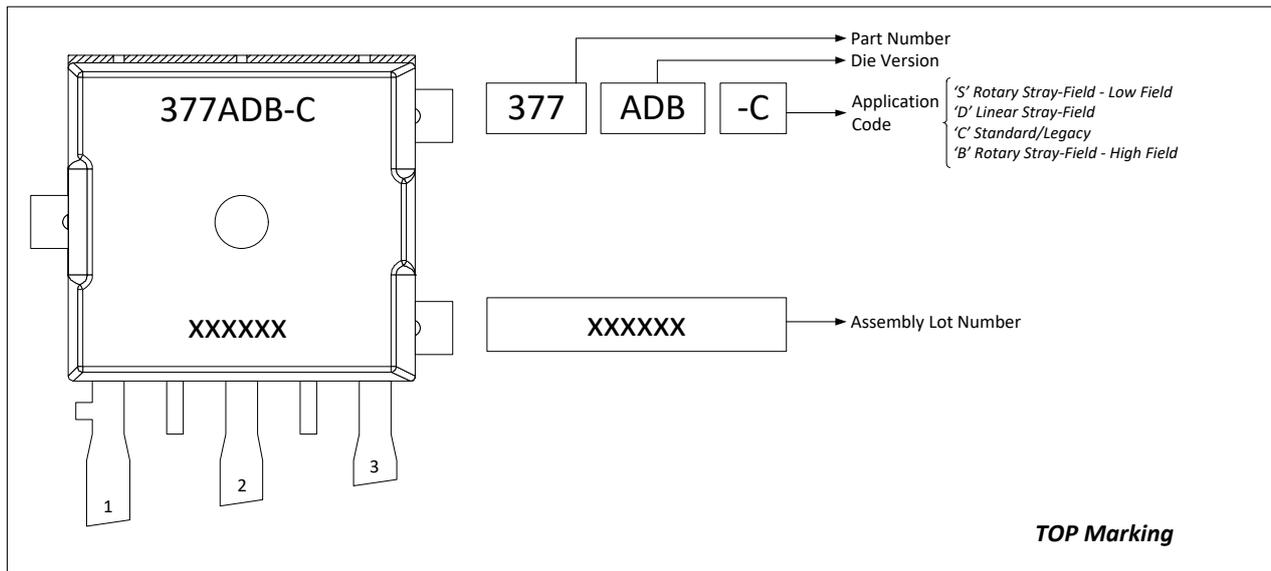


Figure 56 - SMP-3 Marking

19.4.3. SMP-3- Sensitive Spot Positioning

19.4.3.1. Rotary Stray-field Immune or Standard Mode Applications

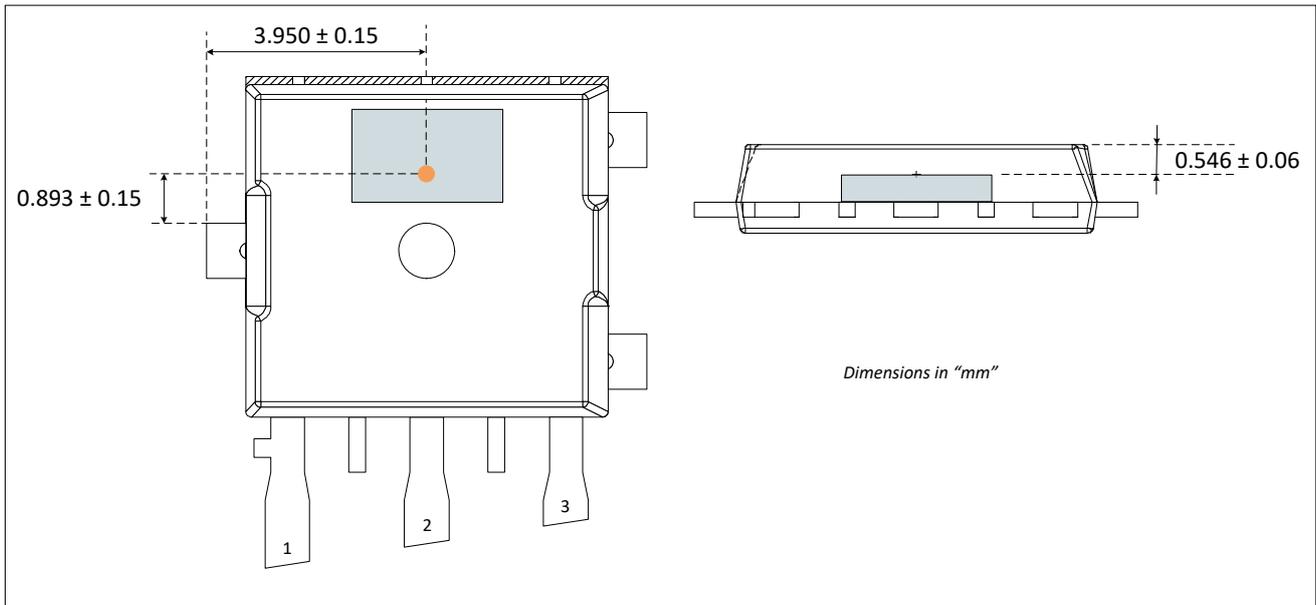


Figure 57 - SMP-3 Rotary Stray-field or legacy Sensitive Spot Position

19.4.3.2. Linear Stray-field Immune Applications

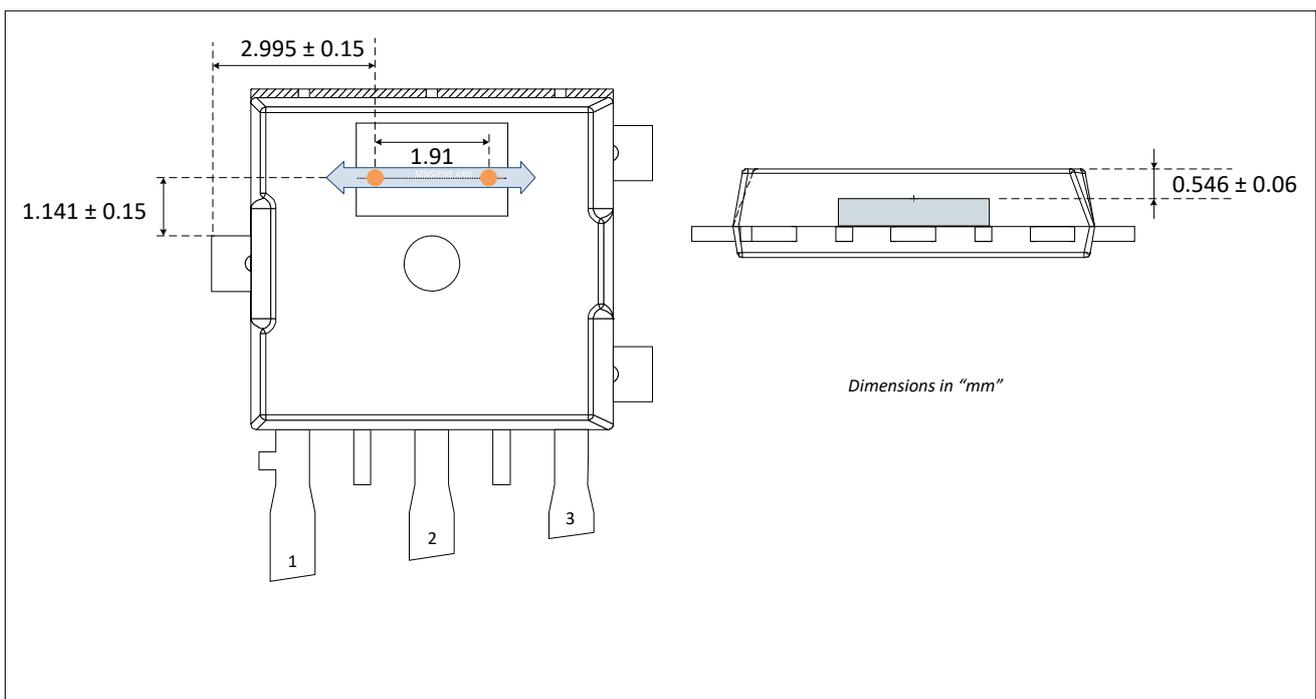


Figure 58 - SMP-3 Rotary Stray-field or legacy Sensitive Spot Position

19.4.4. SMP-3- Angle detection MLX90377 SMP-3

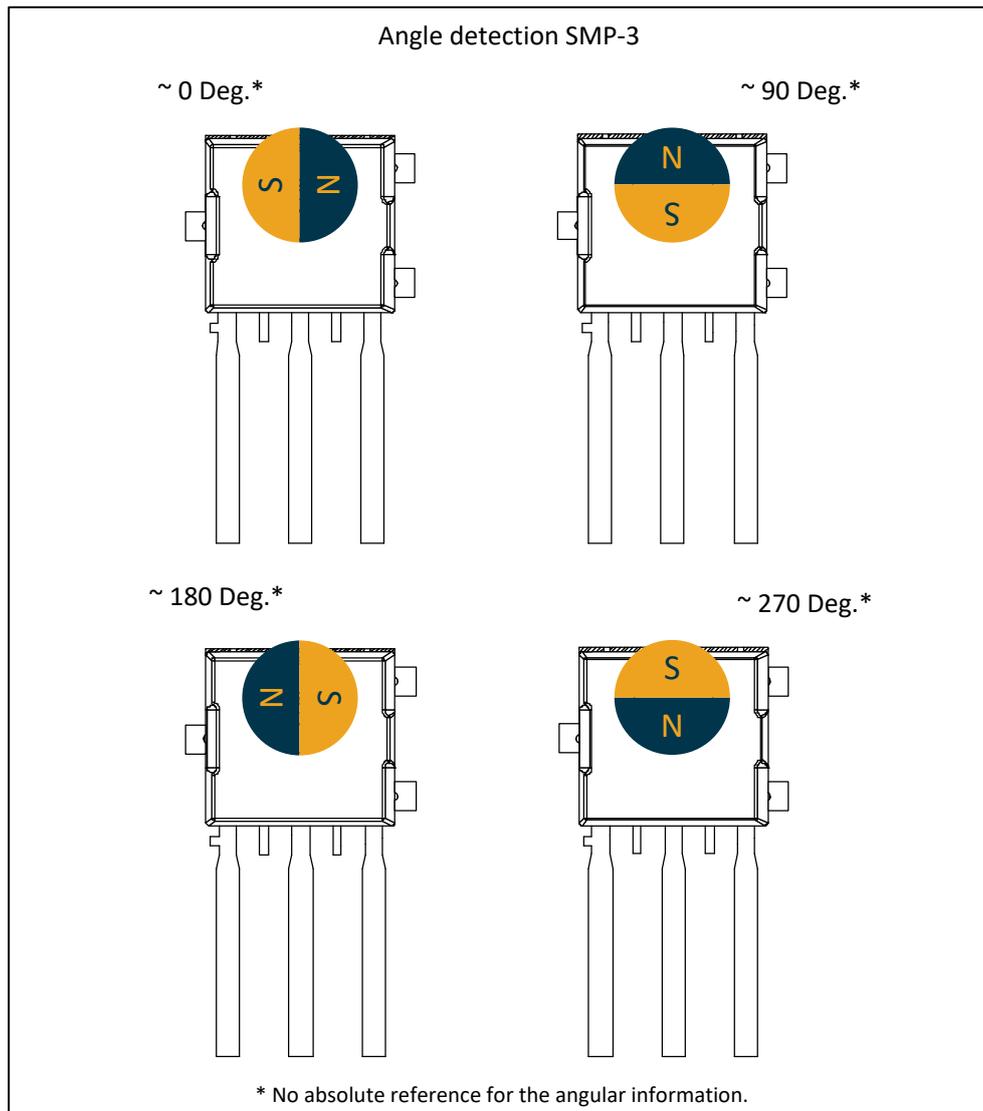


Figure 59 - SMP-3 Angle Detection

The MLX90377 is an absolute angular position sensor but the linearity error (see section 9) does not include the error linked to the absolute reference 0 Deg (which can be fixed in the application through the discontinuity point).

19.5. Packages Thermal Performances

The table below describe the thermal behaviour of available packages following JEDEC EIA/JESD 51.X standard.

| Package | Junction to case - θ_{jc} | Junction to ambient - θ_{ja} (JEDEC 1s2p board) | Junction to ambient - θ_{ja} (JEDEC 1s0p board) |
|----------|----------------------------------|--|--|
| SOIC-8 | 38.8 K/W | 112 K/W | 153 K/W |
| TSSOP-16 | 27.6 K/W | 99.1 K/W | 137 K/W |
| DMP-4 | 32.2 K/W | - | 88.7 K/W ⁽⁴⁵⁾ |
| SMP-3 | 34.4 K/W | - | 206 K/W ⁽⁴⁵⁾ |

Table 79 - Standard Packages Thermal Performances

⁴⁵ PCB-less solution has been evaluated in a typical application case. Values for this package are given as informative.

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